## Blind Digital Modulation Classification based on $M^{th}$ -Power Nonlinear Transformation

Vincent Gouldieff<sup>1,2</sup>, Jacques Palicot<sup>1</sup>, Steredenn Daumont<sup>2</sup>

<sup>1</sup>CentraleSupelec/IETR, Rennes, France <sup>2</sup>Zodiac Data Systems/Zodiac Aerospace, Caen, France





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### **Motivation**



#### Cognitive Radio Context



#### Cognitive Node

- Senses its environment
- Decision:  $C \in \mathcal{C}$

#### Cognitive Terminal

- **Blind** estimation of C
- No prior knowledge
- Low complexity

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# Is the " $M^{th}$ -Power Nonlinear Transformation" a good candidate for the estimation of C?





- System Model & Assumptions
- Basics on  $M^{th}$ -Power Transform (MPT)
- Computation of the References
- Performance and Complexity
- Conclusion and Perspectives





- AMC: a short Review
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### Automatic Modulation Classification [1]

<sup>[1]</sup> O. A. Dobre et al., Survey of AMC Techniques: Classical Approaches and New Trends, IET Communications, 2007





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<sup>[4]</sup> A. Swami and B. M. Sadler, Hierarchical Digital Modulation Classification using Cumulants, IEEE Trans. Commun., 2000

<sup>[5]</sup> O. A. Dobre et al., Cyclostationarity-based Blind Classification of Analog and Digital Modulations, in Proc. MILCOM, 2006





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<sup>[6]</sup> C. W Lim and M. B. Wakin, Automatic Modulation Recognition for Spectrum Sensing using Nonuniform Compressive Samples, in Proc. ICC, 2012





#### System Model & Assumptions

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Single-carrier digital signal in AWGN

$$x(n) = a \cdot e^{i \cdot (2\pi f_r \cdot n + \phi)} \cdot \sum_k s(k) \cdot h(nT_e - kT - \tau) + \omega(n)$$

No synchronization/demodulation

No prior knowledge







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Based on 1PT (classical PSD):

- Normalization step: y(n) has unit useful power (a = 1)
- Spectral centering:  $f_r pprox 0$
- Output parameters:  $\widehat{\eta}=\{\widehat{\sigma}^2_{\omega},\widehat{
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■ *M*PT function:

$$MPT_{y}(f) = \left| \frac{1}{N_{s}} \cdot \sum_{n=0}^{N_{e}-1} y^{M}(n) \cdot e^{-2i\pi nf} \right|^{\frac{2}{M}} = \sqrt[M]{\Gamma_{n}[y^{M}](f)}$$
$$CM_{y}^{n,p,\tau}(f) = \frac{1}{N_{s}} \cdot \sum_{n=0}^{N_{e}-1} \prod_{i=1}^{n} y^{(*)_{i}}(n+\tau_{i}) \cdot e^{-2i\pi nf}$$

Behavior:

#### → Minimum Distance Classification

Basics on MPT



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→ Minimum Distance Classification

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Basics on MPT





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$$\sum_{n=-\infty}^{+\infty} g(t+na) = \frac{1}{a} \sum_{m=-\infty}^{+\infty} G\left(\frac{m}{a}\right) \cdot e^{i2\pi f \frac{m}{a}t}$$

• Theory for 
$$M = 2$$
:  
2F

$$2\mathsf{PT}_{th}^C = \left| \mathbb{E}[s^2] H_2(0) \right|$$

• Theory for M = 4:

$$4\mathsf{PT}_{th}^{C} = \left| \frac{1}{T} \left( \mathbb{E}[s^{4}] H_{4}(0) + 6 \left( \mathbb{E}[s^{2}] \right)^{2} \sum_{k>0} H_{22}^{(k)}(0) \right) \right|^{\frac{1}{2}}$$

**Distributions** for accurate references (or corrective terms)

<sup>[7]</sup> J. E. Mazo, Jitter Comparison of Tones Generated by Squaring and by Fourth-Power Circuits, Bell System Journal, 1978





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Theory for M=2:  $2\mathsf{PT}^C_{th} = \left|\mathbb{E}[s^2]H_2(0)\right|$ 

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#### Theory and Simulation results

Distance Matrix

		Constellation								
		BPSK	QPSK	8PSK	8AMPM	R8QAM	C8QAM	16QAM		
Constellation	BPSK	0	1.004	1.419	0.905	0.564	0.988	1.055		
	QPSK		0	0.761	0.219	0.656	0.068	0.159		
	8PSK			0	0.639	0.861	0.829	0.602		
	8AMPM				0	0.468	0.268	0.161		
	R8QAM					0	0.677	0.629		
	C8QAM						0	0.227		
	16QAM							0		

- PSK/QAM/PAM 25% During Control Cont
- 35% Root-Raised-Cosine



1024 symbols
 ρ = 2
 35% Root-Raised-Cosine



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Correct Classification Rate (CCR)



■ 35% Root-Raised-Cosine

PSK/QAM/PAM35% Root-Raised-Cosine



Comparison between MPT and Cumulant-based classification



Advantages

- No need for pre-demodulation
- More robustness to noise/uncertainty
- Low complexity (FFTs)



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#### This work:

- *M*PT = great tool for AMC (but not only!)
- Basic theory
- Blind performance

#### Future work:

- Distributions & theoretical CCR
- Other contexts (channel, interference,...)



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## Thank you!

