

Can Hardware Distortion Correlation be Neglected When Analyzing Uplink SE in Massive MIMO?

Emil Björnson¹, Luca Sanguinetti², Jakob Hoydis³

¹Department of Electrical Engineering (ISY), Linköping University, Sweden ²Dipartimento di Ingegneria dell'Informazione, University of Pisa, Pisa, Italy ³Nokia Bell Labs, Nozay, France



Università di Pisa

Background

The received signals at a Massive MIMO base station (BS) are correlated between the antennas

- Hardware distortion also becomes correlated
- But the correlation is often † neglected when analyzing the spectral efficiency (SE)
- Recent works call this approximation "physically inaccurate" but does it lead to inaccurate results?

No, we prove that the approximation errors are small



Received signal (at antenna inputs):

$$\mathbf{u} = \begin{bmatrix} u_1 \\ \vdots \\ u_M \end{bmatrix} = \begin{bmatrix} \mathbf{h}_1 & \dots & \mathbf{h}_K \end{bmatrix} \begin{bmatrix} s_1 \\ \vdots \\ s_K \end{bmatrix} = \mathbf{Hs}$$

Received signal is distorted by the hardware:

$$\mathbf{u} \longrightarrow \boxed{ \begin{array}{c} \operatorname{Non-ideal hardware} \\ g(\mathbf{u}) \end{array} } \mathbf{Du} + \eta$$

• Arbitrary function $\boldsymbol{g}(\mathbf{u}) = [g_1(u_1) \dots g_M(u_M)]^{\mathsf{T}}$

Conditional statistics for given channel realization H:

- $\mathbf{D} = \operatorname{diag}(d_1, \dots, d_M)$ and $d_m = \frac{\mathbb{E}_{|\mathbf{H}}\{g_m(u_m)u_m^*\}}{\mathbb{E}_{|\mathbf{H}}\{|u_m|^2\}}$
- Distortion $\eta \in \mathbb{C}^M$ is uncorrelated with **u** and has non-diagonal correlation matrix $\mathbf{C}_{\eta\eta} = \mathbb{E}_{|\mathbf{H}} \{ \eta \eta^{\scriptscriptstyle H} \}$

Output of non-ideal hardware is a scaled version of input **u** plus distortion η that has correlated elements but is uncorrelated with **u**.

Spectral Efficiency

Noisy signal used for detection:

$$\mathbf{y} = oldsymbol{g}(\mathbf{u}) + \mathbf{n} = \mathbf{D}\mathbf{u} + oldsymbol{\eta} + \mathbf{n}$$

• Noise: $\mathbf{n} \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \sigma^2 \mathbf{I}_M)$

Hardware impairments at user-side ($\kappa \in [0,1]$):

• $s_k = \varsigma_k + \omega_k$, with desired signal $\varsigma_k \sim \mathcal{N}_{\mathbb{C}}(0, \kappa p)$ and transmitter distortion $\omega_k \sim \mathcal{N}_{\mathbb{C}}(0, (1 - \kappa)p)$

With perfect CSI and treating interference/distortion

Quantifying Impact of Non-Linearities

Model the low-noise amplifier as third-order strictly memoryless non-linear function

$$g_m(u_m) = u_m - a_m |u_m|^2 u_m, \quad m = 1, \dots, M$$

where $a_m = \frac{\alpha}{b_{\text{off}} \mathbb{E}\{|u_m|^2\}}$, α characterizes saturation level, and b_{off} is the backoff



We obtain **D** and $\mathbf{C}_{\eta\eta}$: (\odot = elementwise product)

$$\mathbf{D} = \mathbf{I}_M - 2\mathbf{A} \odot \mathbf{C}_{uu}$$
$$\mathbf{C}_{\eta\eta} = 2\mathbf{A} \left(\mathbf{C}_{uu} \odot \mathbf{C}_{uu}^* \odot \mathbf{C}_{uu} \right) \mathbf{A}$$

with $\mathbf{C}_{uu} = \mathbb{E}_{|\mathbf{H}} \{ \mathbf{uu}^{\scriptscriptstyle H} \} = p \mathbf{H} \mathbf{H}^{\scriptscriptstyle H}, \mathbf{A} = \operatorname{diag}(a_1, \dots, a_M)$

