

Feature LMS Algorithms

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Presentation Outline

- 1 Introduction
- 2 Feature LMS Algorithms
- 3 Results
- 4 Conclusions

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1 Introduction

2 Feature LMS Algorithms

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Motivations

- When we have some *a priori knowledge* about the unknown system \mathbf{w}_o
 - We can exploit it for accelerating the convergence rate
- When we want to obtain an estimate of the unknown system \mathbf{w}_o such that a *determined characteristic* for the estimate is desirable
 - Lowpass, highpass, linear phase

Proposal

- Feature LMS (F-LMS) algorithms \Rightarrow impose some structure on the adaptive filter's coefficients \Rightarrow exploit **hidden** sparsity in system, such as sparsity in linear combination of coefficients
- In this paper, we present the F-LMS algorithm for:
 - Unknown systems with **lowpass** narrowband spectrum
 - Unknown systems with **highpass** narrowband spectrum

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F-LMS Algorithm: Problem and Solution

- Problem:

$$\xi_{\text{F-LMS}}(k) = \underbrace{\frac{1}{2}|e(k)|^2}_{\text{standard LMS term}} + \underbrace{\alpha \mathcal{P}(\mathbf{F}(k)\mathbf{w}(k))}_{\text{feature-inducing term}},$$

where $\mathcal{P}(\cdot)$ is the sparsity promoting penalty function and $\mathbf{F}(k)$ is the feature matrix that takes the unknown system to a **sparse vector**.

- For example, choose function \mathcal{P} to be the l_1 norm and the feature matrix $\mathbf{F}(k)$ to be time-invariant \mathbf{F}

$$\xi_{\text{F-LMS}}(k) = \frac{1}{2}|e(k)|^2 + \alpha \|\mathbf{F}\mathbf{w}(k)\|_1$$

- Solution:

$$\mathbf{w}(k+1) = \mathbf{w}(k) + \mu e(k)\mathbf{x}(k) - \mu\alpha\mathbf{p}(k),$$

where $\mathbf{p}(k) \in \mathbb{R}^{N+1}$ is the gradient of function $\|\mathbf{F}\mathbf{w}(k)\|_1$.

Example I: F-LMS Algorithm for Lowpass Systems

- Unknown system has **lowpass** narrowband spectrum \Rightarrow its impulse response is **smooth** \Rightarrow the difference between adjacent coefficients is small
- Choose $\mathbf{F} \in \mathbb{R}^{N \times (N+1)}$ as

$$\mathbf{F} = \begin{bmatrix} 1 & -1 & 0 & \cdots & 0 \\ 0 & 1 & -1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & -1 \end{bmatrix} \Rightarrow \mathbf{F}\mathbf{w}(k) \text{ is a sparse vector}$$

- Therefore, $\mathbf{p}(k) = [p_0(k) \cdots p_N(k)]^T$ is given by

$$\begin{cases} p_i(k) = \text{sgn}(w_0(k) - w_1(k)) & \text{if } i = 0, \\ p_i(k) = -\text{sgn}(w_{i-1}(k) - w_i(k)) + \text{sgn}(w_i(k) - w_{i+1}(k)) & \text{if } i = 1, \dots, N-1, \\ p_i(k) = -\text{sgn}(w_{N-1}(k) - w_N(k)) & \text{if } i = N. \end{cases}$$

Example II: F-LMS Algorithm for Highpass Systems

- Unknown system has **highpass** narrowband spectrum \Rightarrow adjacent coefficients have similar absolute values, but with **opposite signs**
- Choose $\mathbf{F} \in \mathbb{R}^{N \times (N+1)}$ as

$$\mathbf{F} = \begin{bmatrix} 1 & 1 & 0 & \cdots & 0 \\ 0 & 1 & 1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 1 \end{bmatrix} \Rightarrow \mathbf{F}\mathbf{w}(k) \text{ is a sparse vector}$$

- Therefore, $\mathbf{p}(k) = [p_0(k) \cdots p_N(k)]^T$ is given by

$$\begin{cases} p_i(k) = \text{sgn}(w_0(k) + w_1(k)) & \text{if } i = 0, \\ p_i(k) = \text{sgn}(w_{i-1}(k) + w_i(k)) + \text{sgn}(w_i(k) + w_{i+1}(k)) & \text{if } i = 1, \dots, N-1, \\ p_i(k) = \text{sgn}(w_{N-1}(k) + w_N(k)) & \text{if } i = N. \end{cases}$$

More Examples for Feature Matrix

- When unknown system is the result of **upsampling** a lowpass system by a factor of L (e.g., $L = 2$)

$$\mathbf{F} = \begin{bmatrix} 1 & 0 & -1 & 0 & \cdots & 0 \\ 0 & 1 & 0 & -1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 0 & -1 \end{bmatrix}.$$

- When unknown system is the result of **interpolating** a highpass system by a factor L (e.g., $L = 2$)

$$\mathbf{F} = \begin{bmatrix} 1 & 0 & 1 & 0 & \cdots & 0 \\ 0 & 1 & 0 & 1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 0 & 1 \end{bmatrix}.$$

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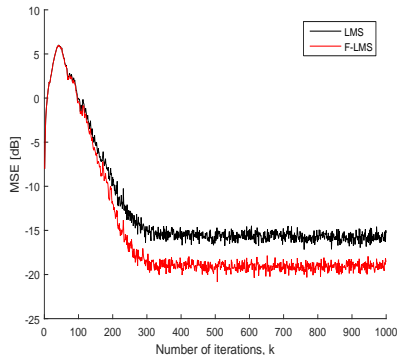
4 Conclusions

