Identification of Bilinear Forms with the Kalman Filter

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Outline



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- 4 Simplified Kalman Filter for Bilinear Forms (SKF BF)
- 5 Practical Considerations
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Conclusions

Kalman filter - used before in system identification
 → reference (desired) signal:

 $d(t) = \mathbf{h}^{\mathsf{T}}(t)\mathbf{x}(t) + \mathbf{v}(t)$

h - unknown system of length L $\mathbf{x}(t) = [x(t)x(t-1) \dots x(n-L+1)]^T$ - input signal v(t) - system noise Kalman filter - used before in system identification
 → reference (desired) signal:

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Our approach - identification of bilinear forms
 → reference (desired) signal:

 $d(t) = \mathbf{h}^{T}(t)\mathbf{X}(t)\mathbf{g}(t) + \mathbf{v}(t)$

h, **g**: unknown systems of lengths *L*, *M* $\mathbf{X}(t) = [\mathbf{x}_1(t) \quad \mathbf{x}_2(t) \quad \dots \quad \mathbf{x}_M(t)]$ - input signal matrix $\mathbf{x}_m(t) = [x_m(t) \quad x_m(t-1) \quad \dots \quad x_m(t-L+1)]^T \quad m = 1, 2, \dots, M$

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- Target: a Kalman algorithm for the identification of bilinear forms
- Timely topic
- Numerous applications:
 - \rightarrow nonlinear acoustic echo cancellation
 - \rightarrow identification of Hammerstein systems
 - \rightarrow tensor algebra Big Data

System Model

• Signal model:

 $d(t) = \mathbf{h}^{T}(t)\mathbf{X}(t)\mathbf{g}(t) + \mathbf{v}(t)$

 \rightarrow bilinear form with respect to the impulse responses

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System impulse responses:

$$\begin{split} \mathbf{h}(t) &= \mathbf{h}(t-1) + \mathbf{w}_{\mathbf{h}}(t) \qquad \mathbf{g}(t) = \mathbf{g}(t-1) + \mathbf{w}_{\mathbf{g}}(t) \\ & \mathbf{w}_{\mathbf{h}}(t), \, \mathbf{w}_{\mathbf{g}}(t): \, \text{zero-mean WGN} \\ & \mathbf{R}_{\mathbf{w}_{\mathbf{h}}}(t) = \sigma_{w_{\mathbf{h}}}^2 \mathbf{I}_L \qquad \mathbf{R}_{\mathbf{w}_{\mathbf{g}}}(t) = \sigma_{w_{\mathbf{g}}}^2 \mathbf{I}_M \end{split}$$

System Model

Signal model:

 $d(t) = \mathbf{h}^{T}(t)\mathbf{X}(t)\mathbf{q}(t) + \mathbf{v}(t)$

 \rightarrow bilinear form with respect to the impulse responses

System impulse responses:

 $h(t) = h(t-1) + w_h(t)$ $\mathbf{g}(t) = \mathbf{g}(t-1) + \mathbf{w}_{\mathbf{q}}(t)$ $\mathbf{w}_{\mathbf{h}}(t), \mathbf{w}_{\mathbf{q}}(t)$: zero-mean WGN $\mathbf{R}_{\mathbf{W}_{\mathbf{b}}}(t) = \sigma_{\mathbf{W}_{\mathbf{b}}}^2 \mathbf{I}_{\mathbf{b}}$ $\mathbf{R}_{\mathbf{w}_{\mathbf{q}}}(t) = \sigma_{\mathbf{w}_{\mathbf{q}}}^{2} \mathbf{I}_{M}$

Equivalent model: $d(t) = \mathbf{f}^{T}(t)\widetilde{\mathbf{x}}(t) + \mathbf{v}(t)$

> $\mathbf{f}(t) = \mathbf{q}(t) \otimes \mathbf{h}(t) - Kronecker product of length ML$ $\widetilde{\mathbf{x}}(t) = \operatorname{vec}[\mathbf{X}(t)] = [\mathbf{x}_1^T(t) \ \mathbf{x}_2^T(t) \ \dots \ \mathbf{x}_M^T(t)]^T$

3 + 4 = +

Scaling Ambiguity

•
$$\mathbf{f}(t) = \mathbf{g}(t) \otimes \mathbf{h}(t) = [\eta \mathbf{g}(t)] \otimes \left[\frac{1}{\eta}\mathbf{h}(t)\right] \quad \eta \in \mathcal{R}^*$$
 - scaling factor
 $\left[\frac{1}{\eta}\mathbf{h}(t)\right]^T \mathbf{X}(t) [\eta \mathbf{g}(t)] = \mathbf{h}^T(t)\mathbf{X}(t)\mathbf{g}(t)$

