

Generalized Approximate Message Passing for Unlimited Sampling of Sparse Signals



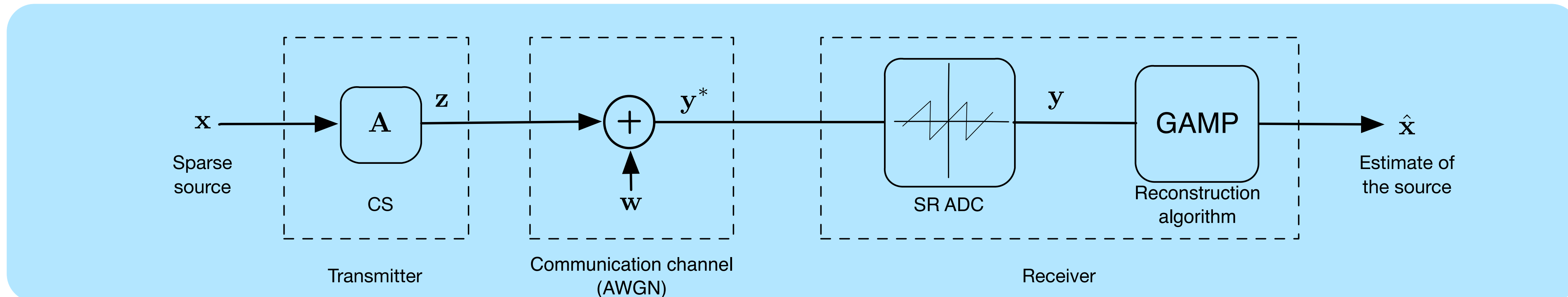
TECHNISCHE
UNIVERSITÄT
WIEN
Vienna University of Technology

Osman Musa^{†*}, Peter Jung[†] and Norbert Goertz^{*}

[†]Communications and Information Theory, Technische Universität Berlin

^{*}Institute of Telecommunications, Technische Universität Wien

{osman.musa,norbert.goertz}@nt.tuwien.ac.at, peter.jung@tu-berlin.de



Introduction

Compressed Sensing

- In Compressed Sensing (CS) we take M linear measurements $\{y_i\}_{i=1}^M$ of an N -dimensional K -sparse (has at most K nonzero components) vector \mathbf{x} , according to

$$\mathbf{y} = \mathbf{A}\mathbf{x} + \mathbf{w}$$

- Recovery is possible if $\mathbf{A} \in \mathbb{R}^{M \times N}$ satisfies the Restricted Isometry Property (RIP)
- Matrices whose elements are randomly drawn from a (sub-)Gaussian distribution satisfy RIP with high probability