Distortion Vectors Less Correlated Than Signals

Correlation coefficient for signals u_i and u_j :

$$\xi_{u_i u_j} = \frac{\rho_{ij}}{\sqrt{\rho_{ii}\rho_{jj}}} \in [0,1]$$

where $\rho_{ij} = \mathbb{E}_{|\mathbf{H}}\{u_i u_j^*\} = [\mathbf{C}_{uu}]_{ij}$

Correlation coefficient for distortion terms η_i and η_j :

$$\xi_{\eta_i\eta_j} = \frac{\mathbb{E}_{|\mathbf{H}}\{\eta_i\eta_j^*\}}{\sqrt{\mathbb{E}_{|\mathbf{H}}\{|\eta_i|^2\}\mathbb{E}_{|\mathbf{H}}\{|\eta_j|^2\}}} = |\xi_{u_iu_j}|^2\xi_{u_iu_j}$$

The distortion terms are less correlated than the corresponding signal terms, since

 $|\xi_{\eta_i\eta_j}| = |\xi_{u_iu_j}|^3 \le |\xi_{u_iu_j}|.$

What if the Distortion Correlation is Neglected?

If distortion correlation is low: analytically tractable to neglect it. Use $\mathbf{C}_{\eta\eta}^{\text{diag}} = \mathbf{C}_{\eta\eta} \odot \mathbf{I}_M$ instead of $\mathbf{C}_{\eta\eta}$?

Assumption: i.i.d. Rayleigh fading channels $\mathbf{h}_k \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0}, \mathbf{I}_M)$ for $k = 1, \dots, K$

Consider maximum ratio (MR) combining $\mathbf{v}_k = \mathbf{h}_k / \sqrt{\mathbb{E}\{\|\mathbf{h}_k\|^2\}}$, then

$$\begin{split} & \left| \mathbb{E} \{ \mathbf{v}_k^{\mathsf{H}} \mathbf{C}_{\eta\eta} \mathbf{v}_k \} = c \left(K + 6 + \frac{9}{K} + \frac{4 + 2M(K+1)}{K^2} \right) \right| \\ & \approx \mathbb{E} \{ \mathbf{v}_k^{\mathsf{H}} \mathbf{C}_{\eta\eta}^{\mathsf{diag}} \mathbf{v}_k \} = c \left(K + 6 + \frac{11}{K} + \frac{6}{K^2} \right) \end{split}$$

Relative Size of BS and User Distortion

Massive MIMO simulation setup:

 $M=200,\,\alpha=1/3,\,b_{\rm off}=7\,{\rm dB},\,\kappa=0.99,$ and a signal-to-noise ratio (SNR) of $p/\sigma^2=0\,{\rm dB}$

• BS hardware is "worse" than the user hardware (larger signal-to-distortion power ratio)



- Distortion correlation has a huge impact on BS distortion when there are few users
- Gap reduces from 15.3 to 5.5 dB in shaded area
- The correlated BS distortion is only the dominant factor for $K \leq 3$, since some terms reduce as $1/K^2$

The correlation of the BS distortion reduces with K. The BS distortion eventually has a smaller impact than the user distortion, which doesn't reduce with K.

SE With and Without Distortion Correlation

We compute SE using $\mathbf{C}_{\eta\eta}$ and $\mathbf{C}_{\eta\eta}^{\text{diag}}$

We compare DA-MMSE and distortion-aware MR (DA-MR) combining $(\mathbf{v}_k = \mathbf{D}\mathbf{h}_k / \|\mathbf{D}\mathbf{h}_k\|)$



- Choice of combining scheme has a large impact
- Approximation error is negligible for $K \ge 5$ with both schemes
- For K < 5, the shaded gap only ranges from 6.7% to 5.5% for DA-MMSE

The distortion correlation has negligible impact on the uplink SE in the studied Massive MIMO scenario; that is, i.i.d. Rayleigh fading and equal SNRs for all users.