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• Normalized projection misalignment (NPM) ¹
NPM[
$$\mathbf{h}(t), \widehat{\mathbf{h}}(t)$$
] = 1 - $\left[\frac{\mathbf{h}^{T}(t)\widehat{\mathbf{h}}(t)}{||\mathbf{h}(t)||||\widehat{\mathbf{h}}(t)||}\right]^{2}$ NPM[$\mathbf{g}(t), \widehat{\mathbf{g}}(t)$] = 1 - $\left[\frac{\mathbf{g}^{T}(t)\widehat{\mathbf{g}}(t)}{||\mathbf{g}(t)||||\widehat{\mathbf{g}}(t)||}\right]^{2}$

• Normalized misalignment (NM) $NM[\mathbf{f}(t), \widehat{\mathbf{f}}(t)] = \|\mathbf{f}(t) - \widehat{\mathbf{f}}(t)\|^2 / \|\mathbf{f}(t)\|^2$

¹ [Morgan et al., IEEE Signal Processing Letters, July 1998]

Kalman Filter for Bilinear Forms (KF - BF)

• estimated output signal: $\hat{y}(t) = \hat{\mathbf{h}}^T(t-1)\mathbf{X}(t)\hat{\mathbf{g}}(t-1)$

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$$\begin{split} \boldsymbol{e}(t) &= \boldsymbol{d}(t) - \widehat{\mathbf{f}}^{\mathsf{T}}(t-1)\widetilde{\mathbf{x}}(t) \\ &= \boldsymbol{d}(t) - \widehat{\mathbf{h}}^{\mathsf{T}}(t-1)\widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}(t) \leftarrow \boldsymbol{e}_{\widehat{\mathbf{g}}}(t) \\ &= \boldsymbol{d}(t) - \widehat{\mathbf{g}}^{\mathsf{T}}(t-1)\widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}(t) \leftarrow \boldsymbol{e}_{\widehat{\mathbf{h}}}(t) \end{split}$$

$$\widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}(t) = [\widehat{\mathbf{g}}(t-1) \otimes \mathbf{I}_L]^T \widetilde{\mathbf{x}}(t) \qquad \widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}(t) = [\mathbf{I}_M \otimes \widehat{\mathbf{h}}(t-1)]^T \widetilde{\mathbf{x}}(t)$$

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optimal estimates of the state vectors:

 $\widehat{\mathbf{h}}(t) = \widehat{\mathbf{h}}(t-1) + \mathbf{k}_{\mathbf{h}}(t)\mathbf{e}(t)$ $\widehat{\mathbf{g}}(t) = \widehat{\mathbf{g}}(t-1) + \mathbf{k}_{\mathbf{g}}(t)\mathbf{e}(t)$

$$\mathbf{k}_{\mathbf{h}}(t), \, \mathbf{k}_{\mathbf{g}}(t)$$
: Kalman gain vectors

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- a posteriori misalignments: $\mu_{\mathbf{h}}(t) = \mathbf{h}(t)/\eta - \widehat{\mathbf{h}}(t)$
- a priori misalignments:

 $\mathbf{m}_{\mathbf{h}}(t) = \mathbf{h}(t)/\eta - \widehat{\mathbf{h}}(t-1)$ $= \mu_{\mathbf{h}}(t-1) + \mathbf{w}_{\mathbf{h}}(t)/\eta$

 $\boldsymbol{\mu}_{\mathbf{g}}(t) = \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t)$

$$\mathbf{m}_{\mathbf{g}}(t) = \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t-1)$$
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$$\begin{split} \mathbf{m}_{\mathbf{g}}(t) &= \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t-1) \\ &= \boldsymbol{\mu}_{\mathbf{g}}(t-1) + \eta \mathbf{w}_{\mathbf{g}}(t) \end{split}$$

 $\boldsymbol{\mu}_{\mathbf{q}}(t) = \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t)$

simplifying notations:

$$\begin{split} \overline{\mathbf{w}}_{\mathbf{h}}(t) &= \mathbf{w}_{\mathbf{h}}(t) / \eta \\ \mathbf{R}_{\mathbf{m}_{\mathbf{h}}}(t) &= \mathbf{R}_{\boldsymbol{\mu}_{\mathbf{h}}}(t-1) + \sigma_{\overline{w}_{\mathbf{h}}}^2 \mathbf{I}_L \end{split}$$

$$\overline{\mathbf{w}}_{\mathbf{g}}(t) = \eta \mathbf{w}_{\mathbf{g}}(t)$$

 $\mathbf{R}_{\mathbf{m}_{\mathbf{g}}}(t) = \mathbf{R}_{\boldsymbol{\mu}_{\mathbf{g}}}(t-1) + \sigma_{\overline{w}_{\mathbf{g}}}^2 \mathbf{I}_M$

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$$\mathbf{m}_{\mathbf{g}}(t) = \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t-1)$$
$$= \mu_{\mathbf{a}}(t-1) + \eta \mathbf{w}_{\mathbf{a}}(t)$$

 $\boldsymbol{\mu}_{\mathbf{q}}(t) = \eta \mathbf{g}(t) - \widehat{\mathbf{g}}(t)$

