



A PARAMETER-FREE MAP IMAGE RECONSTRUCTION ALGORITHM FOR IMPULSE-BASED UWB GROUND PENETRATING RADAR

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Overview

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- Maximum a Posteriori Estimation
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- Conclusions and Acknowledgement





Motivation

- Landmines are self-contained explosive devices that detonate when triggered by a person or vehicle
- Factors that can trigger landmines
 - Pressure
 - Movement
 - Sound
 - Vibration
 - Passage of time
 - Signals
- Types of landmines
 - Anti-tank
 - Anti-personnel
- Different shapes, casings and materials









Motivation

- Currently more than 50-70 million uncleared landmines in at least 70 countries
- It will take about 1,100 years to remove all landmines at the current clearance rate
- Over 26,000 people are killed or maimed every year by landmines
- Over 1,000,000 casualties reported since 1980 [1]
- Half of all casualties in the Iraq and Afghanistan wars were attributed to land mines and improvised explosive devices (IEDs) [2]



[1] U.S. Department of State, "Hidden Killers: The Global Landmine Crisis", US Department of State Publication
[2] C. Wilson, "Improvised explosive devices (IEDs) in Iraq and Afghanistan: effects and countermeasures". CRS Report for Congress, 2007



Landmines Detection Techniques



- Metal Detector
 - Based on disturbances from time-varying magnetic field
 - Most popular but risky and limited to metallic detection
- Acoustic/Seismic methods
 - Based on vibration of materials subjected to sound waves
 - Unaffected by moisture and weather but limited by depth of penetration and interference
- Biological Methods
 - Use of trained dogs, rats, pigs, birds and bees
 - Training required, false alarms common, distraction inevitable
- Mechanical Methods
 - Includes prodding and use of mine clearing machines
 - Efficient but risky and costly
- Electromagnetic Methods
 - Use of microwaves, infrared, X-ray, GPR etc
 - Microwaves produce ambiguous results, infrared-based algorithms are not well developed, Xray results are poor, GPR is promising

Introduction to GPR

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- Electromagnetic pulses are directed at the physical scene-of-interest (SOI)
- Backscatter occurs when the transmitted pulse encounters dielectric constant changes within an SOI
- Backscattered echoes capture the information needed to map the SOI onto the reconstructed imaging space
- Buried targets are expected to have higher dielectric constant values than surrounding material, such as soil and rocks
- GPR data are echoes recorded by receivers for each transmitted pulse







Existing Methods for UWB-GPR Reconstruction



- Delay-and-Sum (DAS)
 - Fast and straightforward implementation
 - Produces images with poor resolution and large side lobes
- Recursive Sidelobe Minimization (RSM)
 - Reconstructs images with reduced clutter and has been applied in other SAR applications
 - Does not incorporate the a priori information that SOI is sparse
- Least Absolute Selection and Shrinkage Operator (LASSO)
 - No straightforward way to choose parameter
- Sparsity Learning Iterative Method (SLIM)
 - Involves matrix inversion that may be too computationally intensive for real applications



Motivation for a Parameter-Free Algorithm

Determining a suitable choice for the parameter of the prior probability density function is not straightforward

Cross-validation is an off-line procedure that is timeconsuming and sacrifices measurement for validation

L-curve is an off-line procedure that is computationally expensive for large-size large scale estimation problems



Overall Methodology



Use the MAP method to incorporate the a priori knowledge that the SOI contains few scatterers

Use "integrate-out" approach to obtain a hyper-parameter-free prior probability density function

Solve the resulting MAP objective functions using the majorize-minimize optimization technique

Jointly estimate noise power



GPR Linear Model



Output of the j^{th} receiver at the i^{th} transmit position $s_{ij}(t) = \sum_{l=1}^{L} x_l \alpha_{ijl} p(t - \tau_{ijl}) + w(t)$

- x_l : Unknown reflection coefficient at l^{th} terrain pixel
- p(t): Transmitted pulse
- α_{ijl} : Round trip pulse attenuation
- τ_{ijl} : Round trip pulse travel time
- w(t): Noise
- $\{y_{ij}\}$: sampled GPR data where

$$\mathbf{y}_{ij} = [s_{ij}(0), s_{ij}(T), \dots, s_{ij}((N-1)T)]^{\mathrm{T}}$$



GPR Linear Model (Cont.)

Model:



y = Ax + w

- Vector y contains sampled GPR data for I transmit positions and J receivers.
- Vector x contains L unknown reflection coefficients
- System matrix A is $(IJN) \times L$
- Problem Statement: Given y and A, estimate x



Maximum A Posteriori Estimation

MAP estimate:

$$\hat{x}_{MAP} = \arg\max_{\boldsymbol{x}} f_{\boldsymbol{X}|\boldsymbol{Y}}(\boldsymbol{x}|\boldsymbol{y})$$

- Assumptions:
 - Noise is WGN with variance σ^2
 - Reflection coefficients are independent and identically distributed

$$\hat{x}_{MAP} = \arg\min_{\mathbf{x}} \frac{1}{2\sigma^2} \|\mathbf{y} - \mathbf{A}\mathbf{x}\|_2^2 + \frac{K}{2} \log\sigma^2 - \log f_{\mathbf{X}}(\mathbf{x})$$

Prior density function is Laplacian distribution $f(a; \lambda) \triangleq \frac{\lambda}{2} \exp(-\lambda |a|)$

Maximum A Posteriori Estimation (Cont.)



