

# Downlink Channel Covariance Estimation in Realistic FDD Massive MIMO Systems

Lorenzo Miretti<sup>†</sup>, Renato L.G. Cavalcante<sup>\*</sup>, and Slawomir Stanczak<sup>\*</sup>

Communication Systems Department, EURECOM<sup>†</sup>

Fraunhofer Heinrich Hertz Institute and Technical University of Berlin<sup>\*</sup>

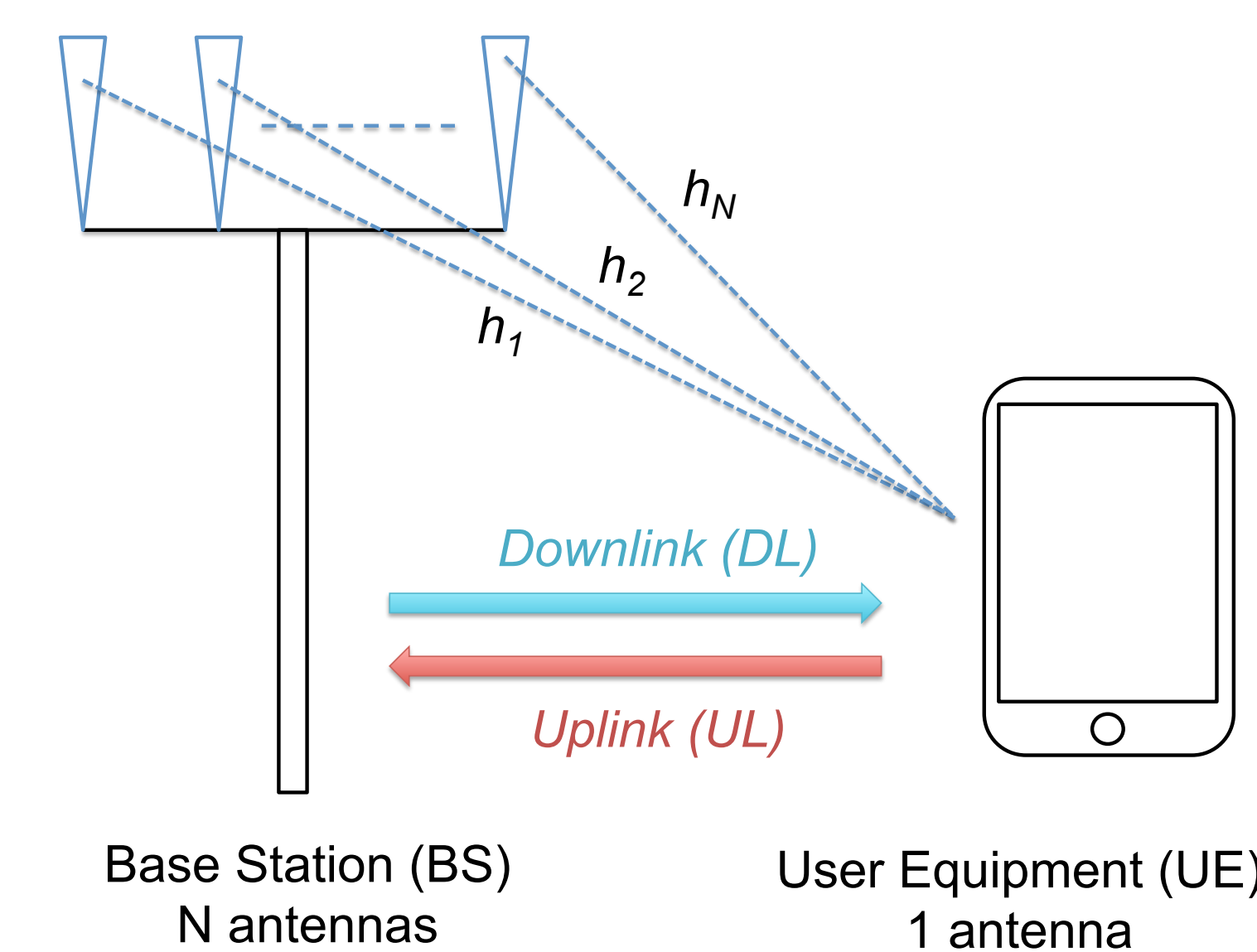


## Spatial Covariance Matrix

- Knowledge of  $\mathbf{R}^d = \mathbb{E}[\mathbf{h}^d(\mathbf{h}^d)^H]$  at the base station:
- Crucial for many DL CSI acquisition and beamforming algorithms in FDD Massive MIMO systems.
- However, **difficult to obtain** in practical FDD Massive MIMO systems.