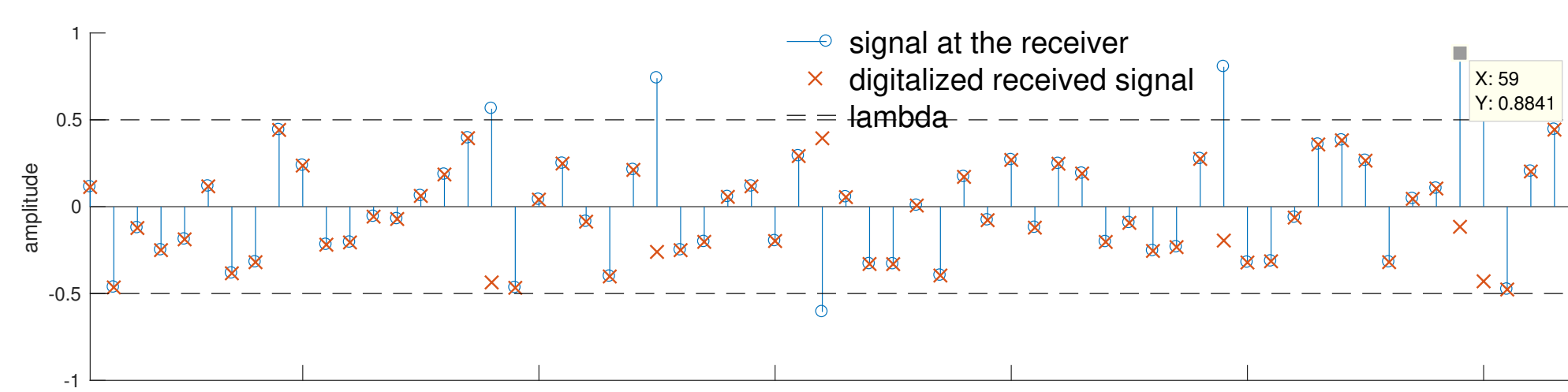
Main Idea and Contributions

- We want to reduce the effects of finite dynamic range
- Image and audio processing, bio-medical applications etc.
- Bhandari et al. propose digitalizing bandlimited signals with a self-reset (SR) analog to digital converter (ADC) defined by

$$\mathcal{M}_\lambda(t) = 2\lambda \left(\left\lfloor \frac{t}{2\lambda} + \frac{1}{2} \right\rfloor - \frac{1}{2} \right)$$

where $\lfloor t \rfloor \triangleq t - [t]$ is the remainder of the division t and λ

- Perfect recovery of a bandlimited signal from its discrete samples is possible if the sampling period $T \leq (2\pi e)^{-1}$
- A. Bhandari et al. provide sufficient conditions for perfect recovery of a K -sparse signal from its low-pass filtered version
- We take CS measurements and digitalize them with a SR ADC
- The main contributions:
 - Consider a new way of digitalizing CS measurements
 - Apply the known GAMP framework
 - Provide closed-form expressions for the nonlinear steps



Unlimited sampling of AWGN corrupted CS measurements

Bernoulli-Gaussian Mixture Prior We assume an i.i.d. source vector where each component x_j of \mathbf{x} is a realization of a Bernoulli-Gaussian distributed random variable

$$p_{x_j}(x_j) = (1 - \epsilon)\delta(x_j) + \epsilon \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{1}{2\sigma^2}x_j^2},$$

with ϵ being the probability of a non-zero value and σ^2 being the variance of the zero-mean Gaussian distribution

Measurement Model

- We obtain M noisy CS measurement according to

$$\mathbf{y} = \mathcal{M}_\lambda(\mathbf{A}\mathbf{x} + \mathbf{w}).$$

- We assume i.i.d. noise vector \mathbf{w} where each $w_i \sim \mathcal{N}(0, \sigma_w^2)$

GAMP for Unlimited sampling

Why GAMP?

- GAMP is very appealing for its efficiency and accurate recovery
- It approximates the computationally intractable high-dimensional integration involved with calculating

$$\hat{\mathbf{x}} \approx \mathbb{E}\{\mathbf{x} | \mathbf{y}\}$$

- It allows to model the quantization as a probabilistic channel with unquantized input and quantized output
- It allows to incorporate measurement noise in the model

The Steps of GAMP

- At $t = 0$, the algorithm is initialized according to (the far right values correspond to the Bernoulli-Gaussian mixture prior)

$$\hat{\mathbf{x}}^0 = \mathbb{E}\{\mathbf{x}\} = 0, \quad \mathbf{v}_x^0 = \text{var}\{\mathbf{x}\} = (1 - \gamma)\sigma^2, \quad \hat{\mathbf{s}}^0 = \mathbf{0}_{M \times 1}$$