Integrate-out approach places a noninformative hyperprior over a parameter [3] to give a hyperparameter-free probability density function

 $f_X(\boldsymbol{x}) = \int_0^\infty f_{X|\Lambda}(\boldsymbol{x}|\lambda) f_{\Lambda}(\lambda) d\lambda$

Conditional probability density function is the Laplacian distribution

$$f_{X|\Lambda}(x|\lambda) = \frac{\lambda}{2} e^{-\lambda|x_l|}$$

Hyperprior is the Jeffreys' prior for the Laplacian distribution

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$$f_{\Lambda}(\lambda) \triangleq [I(\lambda)]^{\frac{1}{2}} \quad (\text{in terms of Fisher's information})$$
$$I(\lambda) = -E_{\lambda} \left[\frac{\partial^2 \log f_{X|\Lambda}(x|\lambda)}{\partial \lambda^2} \right] = \frac{1}{\lambda^2}$$
$$f_{\Lambda}(\lambda) = \frac{1}{\lambda} \quad (\lambda > 0)$$

[3] G.C. Cawley, N.L. Talbot, and M. Girolami, "Sparse multinomial logistic regression via bayesian L1 regularization," Bioinformatics, 10, pp. 209-216, 2007.



Maximum A Posteriori Estimation (Cont.)



$$f_{\boldsymbol{X}}(\boldsymbol{x}) = \int_{0}^{\infty} f_{\boldsymbol{X}|\boldsymbol{\Lambda}}(\boldsymbol{x}|\boldsymbol{\lambda}) f_{\boldsymbol{\Lambda}}(\boldsymbol{\lambda}) d\boldsymbol{\lambda} = \int_{0}^{\infty} \prod_{l=1}^{L} \left(\frac{\lambda}{2} \cdot e^{-\lambda |x_{l}|} \right) \cdot \frac{1}{\lambda} d\lambda$$

- Simplifying and using the gamma integral $f_X(\mathbf{x}) = \left[\frac{1}{2^L} \cdot \frac{\Gamma(L)}{(\sum_{l=1}^L |x_l|)^L}\right]$
- Resulting MAP objective function

$$\phi(x,\sigma^2) = \frac{1}{2\sigma^2}\phi_{LS}(x) + \frac{K}{2}\log(\sigma^2) + L \cdot \gamma(x)$$

where

$$\phi_{LS}(\boldsymbol{x}) \triangleq \left| |\boldsymbol{y} - \boldsymbol{A}\boldsymbol{x}| \right|_{2}^{2}$$
$$\gamma(\boldsymbol{x}) \triangleq \log\left(\sum_{l=1}^{L} |\boldsymbol{x}_{l}|\right)$$



Majorize-Minimize Optimization Technique



- g(x, y) is a majorizing function of f if
 a) g(x, y) ≥ f(x) for all x, y
 b) g(x, x) = f(x) for all x
- MM algorithm $x^{(n+1)} = \arg\min_{x} g(x, x^{(n)})$
- Property of MM algorithm $f(x^{(n+1)}) \le f(x^{(n)})$ for all n



Illustration of MM Concept



PFM Algorithm

Overview:

Suppose Q is a majorizing function for the MAP objective function, ϕ , at the point $x^{(m)}$. Then,

$$\boldsymbol{x}^{(m+1)} \triangleq \arg\min_{\boldsymbol{x}} Q\left(\boldsymbol{x}, \sigma^{2(m)}; \boldsymbol{x}^{(m)}\right)$$

$$\sigma^{2(m+1)} \triangleq \arg\min_{\sigma^2 > 0} \phi\left(\boldsymbol{x}^{(m+1)}, \sigma^2\right)$$
$$= \frac{1}{K} \left\| \left\| \boldsymbol{y} - \boldsymbol{A} \boldsymbol{x}^{(m+1)} \right\|_2^2 \right\|_2$$





What is a suitable choice for Q?

For any $\mathbf{x} \in \mathbb{R}^n$ and function $h(\mathbf{x})$,

$$\log(h(\mathbf{x})) \le \log(h(\mathbf{x}^{(m)})) + \frac{h(\mathbf{x})}{h(\mathbf{x}^{(m)})} - 1$$

$$\gamma(x) = \log\left(\sum_{l=1}^{L} |x_l|\right) \le \log\left(\sum_{l=1}^{L} \left|x_l^{(m)}\right|\right) + \frac{\sum_{l=1}^{L} |x_l|}{\sum_{l=1}^{L} \left|x_l^{(m)}\right|} - 1$$

• Using De Leeuw and Lange's [4] majorization function for |a|

$$q(\mathbf{x}; \mathbf{x}^{(m)}) = log\left(\sum_{l=1}^{L} |x_l^{(m)}|\right) + \frac{\sum_{l=1}^{L} \frac{\gamma'(x_l^{(m)})}{2x_l^m} (x_l^2 - x_l^{(m)}) + \gamma(x_l^{(m)})}{\sum_{l=1}^{L} |x_l^{(m)}|} - 1$$

[4] J. de Leeuw and K. Lange, "Sharp Quadratic Majorization in One Dimension" Computational Statistics and Data Analysis, vol. 53 no.1 pp 2478 February 2004.