## DL Covariance Estimation

- Conventional techniques based on DL training and feedback unfeasible in Massive MIMO.
- At the BS,  $\mathbf{R}^u$  is generally easier to estimate.
- In FDD systems,  $\mathbf{R}^d \neq \mathbf{R}^u$ .
- However, a weaker form of **channel reciprocity in the angular domain** can be assumed.



## FDD UL/DL Covariance Conversion Problem

- Estimation of  $\mathbf{R}^d$  from  $\mathbf{R}^u$ .

## Existing Approaches: Issues

- Most of the available solutions rely on simple channel models and/or require specific array geometries (e.g. ULA).
- They often **fail** to address important effects such as propagation in **3D environments**, **polarization**, and **non-ideal array geometries**.

## A Novel Covariance Model for Dual-polarized Arrays

$$\mathbf{R} = \mathbb{E}[\mathbf{h}\mathbf{h}^H] = \int_{\Omega} \rho_V(\boldsymbol{\theta}) \mathbf{a}_V(\boldsymbol{\theta}) \mathbf{a}_V(\boldsymbol{\theta})^H d\boldsymbol{\theta} + \int_{\Omega} \rho_H(\boldsymbol{\theta}) \mathbf{a}_H(\boldsymbol{\theta}) \mathbf{a}_H(\boldsymbol{\theta})^H d\boldsymbol{\theta}$$

- $\Omega = [-\pi, \pi] \times [0, \pi]$  is a spherical coordinate system.
- $\mathbf{a}_V, \mathbf{a}_H : \Omega \rightarrow \mathbb{C}^{N \times 1}$  are the BS antenna array responses for the vertical and for the horizontal polarizations.
- $\rho_V, \rho_H : \Omega \rightarrow \mathbb{R}^+$  are the frequency invariant angular power spectra for the vertical and for the horizontal polarizations (V-APS, H-APS).
- It can be derived from **3GPP-3D-like** channel models, both for **narrow-band** and **wide-band OFDM** systems.

## Covariance Conversion using Projection Methods

$$\mathbf{R}^u = \int_{\Omega} \rho_V(\boldsymbol{\theta}) \mathbf{a}_V^u(\boldsymbol{\theta}) \mathbf{a}_V^u(\boldsymbol{\theta})^H d\boldsymbol{\theta} + \int_{\Omega} \rho_H(\boldsymbol{\theta}) \mathbf{a}_H^u(\boldsymbol{\theta}) \mathbf{a}_H^u(\boldsymbol{\theta})^H d\boldsymbol{\theta} \quad (1)$$

$$\mathbf{R}^d = \int_{\Omega} \rho_V(\boldsymbol{\theta}) \mathbf{a}_V^d(\boldsymbol{\theta}) \mathbf{a}_V^d(\boldsymbol{\theta})^H d\boldsymbol{\theta} + \int_{\Omega} \rho_H(\boldsymbol{\theta}) \mathbf{a}_H^d(\boldsymbol{\theta}) \mathbf{a}_H^d(\boldsymbol{\theta})^H d\boldsymbol{\theta} \quad (2)$$

- We obtain an estimate  $(\hat{\rho}_V, \hat{\rho}_H)$  of  $(\rho_V, \rho_H)$  based on the knowledge of  $\mathbf{R}^u$ , expression (1), and known properties of  $(\rho_V, \rho_H)$ .
- We compute an estimate of  $\mathbf{R}^d$  from (2), and by substituting  $(\rho_V, \rho_H)$  with its estimate  $(\hat{\rho}_V, \hat{\rho}_H)$ .

**Core idea:** Joint V-APS and H-APS estimation formalized as a **convex feasibility problem**: very effective solutions based on **projection methods** on an infinite-dimensional Hilbert space.

## Proposed Covariance Conversion Scheme

- Extension of the ideas in [1] to the considered realistic covariance model.
- Derived by focusing on the Hilbert space  $\mathcal{H} = L^2[\Omega] \times L^2[\Omega]$  equipped with the inner product  $\langle (f_V, f_H), (g_V, g_H) \rangle := \int_{\Omega} f_V(\boldsymbol{\theta}) g_V(\boldsymbol{\theta}) d^2\boldsymbol{\theta} + \int_{\Omega} f_H(\boldsymbol{\theta}) g_H(\boldsymbol{\theta}) d^2\boldsymbol{\theta}$ .
- We estimate  $(\rho_V, \rho_H) \in \mathcal{H}$  by solving  $\text{find } (\rho_V, \rho_H)^* \in V := \cap_{m=1}^M V_m$ , where  $V_m := \{ \langle (\rho_V, \rho_H), (g_{V,m}^u, g_{H,m}^u) \rangle = r_m^u \}$ .
- $r_m^u$  is the  $m$ th element of  $\text{vec}(\{ \Re\{\mathbf{R}^u\} \Im\{\mathbf{R}^u\} \})$ .
- $g_{(\cdot),m}^u$  is the  $m$ th element of  $\text{vec}(\{ \Re\{\mathbf{a}_{(\cdot)}^u(\boldsymbol{\theta}) \mathbf{a}_{(\cdot)}^u(\boldsymbol{\theta})^H\} \Im\{\mathbf{a}_{(\cdot)}^u(\boldsymbol{\theta}) \mathbf{a}_{(\cdot)}^u(\boldsymbol{\theta})^H\} \})$ .
- Among the solutions, we choose the **projection onto the linear variety**  $P_V(0)$  (**Algorithm 1**).
- Closed-form solution available in terms of inner products of the type  $\langle (g_{V,m}^u, g_{H,m}^u), (g_{V,l}^u, g_{H,l}^u) \rangle$ .
- A more accurate but more complex variant (**Algorithm 2**) is available, which takes into account the positivity of the APS.

