

# SUPER-RESOLUTION DELAY-DOPPLER ESTIMATION FOR SUB-NYQUIST RADAR VIA ATOMIC NORM MINIMIZATION

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#### **CONTRIBUTIONS**

- **Purpose:** Achieve gridless delay-Doppler estimation with sub-Nyquist sampled radar echo signal.
- Key idea:
- ➤ Regard the sub-Nyquist sampling as a linear mapping which maps a low-rank matrix to the sampling vector;
- ➤ Use Atomic Norm Minimization (ANM) to recover the low-rank matrix from the sub-Nyquist samples;
- ☐ **Result:** A two-staged processing is developed:
- > Stage 1: solve a computationally efficient ANM problem;
- > Stage 2: implement the delay-Doppler estimation and pairing.

#### **SUB-NYQUIST RADAR**

#### **Signal Model**

The received radar echo signal is:

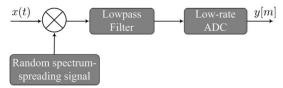
$$x(t) = \sum_{n=0}^{N-1} \sum_{k=1}^{K} \alpha_k g_n (t - nT - \tau_k) e^{j2\pi\nu_k nT}$$

N: the number of transmitted pulses; T: the pulse repetition interval (PRI);  $g_n(t)$ : the pulsed signal transmitted in the n-th PRI;  $\{\tau_k, \nu_k\}$  is the unknown delay-Doppler parameter for the k-th target.

■ **Note:** Here we don't require that the radar transmits the same pulse in each PRI.

### **Sub-Nyquist Sampling**

- The radar echo signal is sampled by an Analog-to-Information Converter (AIC), e.g., random demodulator, Xampling, and so on.
- > Random demodulator is considered here:



 $\triangleright$  During the *n*-th PRI, the sub-Nyquist sample is

$$y_n[m] = \sum_{k=1}^{K} \alpha_k \int_{nT}^{(n+1)T} h(\tau) p(mT_{cs} - \tau) g_n(mT_{cs} - nT - \tau_k - \tau) e^{j2\pi\nu_k nT} d\tau$$

h(t): the lowpass filter; p(t): the random spectrum-spreading signal;

 $T_{cs}$ : the sampling interval of the low-rate ADC.

- lacktriangle Assumption 1: h(t) is an ideal lowpass filter with bandwith  $B_{cs}$ .
- Assumption 2: p(t) is a T-periodic signal  $p(t) = \sum_{l=-L_p}^{L_p} \rho_l e^{j\pi l t/T}$ .
- Assumption 3:  $T_{cs} = 1/B_{cs}, B/Bcs = M$

#### **Frequency-Domain Representation**

The continuous-time Fourier transform of  $y_n(t)$  is

$$Y_n(f) = \sum_{l=-L_0}^{L_0} \sum_{k=1}^K \rho_l \alpha_k e^{j2\pi\nu_k nT} G_n(f - l/T) e^{-j2\pi(f - l/T)\tau_k}$$

The frequency-domain sampling vector of  $Y_n(f)$  can be represented as

$$\tilde{\mathbf{y}}_n = \mathbf{PG}_n \mathbf{W}(\tau) \mathbf{B} \mathbf{V}^T(\nu) \mathbf{e}_n$$

$$\begin{bmatrix} \rho_{1-L_1} & \rho_{2-L_1} & \dots & \rho_{1+L_2} \\ \rho_{2-L_1} & \rho_{2-L_1} & \dots & \rho_{2+L_2} \\ \vdots & \vdots & \ddots & \vdots \\ \rho_{M-L_1} & \rho_{M+1-L_1} & \dots & \rho_{M+L_2} \end{bmatrix} \begin{bmatrix} G_n(f_{(L_1)}) & 0 & \dots & 0 \\ 0 & G_n(f_{(L_1+1)}) & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & G_n(f_{(L_2)}) \end{bmatrix} \\ \begin{bmatrix} \alpha_1 e^{-j2\pi f_{(L_1)}\tau_1} & 0 & \dots & 0 \\ 0 & \alpha_2 e^{-j2\pi f_{(L_1)}\tau_2} & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & \alpha_K e^{-j2\pi f_{(L_1)}\tau_K} \end{bmatrix} \\ \mathbf{W}(\tau) = [\mathbf{w}(\tau_1/T) \dots \mathbf{w}(\tau_M/T)] \text{ with } \mathbf{w}(\theta) = [1 e^{j2\pi\theta} \dots e^{j2\pi(L-1)\theta}]^T$$

