

# Feature LMS Algorithms

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# Presentation Outline

- 1 Introduction
- 2 Feature LMS Algorithms
- 3 Results
- 4 Conclusions

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1 Introduction

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# Motivations

- When we have some *a priori knowledge* about the unknown system  $\mathbf{w}_o$ 
  - We can exploit it for accelerating the convergence rate
- When we want to obtain an estimate of the unknown system  $\mathbf{w}_o$  such that a *determined characteristic* for the estimate is desirable
  - Lowpass, highpass, linear phase

# Proposal

- Feature LMS (F-LMS) algorithms  $\Rightarrow$  impose some structure on the adaptive filter's coefficients  $\Rightarrow$  exploit **hidden** sparsity in system, such as sparsity in linear combination of coefficients
- In this paper, we present the F-LMS algorithm for:
  - Unknown systems with **lowpass** narrowband spectrum
  - Unknown systems with **highpass** narrowband spectrum

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## F-LMS Algorithm: Problem and Solution

- Problem:

$$\xi_{\text{F-LMS}}(k) = \underbrace{\frac{1}{2}|e(k)|^2}_{\text{standard LMS term}} + \underbrace{\alpha \mathcal{P}(\mathbf{F}(k)\mathbf{w}(k))}_{\text{feature-inducing term}},$$

where  $\mathcal{P}(\cdot)$  is the sparsity promoting penalty function and  $\mathbf{F}(k)$  is the feature matrix that takes the unknown system to a **sparse vector**.

- For example, choose function  $\mathcal{P}$  to be the  $l_1$  norm and the feature matrix  $\mathbf{F}(k)$  to be time-invariant  $\mathbf{F}$

$$\xi_{\text{F-LMS}}(k) = \frac{1}{2}|e(k)|^2 + \alpha \|\mathbf{F}\mathbf{w}(k)\|_1$$

- Solution:

$$\mathbf{w}(k+1) = \mathbf{w}(k) + \mu e(k)\mathbf{x}(k) - \mu\alpha\mathbf{p}(k),$$

where  $\mathbf{p}(k) \in \mathbb{R}^{N+1}$  is the gradient of function  $\|\mathbf{F}\mathbf{w}(k)\|_1$ .

## Example I: F-LMS Algorithm for Lowpass Systems

- Unknown system has **lowpass** narrowband spectrum  $\Rightarrow$  its impulse response is **smooth**  $\Rightarrow$  the difference between adjacent coefficients is small
- Choose  $\mathbf{F} \in \mathbb{R}^{N \times (N+1)}$  as

$$\mathbf{F} = \begin{bmatrix} 1 & -1 & 0 & \cdots & 0 \\ 0 & 1 & -1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & -1 \end{bmatrix} \Rightarrow \mathbf{F}\mathbf{w}(k) \text{ is a sparse vector}$$

- Therefore,  $\mathbf{p}(k) = [p_0(k) \cdots p_N(k)]^T$  is given by

$$\begin{cases} p_i(k) = \text{sgn}(w_0(k) - w_1(k)) & \text{if } i = 0, \\ p_i(k) = -\text{sgn}(w_{i-1}(k) - w_i(k)) + \text{sgn}(w_i(k) - w_{i+1}(k)) & \text{if } i = 1, \dots, N-1, \\ p_i(k) = -\text{sgn}(w_{N-1}(k) - w_N(k)) & \text{if } i = N. \end{cases}$$



## Example II: F-LMS Algorithm for Highpass Systems

- Unknown system has **highpass** narrowband spectrum  $\Rightarrow$  adjacent coefficients have similar absolute values, but with **opposite signs**
- Choose  $\mathbf{F} \in \mathbb{R}^{N \times (N+1)}$  as

$$\mathbf{F} = \begin{bmatrix} 1 & 1 & 0 & \cdots & 0 \\ 0 & 1 & 1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 1 \end{bmatrix} \Rightarrow \mathbf{F}\mathbf{w}(k) \text{ is a sparse vector}$$

