

Spatially Oversampled Demultiplexing in mmWave LoS MIMO

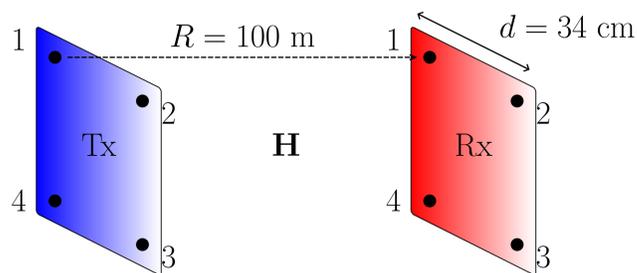
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The potential of LoS MIMO

- * Spatial degrees of freedom $\sim (A_t A_r) / (\lambda R)^2 \sim f_c^2$ (fixed form factor) [1]
 A_t, A_r – transmit and receive antenna apertures
- * Available bandwidth $\sim f_c$
- * Data rates $\sim f_c^3$ – strong incentive to push up carrier frequency to mmWave

Example: 4×4 MIMO at 130GHz



→ Channel for ideally aligned LoS MIMO system

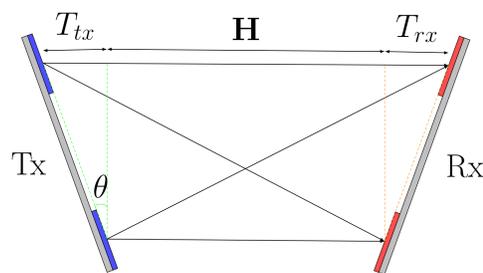
$$\mathbf{H} = \begin{bmatrix} 1 & e^{-j\phi} & e^{-2j\phi} & e^{-j\phi} \\ e^{-j\phi} & 1 & e^{-j\phi} & e^{-2j\phi} \\ e^{-2j\phi} & e^{-j\phi} & 1 & e^{-j\phi} \\ e^{-j\phi} & e^{-2j\phi} & e^{-j\phi} & 1 \end{bmatrix}, \phi = \pi d^2 / \lambda R$$

\mathbf{H} is invertible, if $d = \sqrt{R\lambda/2}$ (34 cm at 130 GHz)

→ Achievable data rates > 100 Gbps
 $20 \times 4 \times 2 = 160$ Gbps (20 GBaud, QPSK)

Spatial demux at 10s of GBaud: How?

- * Simple approach: invert the channel \mathbf{H}
- * Vulnerable to even small misalignments



- * Even a small $\theta(5^\circ)$ introduce 2 channel symbol delay (T_{tx}) at 20 GHz symbol rate ($\frac{d \tan(\theta)}{TC} \approx 2$)
- * Channel becomes frequency selective [2]

$$\begin{bmatrix} \mathbf{y}_K \\ \mathbf{y}_{K-1} \\ \vdots \\ \mathbf{y}_{K-W+1} \end{bmatrix} = \begin{bmatrix} \mathbf{H}_0 & \cdots & \mathbf{H}_{L-1} & \cdots & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{H}_0 & \cdots & \mathbf{H}_{L-1} & \cdots & \mathbf{0} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ \mathbf{0} & \mathbf{0} & \cdots & \mathbf{H}_0 & \cdots & \mathbf{H}_{L-1} \end{bmatrix} \begin{bmatrix} \mathbf{x}_K \\ \mathbf{x}_{K-1} \\ \vdots \\ \mathbf{x}_{K-L-W+1} \end{bmatrix}$$

Conventional DSP architecture ⇒ performance floor

- Symbol rate FIR space-time equalizer → performance floor
- Temporal oversampling out of the question at 20 GBaud

Prior analog approach

- Digitally programmable analog delays – introduce artificial analog delays to mimic the ideal channel

Our observation

- ◇ Channel can be inverted with appropriate spatial oversampling
- ◇ FIR space-time equalizers with symbol rate sampling
- ◇ Small number of taps → both fully digital and hybrid implementations become possible

Spatially oversampled reception

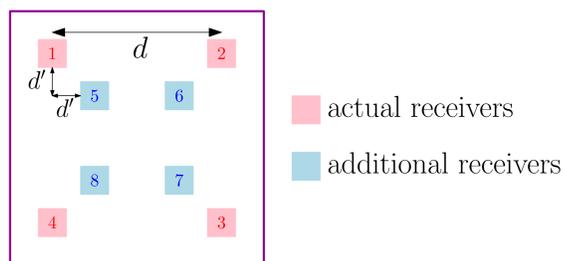


Figure 1: 2X spatial oversampling within the same form factor.