$$\mathbf{W}(\tau) = [\mathbf{w}(\tau_1/T), \cdots, \mathbf{w}(\tau_K/T)] \quad \text{with} \quad \mathbf{w}(\theta) = [1, e^{j2\pi\theta}, \cdots, e^{j2\pi(L-1)\theta}]^T$$

$$\mathbf{V}(\nu) = [\mathbf{v}(\nu_1 T), \cdots, \mathbf{v}(\nu_K T)] \quad \text{with} \quad \mathbf{v}(\theta) = [1, e^{j2\pi\theta}, \cdots, e^{j2\pi(N-1)\theta}]^T$$

Let  $\theta_k=\tau_k/T-1/2, \vartheta_k=-\nu_kT\in (-1/2,1/2],$  the m-th element of  $\tilde{\mathbf{y}}_n$  can be represented as

$$\tilde{y}_n[m] = \sum_{k=1}^{K} \beta_k \mathbf{m}_{m,n}^H \mathbf{w}(\theta_k) \mathbf{v}^H(\vartheta_k) \mathbf{e}_n$$

$$= \text{Tr} \left( \mathbf{e}_n \mathbf{m}_{m,n}^H \sum_{k=1}^K \beta_k \mathbf{w}(\theta_k) \mathbf{v}^H(\vartheta_k) \right)$$

$$= \langle \mathbf{X}, \mathbf{m}_{m,n} \mathbf{e}_n^H \rangle,$$

where  $\mathbf{X} = \sum_{k=1}^{K} \beta_k \mathbf{w}(\theta_k) \mathbf{v}^H(\vartheta_k)$  is a low-rank matrix.

Let  $\tilde{\mathbf{y}} = [\tilde{\mathbf{y}}_0^T, \tilde{\mathbf{y}}_1^T, \cdots, \tilde{\mathbf{y}}_{N-1}^T]^T$ , the process of sub-Nyquist sampling can be

expressed as

 $\mathbf{\tilde{y}} = \mathcal{B}(\mathbf{X})$ 

 $\mathcal{B}(\cdot)$ : the linear operator mapping the matrix  $\boldsymbol{X}$  to  $\tilde{\boldsymbol{y}}$ .

Delay-Doppler Estimation Low-Rank Matrix Recovery

## **Method & Algorithm**

#### Low-Rank Matrix Recovery via ANM

The set of atoms is defined as

$$\mathcal{A} \triangleq \{ \mathbf{A}(\boldsymbol{\theta}, \phi) = e^{j\phi} \mathbf{w}(\theta) \mathbf{v}^H(\vartheta) : \theta, \vartheta \in \mathbb{T}, \phi \in \mathbb{S} \},$$

where  $\mathbb{T} \triangleq (-1/2, 1/2]$  and  $\mathbb{S} \triangleq (0, 2\pi]$ .

The atomic  $l_0$  norm is defined as

$$\|\mathbf{X}\|_{\mathcal{A},0} = \inf \left\{ \mathcal{K} : \mathbf{X} = \sum_{k=1}^{\mathcal{K}} c_k \mathbf{A}(\boldsymbol{\theta}_k, \phi_k), c_k > 0 \right\}$$

 $\|\mathbf{X}\|_{\mathcal{A},0}$  equals the optimal value of the following rank minimization problem:  $\min_{\mathbf{u}_1\in\mathbb{C}^L,\mathbf{u}_2\in\mathbb{C}^N} \quad \mathrm{rank}(\mathbf{M}),$ 

s.t. 
$$\mathbf{M} = \begin{bmatrix} \mathcal{T}(\mathbf{u}_1) & \mathbf{X} \\ \mathbf{X}^H & \mathcal{T}(\mathbf{u}_2) \end{bmatrix} \geq 0,$$

where  $\mathcal{T}(\mathbf{u})$  denotes the Toeplitz matrix with  $\mathbf{u}^T$  as its first row.