• Majorizing function for $\phi_{LS}(\mathbf{x})$

$$\phi_{LS}(\mathbf{x}) = \sum_{k=1}^{K} y_k^2 - 2 \sum_{k=1}^{K} y_k [\mathbf{A}\mathbf{x}]_k + \sum_{k=1}^{K} \left([\mathbf{A}\mathbf{x}^{(m)}]_k \right)^2$$

- De Pierro [5] developed a majorizing function for $\left(\left[Ax^{(m)}\right]_{k}\right)^{2}$ $r_{k}(x;x^{(m)}) = \sum_{l=1}^{L} c_{kl} \left(n_{k}A_{kl}x_{l} - n_{k}A_{kl}x_{l}^{(m)} + \left[Ax^{(m)}\right]_{k}\right)^{2}$
- Therefore,

$$q_{LS}(\boldsymbol{x}; \boldsymbol{x}^{(m)}) = \sum_{k=1}^{K} y_k^2 - 2 \sum_{k=1}^{K} y_k [\boldsymbol{A}\boldsymbol{x}]_k + \sum_{k=1}^{K} r_k(\boldsymbol{x}; \boldsymbol{x}^{(m)})$$

[5] A.R De Pierro, "A modified expectaction Maximization algorithm for penalized likelihood estimation in emission tomography", IEEE transactions, medical imagery, pp 132-137, 1995





Majorizing function for ϕ :

$$Q(\mathbf{x}; \mathbf{x}^{(m)}) = \frac{1}{2\sigma^2} \sum_{k=1}^{K} \left(y_k^2 - 2y_k [\mathbf{A}\mathbf{x}]_k + \sum_{l=1}^{L} c_{kl} \left(n_k A_{kl} x_l - n_k A_{kl} x_l^{(m)} + [\mathbf{A}\mathbf{x}^{(m)}]_k \right)^2 \right) + \frac{k}{2} \log(\sigma^2) + L \cdot \log\left(\sum_{l=1}^{L} |x_l^{(m)}| \right) + L \cdot \frac{\sum_{l=1}^{L} \frac{\gamma' \left(x_l^{(m)} \right)}{2x_l^m} \left(x_l^2 - x_l^{(m)} \right) + \gamma(x_l^{(m)})}{\sum_{l=1}^{L} |x_l^{(m)}|} - L$$

Taking the derivative of the majorizing function Q with respect to x_l and setting it to zero yields the desired update for the l^{th} reflectance coefficient





Updates for reflection coefficients and noise power:

$$x_{l}^{(m+1)} = \frac{D_{l}^{(m)} \left(G_{l}^{(m)} + x_{l}^{(m)}H_{l}\right)}{D_{l}^{(m)} + L \cdot \sigma^{2(m)}}, \qquad l = 1, 2, \dots, L$$

$$H(l) = \sum_{k=1}^{K} n_k A_{kl}^2, \quad G^{(m)}(l) = \sum_{k=1}^{K} A_{kl} \left(y_k - \left[A x^{(m)} \right]_k \right)$$

$$D_l^{(m)} = \left[x_l^{(m)} \right] \sum_{l=1}^L \left| x_l^{(m)} \right|$$

$$\boldsymbol{\sigma}^{2(m+1)} = \frac{1}{K} \left| \left| \boldsymbol{y} - \boldsymbol{A} \boldsymbol{x}^{(m+1)} \right| \right|_{2}^{2}$$



PFM Algorithm in Practice

• $\gamma(x)$ is defined as

$$\gamma(\boldsymbol{x}) \triangleq \log\left(\sum_{l=1}^{L} |x_l| + c\right)$$

- C>O insures the parameter-free MAP objective function has a minimizer
- C is chosen such that

 $\phi(\widehat{x}) < \phi(\mathbf{0})$

- DAS image is used as initial estimate
- Acceleration techniques for computing H(l) and $G^{(m)}(l)$ from previous work [6] are used for a fast memory-efficient implementation.

Simulation Results: ARL SIRE System





- A monocycle UWB pulse (300 -3000 MHz)
- 2 transmit antennas

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- 1 active transmit antenna per shot
- 16 receive antennas



ARL SIRE System Prototype

Simulation Results: Real Data

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Image formed using DAS algorithm



Image formed using LMM algorithm



Image formed using PFM algorithm

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Comparison Using Threshold Detector





SNR=20dB, I= 121, J=16



Conclusions



- Developed Parameter-Free MAP algorithms have been successfully applied to synthetic and real data from the impulse-based ARL SIRE system
- Algorithms produced images that are sparse with suppressed background noise while retaining known scatterers

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