

Multi-User MIMO-OFDM for Indoor Visible Light Communication Systems

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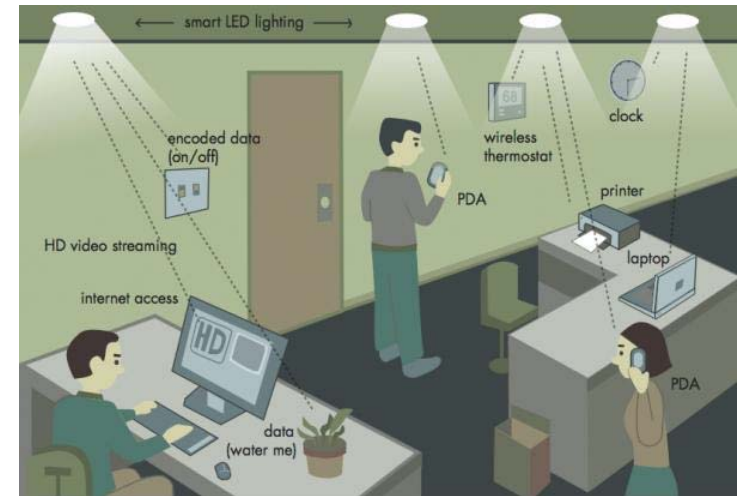
Visible light communication

▶ Advantages

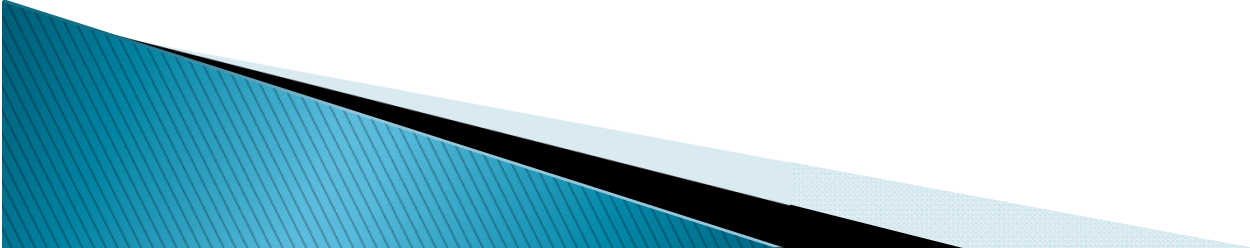
- No electromagnetic interference
- High security
- Low cost

▶ Multiple LED units are usually installed in a single room

- Provide sufficient illumination
- Bandwidth of off-the-shelf LED is limited
- MIMO can be naturally used to boost the data rate