- At every iteration $t = 1, 2, \dots$ compute the measurement and estimation updates

$$\mathbf{v}_p^t = (\mathbf{A} \bullet \mathbf{A}) \mathbf{v}_x^{t-1}, \quad \mathbf{v}_r^t = ((\mathbf{A} \bullet \mathbf{A})^T \mathbf{v}_s^t)^{-1}$$

$$\hat{\mathbf{p}}^t = \mathbf{A} \hat{\mathbf{x}}^{t-1} - \mathbf{v}_p^t \bullet \hat{\mathbf{s}}^{t-1}, \quad \hat{\mathbf{r}}^t = \hat{\mathbf{x}}^{t-1} + \mathbf{v}_r^t \bullet (\mathbf{A}^T \hat{\mathbf{s}}^t)$$

$$\hat{\mathbf{s}}^t = F_1(\mathbf{y}, \hat{\mathbf{p}}^t, \mathbf{v}_p^t), \quad \hat{\mathbf{x}}^t = G_1(\hat{\mathbf{r}}^t, \mathbf{v}_r^t; p_x)$$

$$\mathbf{v}_s^t = F_2(\mathbf{y}, \hat{\mathbf{p}}^t, \mathbf{v}_p^t), \quad \mathbf{v}_x^t = G_2(\hat{\mathbf{r}}^t, \mathbf{v}_r^t; p_x)$$

- The functions $F_1(\cdot)$, $F_2(\cdot)$, $G_1(\cdot)$ and $G_2(\cdot)$ are applied component-wise and are given by

$$F_1(y, \hat{p}, v_p) = \frac{\mathbb{E}\{z | y\} - \hat{p}}{v_p}, \quad G_1(\hat{r}, v_r; p_x) = \mathbb{E}\{x | \hat{r}\}$$

$$F_2(y, \hat{p}, v_p) = \frac{v_p - \text{var}\{z | y\}}{v_p^2}, \quad G_2(\hat{r}, v_r; p_x) = \text{var}\{x | \hat{r}\}$$

- The first and the second moment of $z|y$ and $x|\hat{r}$ are evaluated with respect to

$$p_{z|y} \propto p_{y|z}(\cdot | \cdot) p_z(\cdot) \quad \text{and} \quad p_{x|\hat{r}} \propto g(\cdot; \hat{r}, v_r) p_x(\cdot)$$

