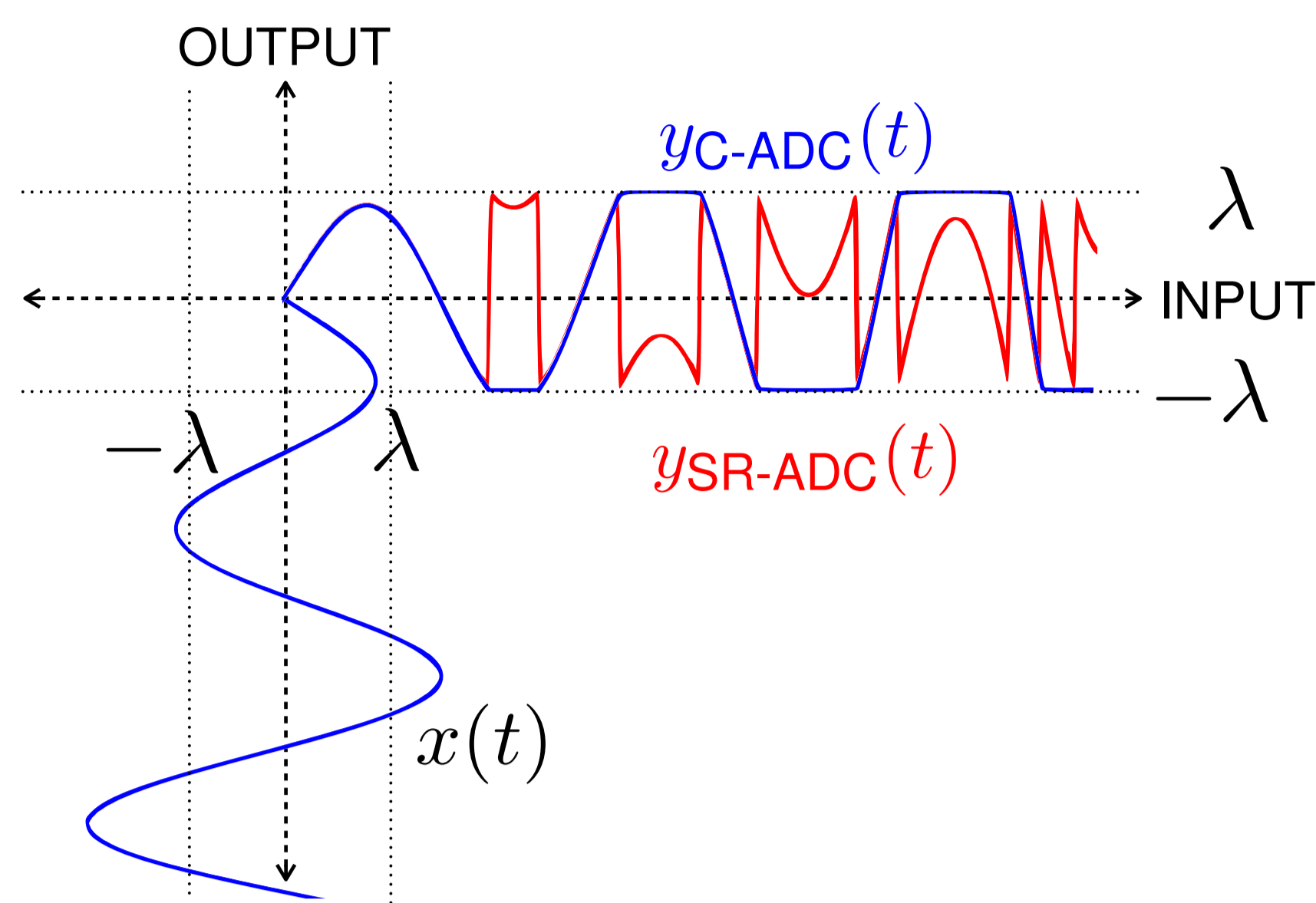


1. Overview

Self-reset analog-to-digital converter (SR-ADC):