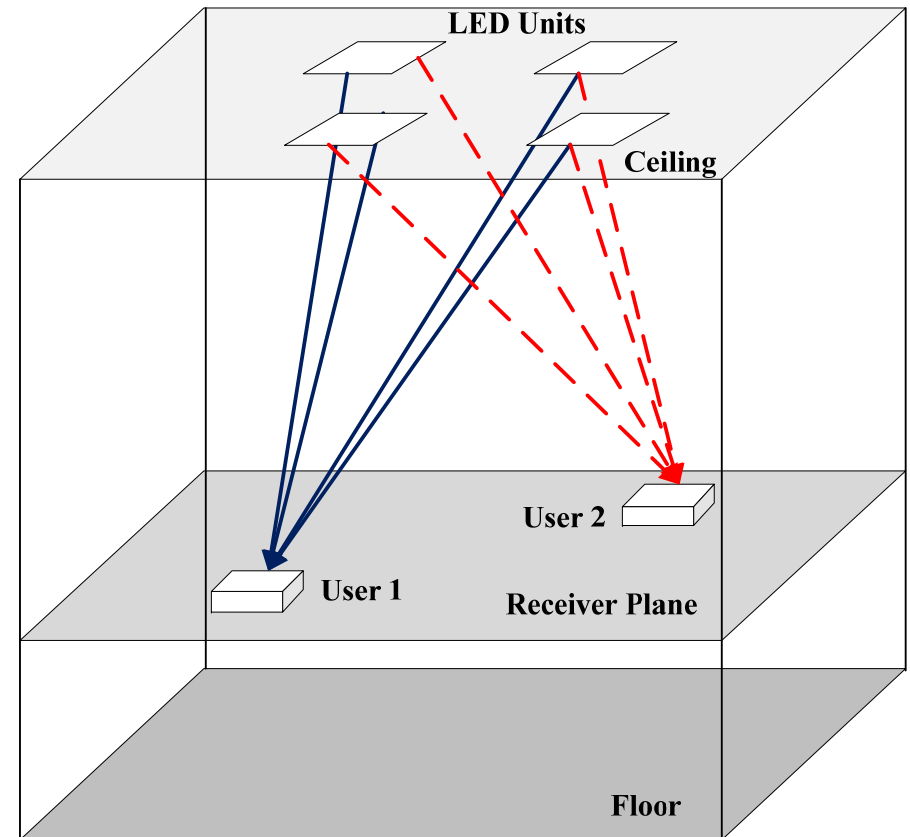


Multuser MIMO VLC

- ▶ Precoding
 - Zero forcing
 - Dirty paper coding
 - Minimum mean-squared error (MMSE)
 - Block diagonalization
 - ▶ OFDM → High spectral efficiency
 - ▶ MIMO-OFDM
- 

System Model

- ▶ N_t LED units
- ▶ N_r users with single PD



System Model

- ▶ DC channel gain
$$h_{p,q}^{DC} = \begin{cases} \frac{\rho_p A_p}{d_{p,q}^2} R(\phi_q) \cos(\varphi_{p,q}), \varphi_{p,q} \leq \Psi_{c,p} \\ 0, \varphi_{p,q} > \Psi_{c,p} \end{cases}$$
- ρ_p : responsivity coefficient of PD
 - $d_{p,q}$: distance between the qth LED unit and the pth user
 - $\varphi_{p,q}$: incidence angle of the light
 - $\Psi_{c,p}$: receiver FOV

▶ Noise

$$\sigma_p^2 = 2eP_p B + 2e\rho_p \chi_{amb} A_p \left(1 - \cos(\Psi_{c,p})\right) B + i_{amp}^2$$

- P_p : the average received optical power at the pth user

MU-MIMO-OFDM for VLC

- ▶ In existing MU-MIMO VLC systems, single-carrier modulations are utilized
- ▶ Time domain precoding
- ▶ Distances of the multiple transmitter-receiver links are different \rightarrow different delays

$$h_{p,q}(t) = h_{p,q}^{DC} \delta\left(t - \frac{d_{p,q}}{c}\right)$$

- ▶ Frequency-domain channel response

$$H_{p,q,k} = h_{p,q}^{DC} \exp\left(-\frac{j2\pi k B d_{p,q}}{Nc}\right)$$

- ▶ Complex-valued

MU-MIMO-OFDM for VLC

- ▶ Precoding is performed on each subcarrier to eliminate multiuser interference
 - Data: $D_{p,k}$ $D_{p,k} = D_{p,N-k}^*$
 - Precoding weights: $W_{p,q,k}$
 - Frequency-domain signal: $X_{q,k} = \sum_{p=1}^{N_r} W_{p,q,k} D_{p,k}, k = 0, 1, \dots, N-1$
- ▶ After IFFT

$$x_{q,n} = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} X_{q,k} \exp\left(j \frac{2\pi}{N} nk\right), n = 0, 1, \dots, N-1$$

- ▶ CP is added to eliminate ISI

Precoding Design

- ▶ At the receiver of the p th user

$$R_{p,k} = \sum_{q=1}^{N_t} H_{p,q,k} X_{q,k} + Z_{p,k} = \mathbf{H}_{p,k}^T \mathbf{W}_{p,k} D_{p,k} + \sum_{l \neq p}^{N_t} \mathbf{H}_{l,k}^T \mathbf{W}_{l,k} D_{l,k} + Z_{p,k}$$

- ▶ Rewrite in the matrix form

$$R_{p,k} = \sum_{q=1}^{N_t} H_{p,q,k} X_{q,k} + Z_{p,k} = \mathbf{H}_{p,k}^T \mathbf{W}_{p,k} D_{p,k} + \sum_{l \neq p}^{N_t} \mathbf{H}_{l,k}^T \mathbf{W}_{l,k} D_{l,k} + Z_{p,k}$$