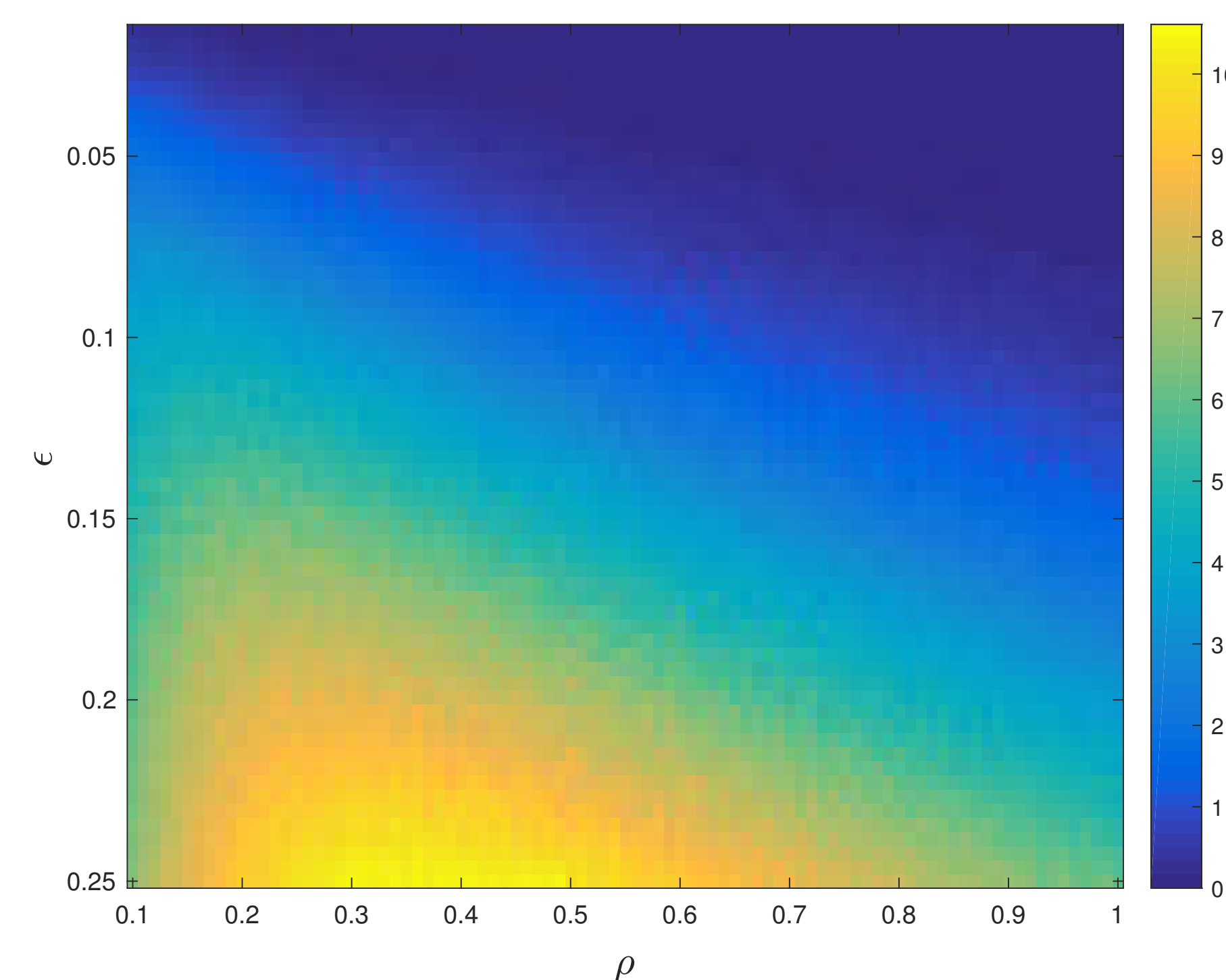
where $z \sim \mathcal{N}(\hat{p}, v_p)$

- Stop iterating if $\|\hat{\mathbf{x}}^t - \hat{\mathbf{x}}^{t-1}\|_2 < \epsilon \|\hat{\mathbf{x}}^t\|_2$ with a small $\epsilon > 0$ (e.g. $\epsilon = 10^{-2}$) or when $t \geq t_{\max}$
- GAMP nonlinear operators are computed for the specific input and output channels and given in the paper

Numerical Results

Simulation Setup

- We averaged our results over 4000 independent realizations of the source vector \mathbf{x} , the sensing matrix \mathbf{A} and the AWGN \mathbf{w}
 - In each simulation we fix $N = 256$, and acquire $n = \rho N$ measurements of the $K = \epsilon N$ sparse vector
 - Each CS measurement vector is corrupted with AWGN noise with power $\sigma_w^2 = 10^{-\text{SNR}/10}$, where the SNR is defined as
- $$\text{SNR/dB} = 10 \log_{10} \{ \|\mathbf{y}^*\|^2 / \|\mathbf{w}\|^2 \}$$
- In the noiseless case $\text{SNR} = \infty$. The SR ADC threshold $\lambda = 1$
 - Successful recovery := $\text{MSE} \leq -30\text{dB}$



Results

- Noiseless Case: There's a clear phase transition (PT) between unsuccessful (blue) and successful (yellow) regions
- Classical CS algorithms completely fail when $\|\epsilon_g\|_0 \neq 0$
- GAMP can cope with distortion; the PT is almost linear in ϵ
- Noisy Case: The PT curve shifted to the right lower corner
- More measurements are needed when the CS measurements are corrupted with AWGN (SNR = 20dB) before digitalization

Conclusion

- For certain choice of the signal parameters, the GAMP is able to successfully recover a sparse signal from folded measurements
- Unlike the classical algorithms for recovery of sparse signals from folded measurements, the GAMP algorithm can cope with the noise introduced by a communication channel