With relaxation, the low-rank matrix recovery can be cast as:

$$\begin{aligned} \min_{\mathbf{u}_1 \in \mathbb{C}^L, \mathbf{u}_2 \in \mathbb{C}^N, \mathbf{X}} & \operatorname{trace}(\mathcal{T}(\mathbf{u}_1)) + \operatorname{trace}(\mathcal{T}(\mathbf{u}_2)), \\ \text{s.t.} & \mathbf{M} = \begin{bmatrix} \mathcal{T}(\mathbf{u}_1) & \mathbf{X} \\ \mathbf{X}^H & \mathcal{T}(\mathbf{u}_2) \end{bmatrix} \geq 0, \\ & \tilde{\mathbf{y}} = \mathcal{B}(\mathbf{X}). \end{aligned}$$

#### **Delay-Doppler Estimation and Pairing Algorithm**

- > Input:  $u_1^*, u_2^*, X^*, K^* = rank(M)$ .
- > Step1: Perform Vandermonde decomposition  $\mathcal{T}(\mathbf{u}_1^*) = \mathbf{W}_1 \mathbf{\Sigma}_1 \mathbf{W}_1^H$  and  $\mathcal{T}(\mathbf{u}_2^*) = \mathbf{V}_2 \mathbf{\Sigma}_2 \mathbf{V}_2^H$ , acquire two sets of parameters  $\{\theta_1^*, \theta_2^*, \cdots, \theta_{K_1}^*\}$  and  $\{\vartheta_1^*, \vartheta_2^*, \cdots, \vartheta_{K_2}^*\}$ .
- > Step 2: Compute a  $K_1 \times K_2$  dimensional matrix  $\mathbf{O} = \mathbf{\Sigma}_1^{-1/2} \mathbf{W}_1^{\dagger} \mathbf{X} \mathbf{V}_2^{\dagger} \mathbf{\Sigma}_2^{-1/2}$ .
- > Step 3: Select the positions of the  $K^*$  largest elements in the matrix **0** to determine the  $K^*$  delay-Doppler pairs.
- **Note:** There exists a matrix  $\mathbf{O} \in \mathbb{C}^{K_1 \times K_2}$  satisfying  $\mathbf{X} = \mathbf{W}_1 \mathbf{\Sigma}_1^{1/2} \mathbf{O} \mathbf{\Sigma}_2^{1/2} \mathbf{V}_2^H$ . The matrix  $\mathbf{O}$  determines the pairing relationship between the two sets of parameters.

#### **Simulation Results**

A LFM signal with bandwidth 20MHz and pulse-width  $2\mu s$  is transmitted. The receiver samples the radar echoes at 1/4 Nyquist rate, i.e.,  $B_{cs} = 5$ MHz. Other parameters are PRI= $4\mu s$ , N = 50. The performance of ANM-based method is compared with Doppler focusing (in *O. Bar-Ilan and Y. C. Eldar "Sub-Nyquist radar via Doppler focusing," IEEE T-SP, 2014*) and Cramér-Rao bound (CRB).

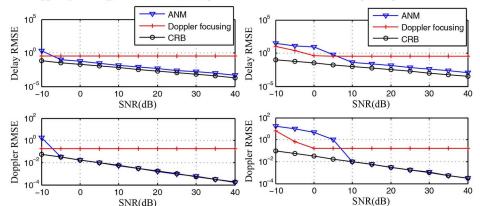


Fig 1. Two delay-Doppler pairs, *i.e.*, K = 2.

Fig 2. Four delay-Doppler pairs, *i.e.*, K = 4.

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