- ▶ Two simple precoding methods

- Zero-forcing $\mathbf{H}_k \mathbf{W}_k = \text{diag}(\boldsymbol{\lambda}_k)$

$$\mathbf{W}_k = \mathbf{H}_k^\dagger \text{diag}(\boldsymbol{\lambda}_k) = \mathbf{H}_k^H \left(\mathbf{H}_k \mathbf{H}_k^H \right)^{-1} \text{diag}(\boldsymbol{\lambda}_k)$$

- MMSE

$$\mathbf{W}_k = \mathbf{H}_k^H \left(\mathbf{H}_k \mathbf{H}_k^H + \text{diag}(\boldsymbol{\sigma}_{Z_k}^2) \right)^{-1} \text{diag}(\boldsymbol{\lambda}_k)$$

DCO-OFDM-based MU-MIMO

- ▶ Intensity modulation with direct detection
- ▶ Non-negative real signals are transmitted
- ▶ DC bias is added → DCO-OFDM

$$\bar{P}_{DC,q} = \eta \sqrt{E\{x_{q,n}^2\}}$$

- ▶ Low DC bias → clipping distortion
- ▶ High DC bias → inefficient in terms of power
- ▶ Define a minimum DC bias ratio η_0 to avoid clipping distortion
- ▶ Average optical power: P

DCO-OFDM-based MU-MIMO

▶ Minimum DC bias scheme

- Each transmitter use different DC biases

$$\bar{P}_{DC,q} = \eta_0 \sqrt{E\{x_{q,n}^2\}}$$

- Optical power: $P_{opt,q} = E\{x_{q,n} + \bar{P}_{DC,q}\} = E\{x_{q,n}\} + \bar{P}_{DC,q} = \bar{P}_{DC,q}$

$$y_{q,n} = \alpha (x_{q,n} + \bar{P}_{DC,q})$$

- Scaling factor: $\alpha = N_t P / \eta_0 \sum_{q=1}^{N_t} \sqrt{E\{x_{q,n}^2\}}$

- DC bias: $P_{DC,q} = \alpha \bar{P}_{DC,q} = \sqrt{E\{x_{q,n}^2\}} N_t P / \sum_{q=1}^{N_t} \sqrt{E\{x_{q,n}^2\}}$

▶ Unified DC bias scheme

- Each transmitter use the same DC bias

$$\alpha = \frac{P}{\eta_0 \sqrt{\max_{1 \leq q \leq N_t} E\{x_{q,n}^2\}}} \quad P_{DC,q} = P$$

ACO-OFDM-based MU-MIMO

- ▶ Only odd subcarriers are modulated

$$x_{q,n} = -x_{q,n+N/2}, n = 0, 1, \dots, N-1$$

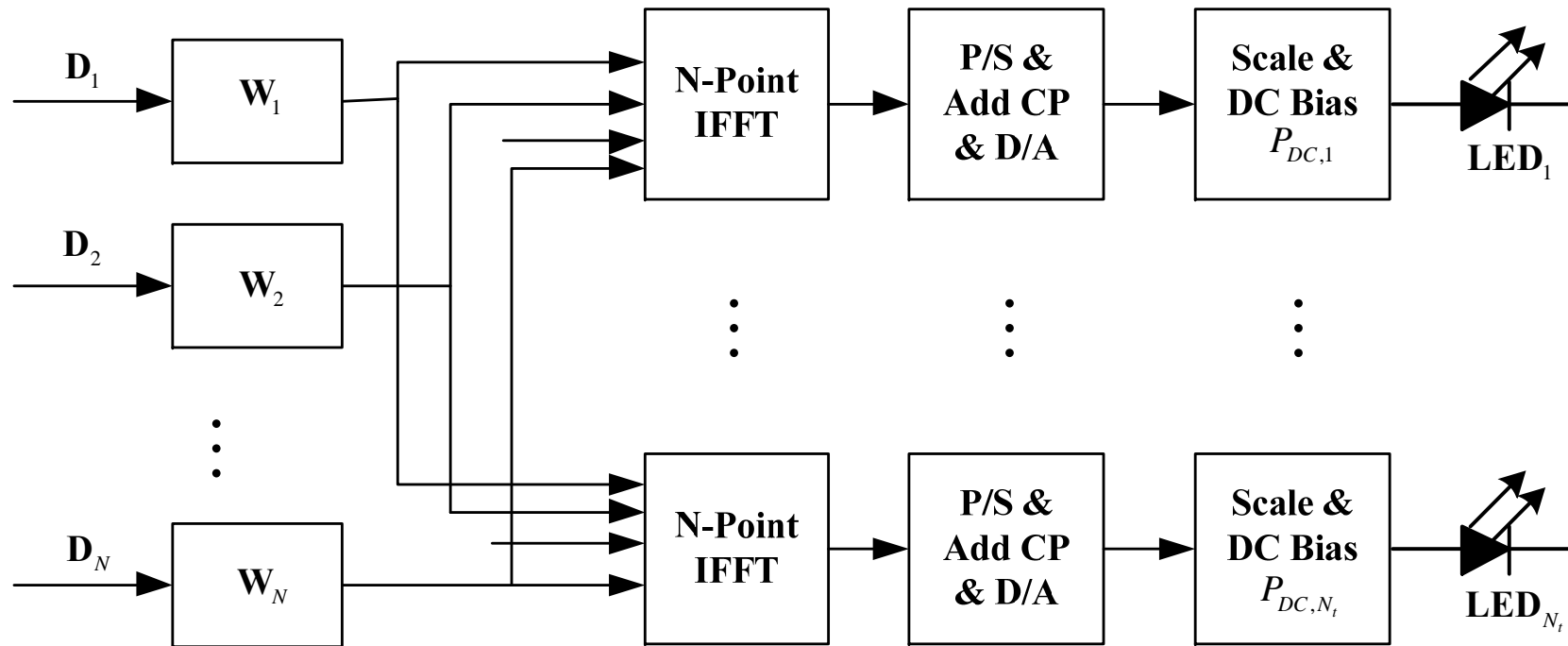
- ▶ Asymmetrically clipping at zero
- ▶ No DC bias: better performance?