- ✓ Spatial oversampling is natural as distance b/w actual receivers $d = 34$ cm $\gg \lambda$

Basic ZF conditions

- 1 $N_r \geq N_t$
- 2 The window length
$$W \geq W_0 = \left\lceil \frac{L-1}{\left(\frac{N_r}{N_t} - 1\right)} \right\rceil$$
- 3 \mathbf{H} has full column rank

Table 1: Trade-off between N_r and W_0 for $L = 6$ and $N_t = 4$

N_r	5	6	7	8
W_0	20	10	7	5

- ✓ Time domain complexity reduces as spatial oversampling increases

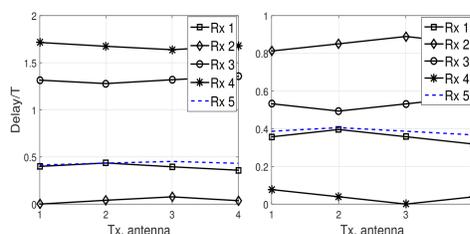


Figure 2: The variation of delays at the different receive antennas

- \mathbf{H} is not a well-conditioned channel!
- Offset sampling (T/2 a robust choice) for additional antennas helps in obtaining a well-conditioned channel

Numerical results

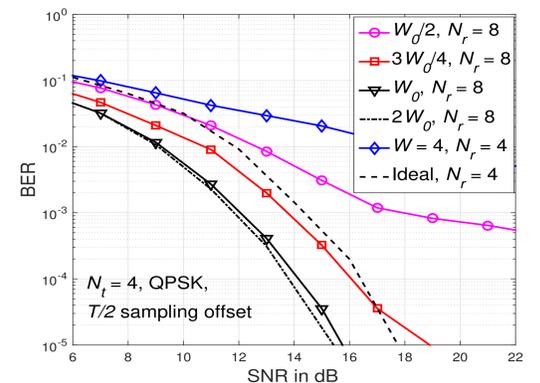


Figure 3: BER of the proposed oversampled LoS MIMO system for different window lengths with $N_t = 4$ and QPSK

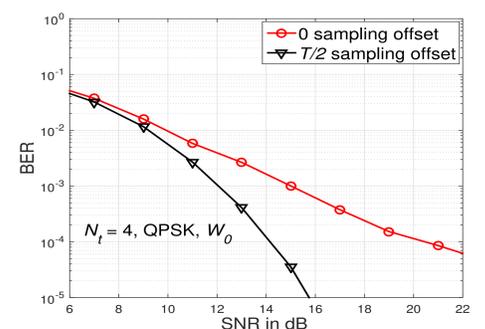


Figure 4: BER of the proposed oversampled LoS MIMO system for different sampling offsets at extra receive antennas

- Performance improves with window length until W_0 and saturates thereafter
- Window length of W_0 and T/2 sampling offset are necessary to avoid error floors
- Misaligned system with $N_r = 8$ performs better than an ideally aligned system with $N_r = 4$ due to better noise averaging

Conclusion

- 1 Geometric misalignments in LoS MIMO cause frequency selectivity
- 2 Spatial oversampling, along with designed delay diversity, is an effective approach to combat the frequency selectivity
- 3 Can trade spatial oversampling against time complexity
- 4 Particularly attractive architecture: double the number of receivers within the same form factor

References

- [1] E. Torkildson, U. Madhow, and M. Rodwell, "Indoor millimeter wave MIMO: Feasibility and performance," *IEEE Trans. Wireless Commun.*, vol. 10, no. 12, pp. 4150-4160, Dec. 2011.
- [2] B. Mamandipoor, M. Sawaby, A. Arbabian, and U. Madhow, "Hardware-constrained signal processing for mmWave LoS MIMO," in *IEEE 49th Asilomar Conference on Signals, Systems and Computers*, Pacific Grove, CA, Nov. 2015.

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