- Therefore,  $\mathbf{p}(k) = [p_0(k) \cdots p_N(k)]^T$  is given by

$$\begin{cases} p_i(k) = \text{sgn}(w_0(k) + w_1(k)) & \text{if } i = 0, \\ p_i(k) = \text{sgn}(w_{i-1}(k) + w_i(k)) + \text{sgn}(w_i(k) + w_{i+1}(k)) & \text{if } i = 1, \dots, N-1, \\ p_i(k) = \text{sgn}(w_{N-1}(k) + w_N(k)) & \text{if } i = N. \end{cases}$$

## More Examples for Feature Matrix

- When unknown system is the result of **upsampling** a lowpass system by a factor of  $L$  (e.g.,  $L = 2$ )

$$\mathbf{F} = \begin{bmatrix} 1 & 0 & -1 & 0 & \cdots & 0 \\ 0 & 1 & 0 & -1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 0 & -1 \end{bmatrix}.$$

- When unknown system is the result of **interpolating** a highpass system by a factor  $L$  (e.g.,  $L = 2$ )

$$\mathbf{F} = \begin{bmatrix} 1 & 0 & 1 & 0 & \cdots & 0 \\ 0 & 1 & 0 & 1 & \cdots & 0 \\ \vdots & & \ddots & \ddots & \ddots & \\ 0 & 0 & \cdots & 1 & 0 & 1 \end{bmatrix}.$$

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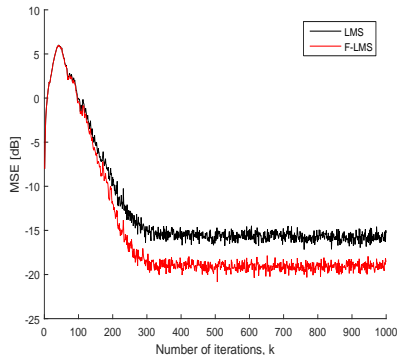
4 Conclusions

## Scenario: System Identification

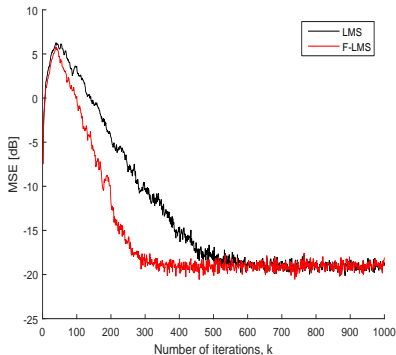
- Algorithms tested: LMS and F-LMS algorithms
- Input signal:  $\mathbf{x} \sim \mathcal{N}(0, 1)$
- Filter order:  $N = 39$ , i.e., 40 coefficients
- $\mathbf{w}(0) = [0, \dots, 0]^T$
- $\alpha = 0.05$
- SNR: 20 dB
- Unknown lowpass system:  $\mathbf{w}_{o,l} = [0.4, 0.4, \dots, 0.4]^T$
- Unknown highpass system:  $\mathbf{w}_{o,h} = [0.4, -0.4, 0.4 \dots, -0.4]^T$