- ▶ Optical power: $P_{opt,q} = E\{x_{q,n}^{(c)}\} = \sqrt{E\{x_{q,n}^2\} / 2\pi}$

- ▶ Scaling factor:

$$\alpha = \frac{N_t P}{\sum_{q=1}^{N_t} P_{opt,q}} = \frac{N_t P}{\sum_{q=1}^{N_t} \sqrt{E\{x_{q,n}^2\} / 2\pi}}$$

System Block Diagram



Simulation Results

SIMULATION PARAMETERS FOR VLC SYSTEM CONFIGURATION

Room size (length \times width \times height)	5 m \times 5 m \times 3 m
LED 1 coordinate	[1.25 1.25 3]
LED 2 coordinate	[1.25 3.75 3]
LED 3 coordinate	[3.75 1.25 3]
LED 4 coordinate	[3.75 3.75 3]
LED emission angle ϕ_q	60 deg
PD area $A_{PD,p}$	1 cm ²
PD responsivity coefficient ρ_p	0.4 A/W
PD concentrator refractive index γ	1.5
Lambertian emission mode number m	1
Receiver FOV angle $\Psi_{c,p}$	62 deg
Pre-amplifier noise density i_{amp}	5 pA/Hz ^{-1/2}
Ambient light photocurrent χ_{amp}	10.93 A/m ² /Sr
System bandwidth B	1 GHz
OFDM subcarrier number N	64
Cyclic prefix length N_{CP}	3