Scenario: System Identification

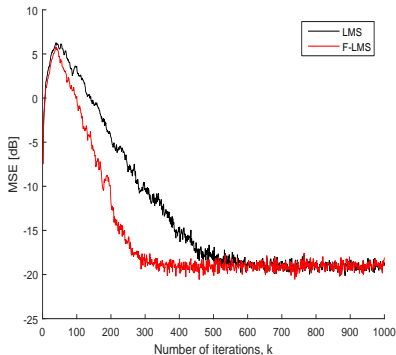
- Algorithms tested: LMS and F-LMS algorithms
- Input signal: $\mathbf{x} \sim \mathcal{N}(0, 1)$
- Filter order: $N = 39$, i.e., 40 coefficients
- $\mathbf{w}(0) = [0, \dots, 0]^T$
- $\alpha = 0.05$
- SNR: 20 dB
- Unknown lowpass system: $\mathbf{w}_{o,l} = [0.4, 0.4, \dots, 0.4]^T$
- Unknown highpass system: $\mathbf{w}_{o,h} = [0.4, -0.4, 0.4 \dots, -0.4]^T$

F-LMS Algorithm Identifying Unknown System with Lowpass Spectrum

- Unknown lowpass system: $\mathbf{w}_{o,l} = [0.4, 0.4, \dots, 0.4]^T$



(a) $\mu_{LMS} = \mu_{F-LMS} = 0.03$

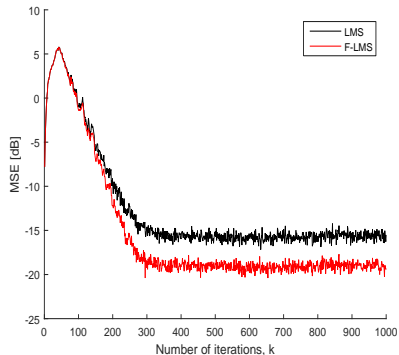


(b) $\mu_{LMS} = 0.01, \mu_{F-LMS} = 0.03$

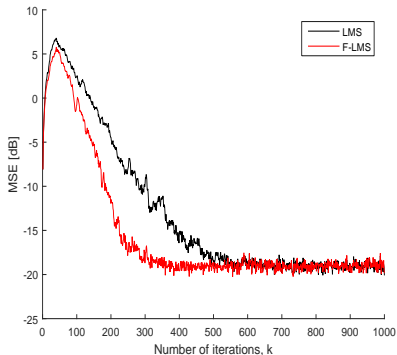
Figure: Learning (MSE) curves

F-LMS Algorithm Identifying Unknown System with Highpass Spectrum

- Unknown highpass system: $\mathbf{w}_{o,h} = [0.4, -0.4, 0.4 \dots, -0.4]^T$



(a) $\mu_{LMS} = \mu_{F-LMS} = 0.03$

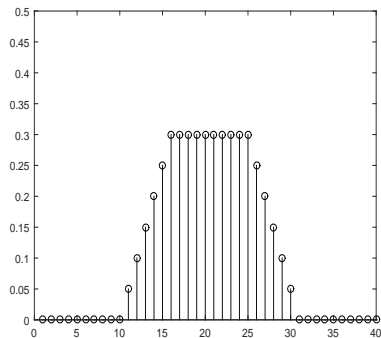


(b) $\mu_{LMS} = 0.01, \mu_{F-LMS} = 0.03$

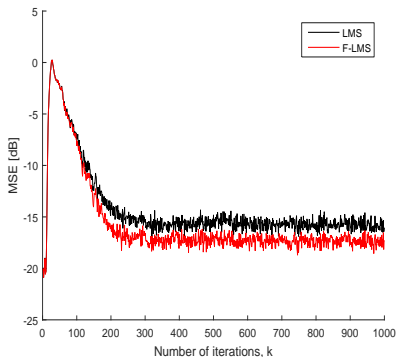
Figure: Learning (MSE) curves

F-LMS Algorithm Identifying Unknown Block Sparse System

- Block sparse system with lowpass narrowband spectrum



(a) Unknown system

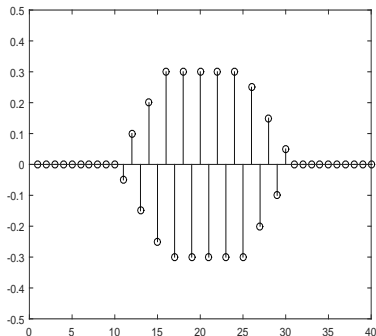


(b) $\mu_{LMS} = \mu_{F-LMS} = 0.03$

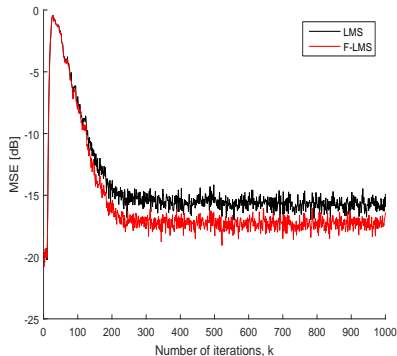
Figure: Learning (MSE) curves

F-LMS Algorithm Identifying Unknown Block Sparse System

- Block sparse system with highpass narrowband spectrum



(a) Unknown system



(b) $\mu_{LMS} = \mu_{F-LMS} = 0.03$

Figure: Learning (MSE) curves

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Conclusions

- In this presentation:
 - We have proposed a family of algorithms called feature LMS algorithm
 - We have presented some examples of the F-LMS algorithms for exploiting the lowpass and highpass characteristics of unknown systems
 - Some other characteristics can be exploited (e.g., linear phase)
 - The F-LMS algorithms have some advantages such as higher convergence rate or lower steady-state MSE
 - The computational complexity of the proposed F-LMS algorithms are close to that of the LMS algorithm

Thank You!