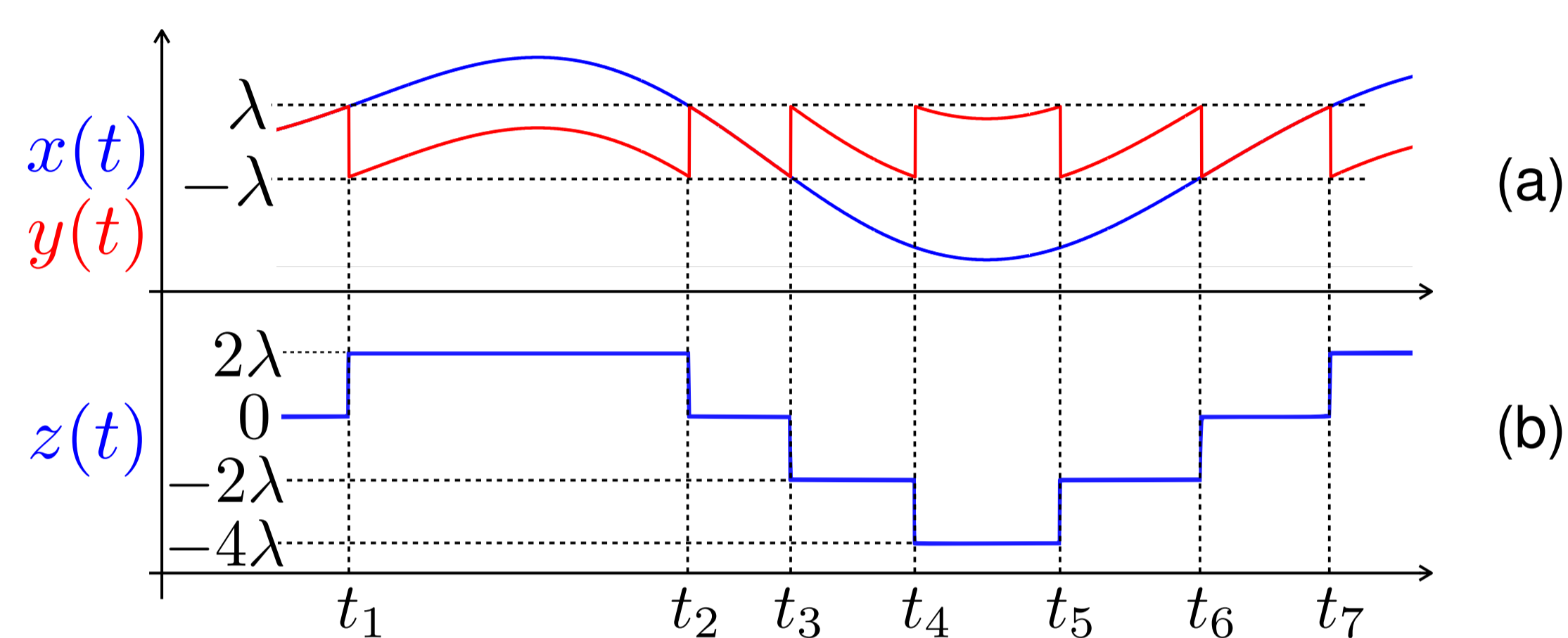


An illustration of the transfer characteristics of a clipping-ADC and a self-reset-ADC with the thresholds at $\pm\lambda$.

Output:

$$y(t) = \mathcal{M}_\lambda\{x(t)\} := \text{mod}(x(t) + \lambda, 2\lambda) - \lambda, \\ = x(t) - z(t),$$

where $z(t) = \sum_k \alpha_k \mathbf{1}_{[t_k, t_{k+1}]}(t)$ is a piecewise-constant signal.



Modulo operation: (a) Input signal $x(t)$ and its modulo version $y(t)$; and (b) piecewise-constant signal $z(t)$.

Discrete version:

$$y(nT) = \mathcal{M}_\lambda\{x(nT)\} = x(nT) - z(nT),$$

where $z(nT) = \sum_k \alpha_k \mathbf{1}_{[n_k, n_{k+1}]}(nT)$.

2. Problem Formulation

- Given $y(t) = \mathcal{M}_\lambda\{x(t)\}$, reconstruct $x(t)$.
- Given $y(nT) = \mathcal{M}_\lambda\{x(nT)\}$, reconstruct $x(nT)$.

3. Continuous-Domain Analysis

Definition 1 (Vanishing moments)

A wavelet ψ has $p+1$ vanishing moments if $\int t^k \psi(t) dt = 0$, for $k \in \llbracket 0, p \rrbracket$.

Key idea: Use wavelets to annihilate the smooth parts of $y(t)$.

Lemma 1 Let $x(t)$ be a polynomial of degree p , $\psi(t)$ be a wavelet with $p+1$ vanishing moments, and $y(t) = \mathcal{M}_\lambda\{x(t)\}$. Then, $y_\psi(t) := (y * \psi)(t) = \sum_k (\alpha_k - \alpha_{k-1}) \xi(t - t_k)$, where $\xi(t) = -\int_{-\infty}^t \psi(\tau) d\tau$.