Case 1

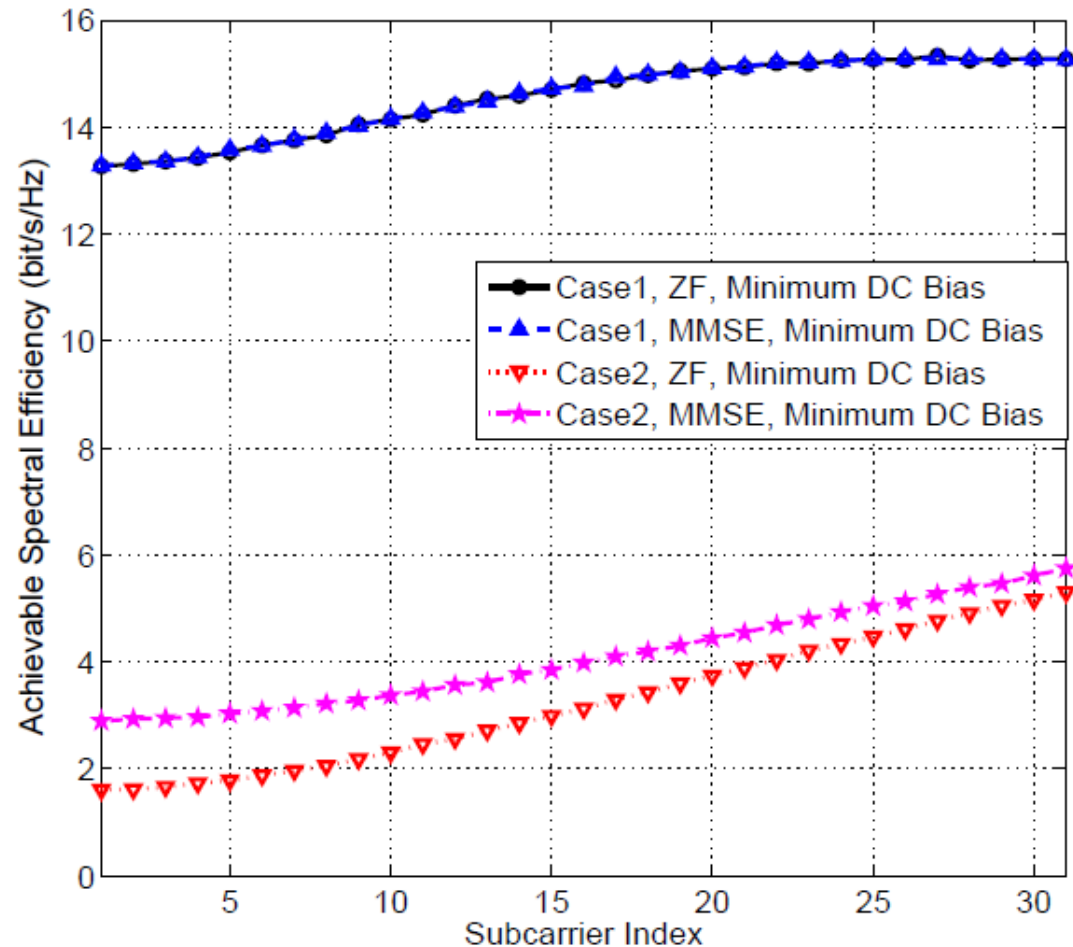
- User 1: [2.5 2.5 0.85]
- User 2: [3.2 3.9 0.85]

Case 2:

- User 1: [2.05 1.6 0.85]
- User 2: [2.05 1.4 0.85]

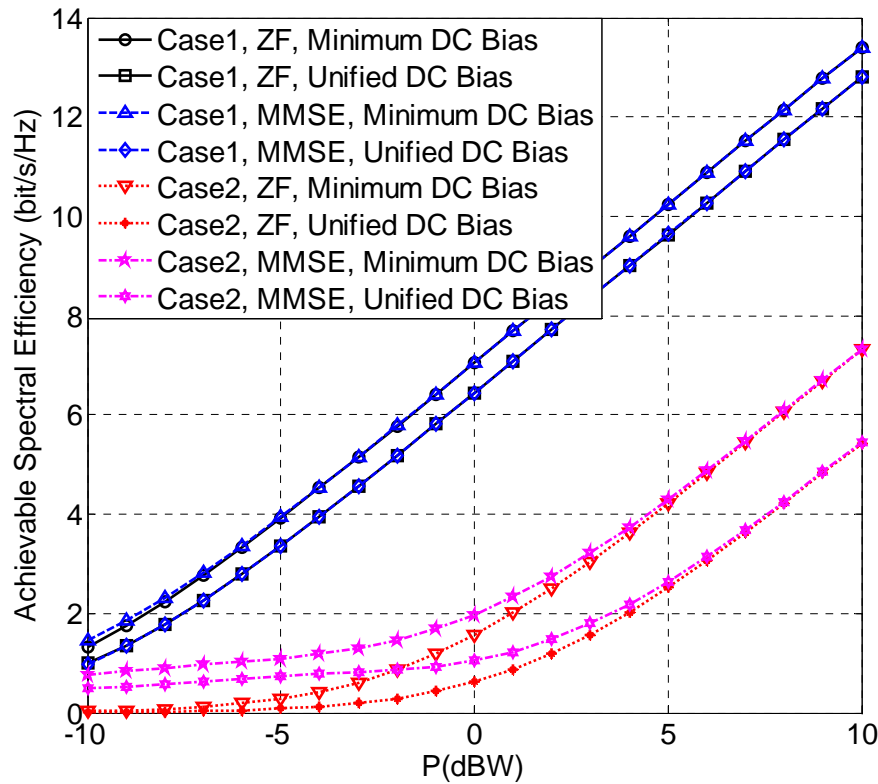
Simulation Results

- ▶ Spectral efficiency of each subcarrier with the average emitted optical power $P=0$ dB