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• minimizing (1/L)tr $[\mathbf{R}_{\mu_{\mathbf{h}}}(t)]$, (1/M)tr $[\mathbf{R}_{\mu_{\mathbf{g}}}(t)]$ yields: $\mathbf{k}_{\mathbf{h}}(t) = \mathbf{R}_{\mathbf{m}_{\mathbf{h}}}(t)\widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}(t)[\widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}^{T}(t)\mathbf{R}_{\mathbf{m}_{\mathbf{h}}}(t)\widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}(t) + \sigma_{v}^{2}]^{-1}$ $\mathbf{k}_{\mathbf{g}}(t) = \mathbf{R}_{\mathbf{m}_{\mathbf{g}}}(t)\widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}(t)[\widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}^{T}(t)\mathbf{R}_{\mathbf{m}_{\mathbf{g}}}(t)\widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}(t) + \sigma_{v}^{2}]^{-1}$

Simplified Kalman Filter for Bilinear Forms (SKF - BF)

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- $\bullet\,$ misalignments of the individual coefficients: uncorrelated $\rightarrow\,$

$$\begin{array}{l} \rightarrow \text{ approximate:} \\ \mathbf{I}_{\mathcal{L}} - \mathbf{k}_{\mathbf{h}}(t) \widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}^{T}(t) \approx \left[1 - \frac{1}{\mathcal{L}} \mathbf{k}_{\mathbf{h}}^{T}(t) \widetilde{\mathbf{x}}_{\widehat{\mathbf{g}}}(t)\right] \mathbf{I}_{\mathcal{L}} \\ \mathbf{I}_{\mathcal{M}} - \mathbf{k}_{\mathbf{g}}(t) \widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}^{T}(t) \approx \left[1 - \frac{1}{\mathcal{M}} \mathbf{k}_{\mathbf{g}}^{T}(t) \widetilde{\mathbf{x}}_{\widehat{\mathbf{h}}}(t)\right] \mathbf{I}_{\mathcal{M}} \end{array}$$

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 \Rightarrow Simplified Kalman Filter for bilinear forms (SKF - BF)

• The parameters related to uncertainties in **h**, **g**: $\sigma_{\overline{w}_{h}}^{2}$, $\sigma_{\overline{w}_{h}}^{2}$.

- $\bullet \ \text{small} \Rightarrow \text{good misalignment, poor tracking}$
- large (i.e., high uncertainty in the systems) \Rightarrow
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- A good compromise is needed!
- In practice \rightarrow some a priori information (e.g., if **g** - time-invariant $\Rightarrow \sigma_{\overline{w}_{g}}^{2} = 0$)
- By applying the ℓ_2 norm on the state equation:

$$\widehat{\sigma}_{\overline{w}_{\mathbf{h}}}^{2}(t) = \frac{1}{L} \left\| \widehat{\mathbf{h}}(t) - \widehat{\mathbf{h}}(t-1) \right\|_{2}^{2}$$

Conditions

- input signals x_m(t), m = 1, 2, ..., M independent WGN, respectively AR(1) generated by filtering a white Gaussian noise through a first-order system 1/(1 − 0.8z⁻¹)
- h, g Gaussian, randomly generated, of lengths L = 64, M = 8
- v(t) independent white Gaussian noise signal

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Compared algorithms

- KF-BF and KF
- SKF-BF and SKF when $\sigma_{\overline{w}_{h}}^{2} = \sigma_{\overline{w}_{a}}^{2} = \sigma_{w}^{2} = 10^{-9}$
- SKF-BF and SKF when $\sigma_{\overline{w}_{g}}^{2} = 0$ and $\widehat{\sigma}_{\overline{w}_{h}}^{2}(t) = \frac{\|\widehat{\mathbf{h}}(t) \widehat{\mathbf{h}}(t-1)\|_{2}^{2}}{L}$



Figure 1: Normalized misalignment of the KF-BF and regular KF for different types of input signals. ML = 512, $\sigma_v^2 = 0.01$, $\sigma_{\overline{w}_h}^2 = \sigma_{\overline{w}_g}^2 = \sigma_w^2 = 10^{-9}$, and $\epsilon = 10^{-5}$.



Figure 2: Normalized misalignment of the SKF-BF and regular SKF for different types of input signals. Other conditions are the same as in Fig. 1.

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Figure 3: Normalized misalignment of the SKF-BF and regular SKF for different types of input signals, using the recursive estimates $\hat{\sigma}_{w_h}^2(t)$ and $\hat{\sigma}_w^2(t)$, respectively. $ML = 512, \sigma_v^2 = 0.01, \sigma_w^2 = 0, \text{ and } \epsilon = 10^{-5}.$ • **KF-BF, SKF-BF**: improvement in convergence rate and tracking with respect to regular KF, SKF

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- **KF-BF, SKF-BF**: improvement in convergence rate and tracking with respect to regular KF, SKF
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- SKF-BF provides:
 - reduced computational complexity , but also
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The experimental results indicate the good performance of the proposed algorithms

Thank you!

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