- Estimation of α_k s: by matched filtering.
- Sufficient condition for exact reconstruction: $(t_k - t_{k-1}) > (2p-1) \implies \frac{2\lambda}{L} < (2p-1)$.

4. Discrete-Domain Analysis

$$y(nT) = \mathcal{M}_\lambda\{x(nT)\} = x(nT) - z(nT)$$

Annihilation of a sampled polynomial of degree p : $\sum_{n \in \mathbb{Z}} n^k g[n] = 0$, for $k \in \llbracket 0, p \rrbracket$.

Compactly supported $g[n]$: Daubechies filter of order p with support $\llbracket 0, 2p-1 \rrbracket$.

Lemma 2 Let $x[n]$ be the samples of a polynomial of degree p , and let $g[n]$ be a discrete wavelet filter with $p+1$ vanishing moments. Then, $y_g[n] := (y * g)[n] = \sum_k \beta_k m[n - n_k]$ with $\beta_k = (\alpha_k - \alpha_{k-1})$

and $m[n] := -\sum_{k=-\infty}^n g[k]$.

Computing $\{n_k\}$:

$$\text{LASSO: } \arg \min_{\mathbf{h}} \|\mathbf{y}_g - \mathbf{A}\mathbf{h}\|_2^2 + \gamma \|\mathbf{h}\|_1$$

\mathbf{A} – convolutional dictionary of time-shifted versions of $m[n]$

\mathbf{h} – sparse vector with $\mathbf{h}[n_k] = \beta_k$ and $\text{supp}\{\mathbf{h}\} = \text{cardinality of } \{n_k\}$

Selection of sampling interval (T):

- Let $T_f := \min_k (n_k - n_{k-1})$ be the minimum sampling interval.
- For no overlap: $T_f > \text{supp}\{m[n]\}$.
- Support of Daubechies wavelet filter of order p is $2p$.
- A sufficient condition on T : $T < \frac{T_f}{2p}$.

References

- [1] A. Bhandari, F. Krahmer, and R. Raskar, "On unlimited sampling," in *Proceedings of International Conference on Sampling Theory and Applications (SampTA)*, July 2017, pp. 31–35.
- [2] K. Sasagawa, T. Yamaguchi, M. Haruta, Y. Sunaga, H. Takehara, T. Noda, T. Tokuda, and J. Ohta, "An implantable CMOS image sensor with self-reset pixels for functional brain imaging," *IEEE Transactions on Electron Devices*, vol. 63, no. 1, pp. 215–222, Jan 2016.
- [3] I. Daubechies, "Ten Lectures on Wavelets," *SIAM*, 1992.
- [4] S. Mallat, *A Wavelet Tour of Signal Processing*, Academic Press, 2008.
- [5] R. Tibshirani, "Regression shrinkage and selection via the lasso," *Journal of the Royal Statistical Society, Series B*, vol. 58, pp. 267–288, 1994.

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5. Main Result: WAVE-BUS

WAVE-BUS: WAVElet-Based Unlimited Sampling

Lemma 3 Let $y(t) = \mathcal{M}_\lambda\{x(t)\}$, where $x(t)$ is a Lipschitz-continuous signal that satisfies $|x(a) - x(b)| \leq L|b - a|$. Then, we have $\tilde{T}_f := \frac{2\lambda}{L} \leq T_f$.

Proof: From Lipschitz continuity, $|x(t + \tilde{T}_f) - x(t)| \leq L\tilde{T}_f < 2\lambda, \forall t$.

Hence, at any t_k , no folding happens in the interval $(t_k, t_k + \tilde{T}_f)$. Thus, $\tilde{T}_f \leq T_f$.

- Sufficient condition for exact reconstruction: Sampling interval $T \leq \frac{T_f}{2p} \leq \frac{2\lambda}{L(2p)}$.