as noise, the SE for user k is $\mathbb{E}_{\mathbf{H}}\{\log_2(1+\gamma'_k)\}, \gamma'_k =$ $\kappa p |\mathbf{h}_k^{\scriptscriptstyle \mathrm{H}} \mathbf{D}^{\scriptscriptstyle \mathrm{H}} \mathbf{v}_k|^2$ $\frac{1}{\sum_{i \neq k} p |\mathbf{h}_i^{\mathrm{H}} \mathbf{D}^{\mathrm{H}} \mathbf{v}_k|^2 + \mathbf{v}_k^{\mathrm{H}} \mathbf{C}_{\eta \eta} \mathbf{v}_k + (1 - \kappa) p |\mathbf{h}_k^{\mathrm{H}} \mathbf{D}^{\mathrm{H}} \mathbf{v}_k|^2 + \sigma^2 \|\mathbf{v}_k\|^2}$

- Receive combining vector \mathbf{v}_k
- BS distortion: $\mathbf{v}_k^{\text{H}} \mathbf{C}_{\eta\eta} \mathbf{v}_k$
- User distortion: $(1\!-\!\kappa)p|\mathbf{h}_k^{\scriptscriptstyle \mathrm{H}}\mathbf{D}^{\scriptscriptstyle \mathrm{H}}\mathbf{v}_k|^2$

SE maximized by *distortion-aware minimum-mean squared error* (DA-MMSE) combining:

$$\mathbf{v}_k = p \left(\sum_{i=1, i \neq k}^{K} p \mathbf{D} \mathbf{h}_i \mathbf{h}_i^{\mathsf{H}} \mathbf{D}^{\mathsf{H}} + \mathbf{C}_{\eta\eta} + \sigma^2 \mathbf{I}_M \right)^{-1} \mathbf{D} \mathbf{h}_k.$$

The optimal receive combining takes the BS distortion correlation into account, but is unaffected by the user distortion.

where $c = \frac{2\alpha^2 p}{b_{off}^2}$ and the approximation neglects the distortion correlation.

The average distortion power is larger when the BS distortion is correlated:

$$\frac{\mathbb{E}\{\mathbf{v}_{k}^{\mathsf{H}}\mathbf{C}_{\eta\eta}\mathbf{v}_{k}\}}{\mathbb{E}\{\mathbf{v}_{k}^{\mathsf{H}}\mathbf{C}_{\eta\eta}^{\mathsf{diag}}\mathbf{v}_{k}\}} = 1 + \frac{2(M-1)}{(K+2)(K+3)}$$

Second term grows as M and decays as $1/K^2$

$$\begin{split} & \left[\begin{array}{l} \text{Under same assumptions, average user distortion is } (1-\kappa)p\mathbb{E}\{|\mathbf{h}_k^{\scriptscriptstyle H}\mathbf{D}^{\scriptscriptstyle H}\mathbf{v}_k|^2\} \text{ with} \\ & \mathbb{E}\{|\mathbf{h}_k^{\scriptscriptstyle H}\mathbf{D}^{\scriptscriptstyle H}\mathbf{v}_k|^2\} = (M+1) - \frac{4\alpha(MK+K+M+3)}{b_{\text{off}}K} \\ & + \frac{4\alpha^2(MK^2+8K+11+2MK+K^2+M)}{b_{\text{off}}^2K^2} \end{split} \right] \end{split}$$

Conclusion

Yes, hardware distortion correlation can be neglected when analyzing uplink SE in Massive MIMO!

Our journal paper proves the conclusion with

- Different channel models and varying SNRs
- Quantization distortion
- Imperfect CSI and as $M \to \infty$

In practice, frequency-selective fading and compensation algorithms further supports the conclusion

But, one can create setups (ideal user hardware, freespace propagation) when the correlation is influential

[†]Our book "Massive MIMO Networks: Spectral, Energy, and Hardware Efficiency" reviews how to quantify the SE with uncorrelated distortion. We would happily give you a copy!