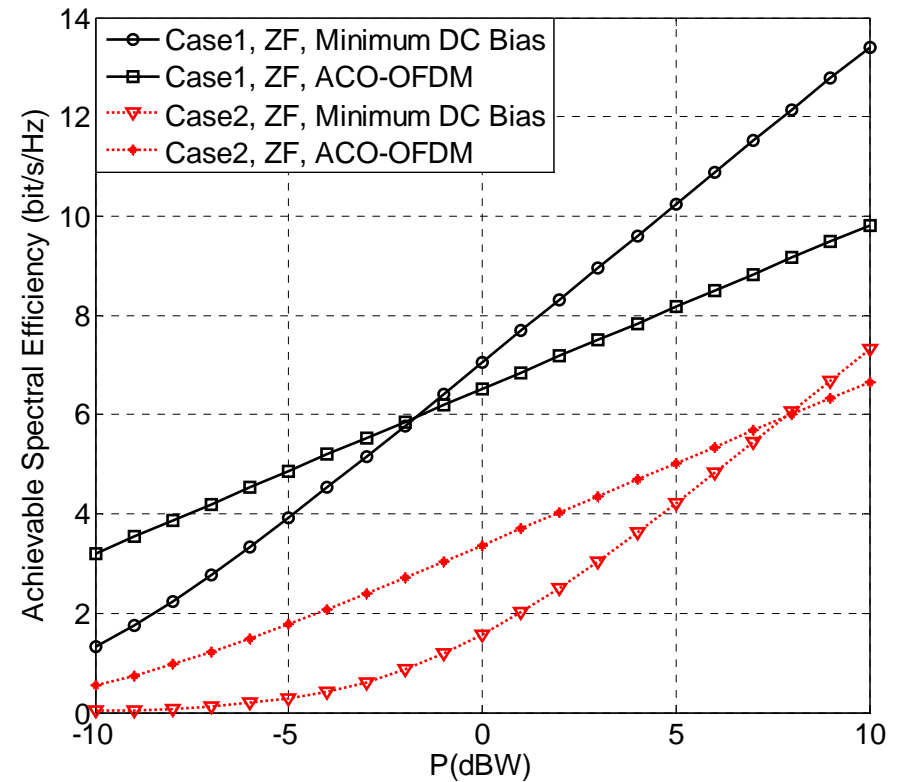


Simulation Results

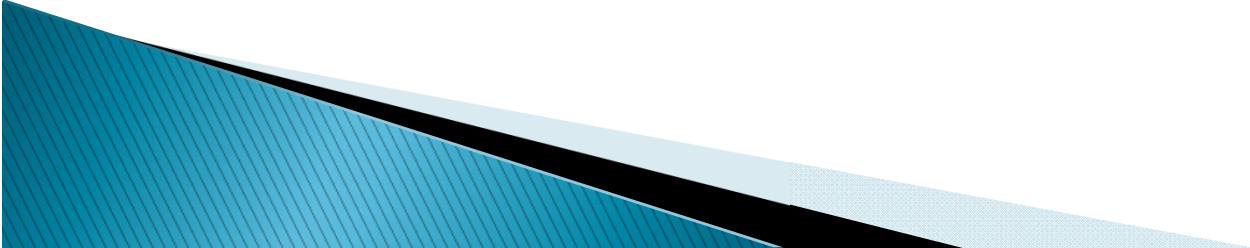
DCO-OFDM-based MU-MIMO VLC



ACO-OFDM-based MU-MIMO VLC



Conclusions

- ▶ MU-MIMO-OFDM is investigated for indoor VLC systems
 - ▶ The distance differences of the multiple transmitter-receiver links are considered
 - ▶ DCO-OFDM with minimum and unified DC bias, and ACO-OFDM are compared in the proposed system
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Thanks for your
suggestions!