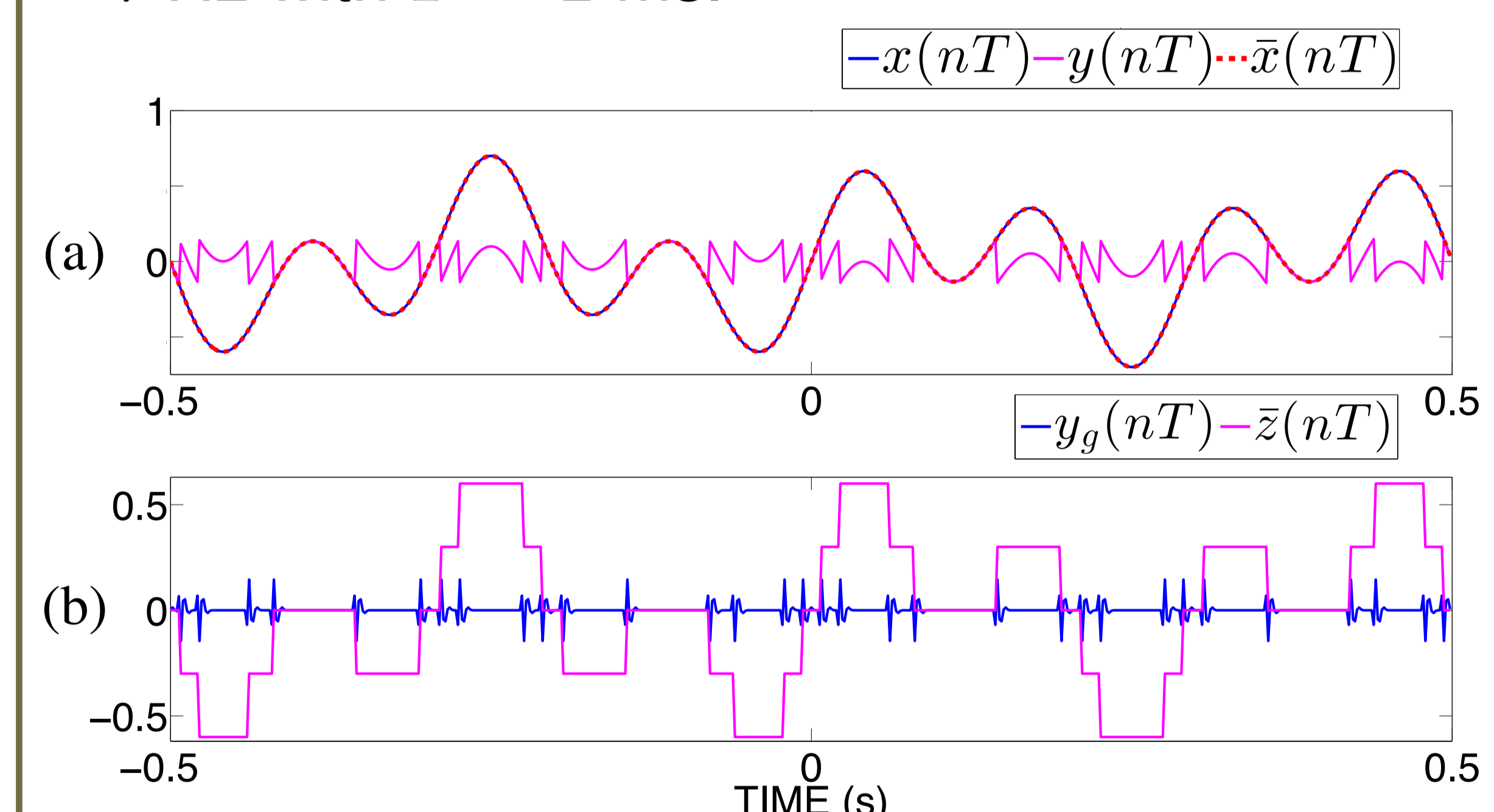
6. Reconstruction Algorithm

Algorithm 1 : WAVE-BUS.

- Input: $y(nT) = \mathcal{M}_\lambda\{x(nT)\}$, L , λ , p , T
- Output: $\bar{x}(nT)$
- Method:
 - Wavelet filtering: $(y_g)[n] = (y * g)[n]$
 - LASSO (compute n_k s and α_k s): $\arg \min_{\mathbf{h}} \|\mathbf{y}_g - \mathbf{A}\mathbf{h}\|_2^2 + \gamma \|\mathbf{h}\|_1$
 - Compute $\bar{z}[n] : \sum_k \alpha_k \mathbf{1}_{[n_k, n_{k+1}]}$
 - Reconstruct $x[n] : \bar{x}[n] = y[n] + \bar{z}[n]$

7. Simulation Results

- Sum of sinusoids of frequency 4 Hz and 7 Hz with $T = 2$ ms.



WAVE-BUS with $\lambda = 0.15$: (a) Signal $x(nT)$, its modulo samples $y(nT)$, and the reconstruction $\bar{x}(nT)$; (b) wavelet filter output $y_g(nT)$, and the estimate $\bar{z}(nT)$. The reconstruction error was computed to be -330 dB.

- Signal-to-reconstruction-noise ratio (SRNR) vs input SNR for WAVE-BUS and repeated finite-difference (RFD)¹ method.