## Main Advantages

- Algorithm 1** is a simple **matrix/vector multiplication**  $\hat{\mathbf{r}}^d = \mathbf{F} \mathbf{r}^u$  over vectorized covariances.
- $\mathbf{F}$  depends only on the **array geometry** and it is **computed once** for the entire system lifetime.
- It takes into account **polarization** and **3D propagation**.
- No specific **array geometry** is required.

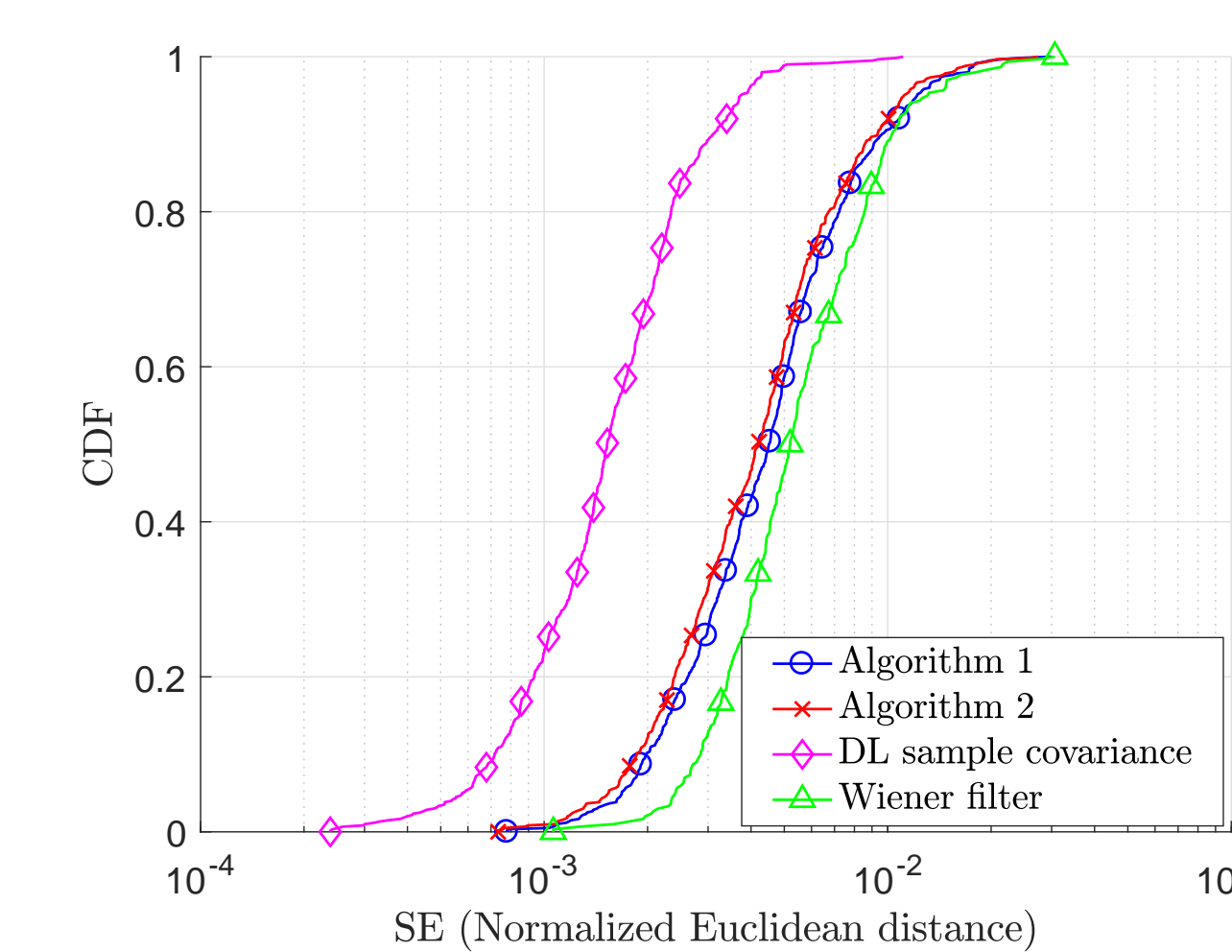


Figure: CDF of the DL covariance estimation error.