# F-LMS Algorithm Identifying Unknown System with Lowpass Spectrum

- Unknown lowpass system:  $\mathbf{w}_{o,l} = [0.4, 0.4, \dots, 0.4]^T$



(a)  $\mu_{LMS} = \mu_{F-LMS} = 0.03$

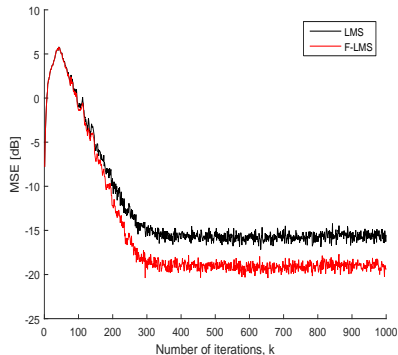


(b)  $\mu_{LMS} = 0.01, \mu_{F-LMS} = 0.03$

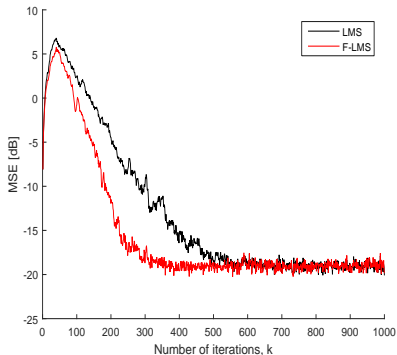
Figure: Learning (MSE) curves

# F-LMS Algorithm Identifying Unknown System with Highpass Spectrum

- Unknown highpass system:  $\mathbf{w}_{o,h} = [0.4, -0.4, 0.4 \dots, -0.4]^T$



(a)  $\mu_{LMS} = \mu_{F-LMS} = 0.03$

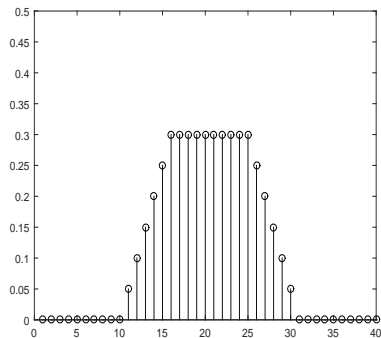


(b)  $\mu_{LMS} = 0.01, \mu_{F-LMS} = 0.03$

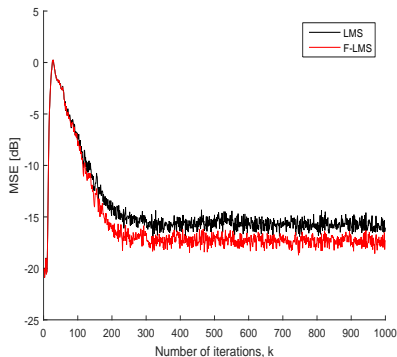
Figure: Learning (MSE) curves

# F-LMS Algorithm Identifying Unknown Block Sparse System

- Block sparse system with lowpass narrowband spectrum



(a) Unknown system

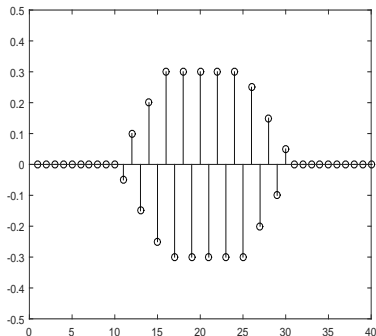


(b)  $\mu_{LMS} = \mu_{F-LMS} = 0.03$

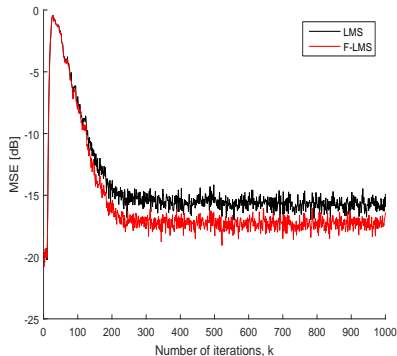
Figure: Learning (MSE) curves

# F-LMS Algorithm Identifying Unknown Block Sparse System

- Block sparse system with highpass narrowband spectrum



(a) Unknown system



(b)  $\mu_{LMS} = \mu_{F-LMS} = 0.03$

Figure: Learning (MSE) curves



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# Conclusions

- In this presentation:
  - We have proposed a family of algorithms called feature LMS algorithm
    - We have presented some examples of the F-LMS algorithms for exploiting the lowpass and highpass characteristics of unknown systems
    - Some other characteristics can be exploited (e.g., linear phase)
  - The F-LMS algorithms have some advantages such as higher convergence rate or lower steady-state MSE
  - The computational complexity of the proposed F-LMS algorithms are close to that of the LMS algorithm

Thank You!