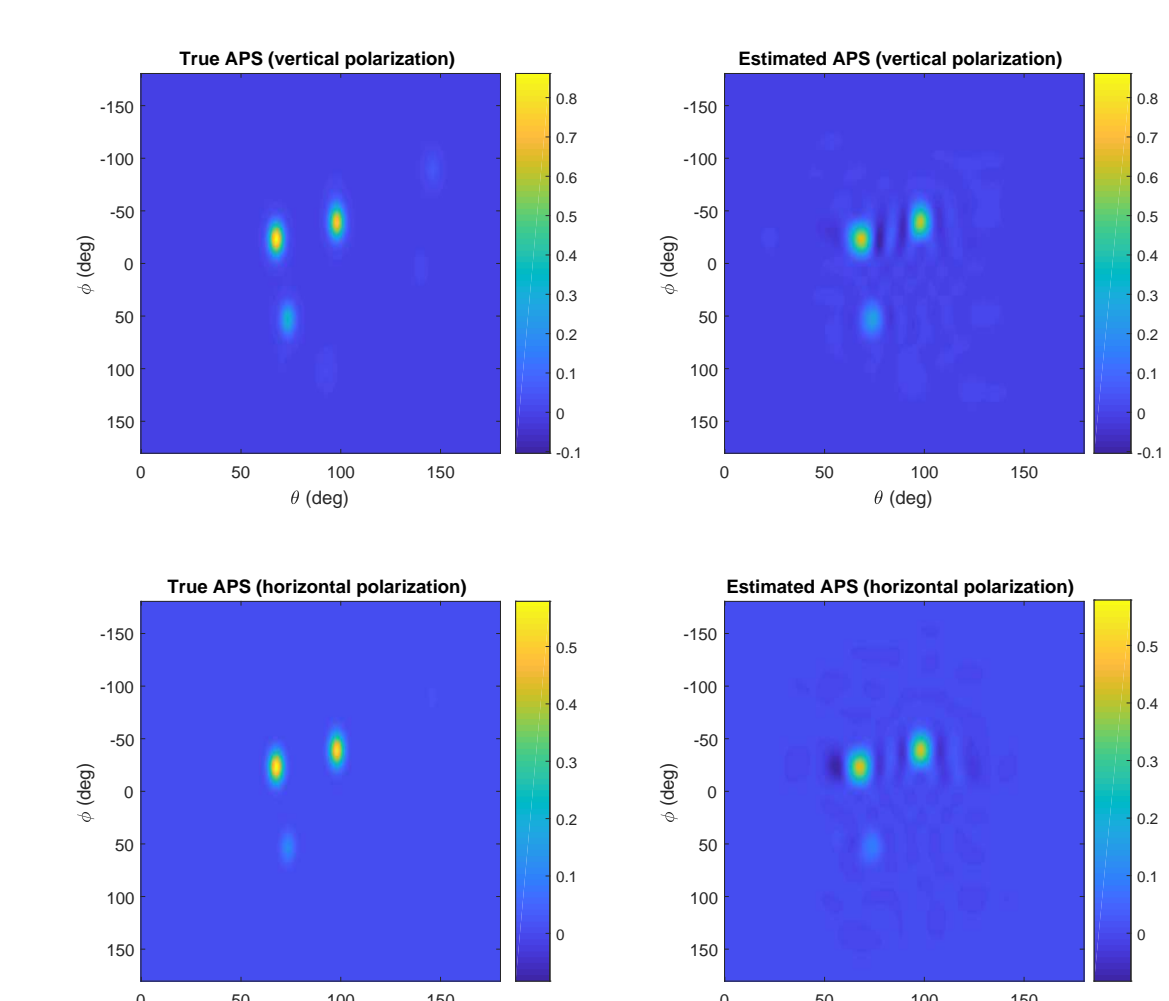


Figure: Example of joint V-APS and H-APS estimation.

[1] L. Miretti, Renato L.G. Cavalcante and Slawomir Stanczak, "FDD massive MIMO channel spatial covariance conversion using projection methods", *IEEE ICASSP*, 2018

[2] L. Miretti, Renato L.G. Cavalcante and Slawomir Stanczak, "Downlink channel spatial covariance estimation in realistic FDD massive MIMO systems", *IEEE GlobalSIP*, 2018