



# Performance Analysis for Pilot-based 1-bit Channel Estimation with Unknown Quantization Threshold



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## Introduction (I)

### Considered Problem:

- Channel parameter estimation with a low-complexity receiver performing 1-bit ADC
- Goal: Characterization of performance loss associated with hard-limiting under an unknown threshold
- Approach: Derivation of performance bounds for parameter estimation with nuisance parameter (offset)

## Motivation (II)

### Statistical Signal Processing with 1-bit ADC:

- Advantage of 1-bit ADC: simple, cheap and robust analog radio-front end, fast binary DSP
- 1-bit ADC allows high temporal sampling rates → high receive bandwidth
- In practice circuit imperfections and external effects will result in asymmetric 1-bit ADCs

## System Model - Ideal and 1-bit System (III)

### Receive Signal - Ideal Receiver ( $\infty$ -bit ADC)

$$\mathbf{y} = \zeta \mathbf{x} + \boldsymbol{\eta} \quad \text{where} \quad \mathbf{E}_{\boldsymbol{\eta}}[\boldsymbol{\eta}\boldsymbol{\eta}^T] = \mathbf{I}_N$$

### Transmit Signal (BPSK, Known Structure)

$$\mathbf{x} \in \{-1, 1\}^N \quad \text{with} \quad \frac{1}{N} \sum_{n=1}^N x_n = 0$$

### Receive Signal - Low-Complexity Receiver (1-bit ADC)

$$\mathbf{r} = \text{sign}(\mathbf{y} - \alpha \mathbf{1}) = \text{sign}(\zeta \mathbf{x} + \boldsymbol{\eta} - \alpha \mathbf{1})$$

### 1-bit ADC - Hard-limiter Model

$$[\text{sign}(\mathbf{x})]_n = \begin{cases} +1 & \text{if } x_n \geq 0 \\ -1 & \text{if } x_n < 0 \end{cases}, \quad n = 1, \dots, N$$

## Deterministic and Hybrid Approach (IV)

### Deterministic - Model Assumptions

$$\zeta \rightarrow \text{deterministic unknown} \\ \alpha \rightarrow \text{deterministic unknown}$$

### Hybrid - Model Assumptions

$$\zeta \rightarrow \text{random unknown} \\ \alpha \rightarrow \text{deterministic unknown}$$

### Deterministic - Estimator (MLE)

$$\hat{\zeta}_{\mathbf{y}}(\mathbf{y}) \triangleq \arg \max_{\zeta \in \mathbb{R}} p_{\mathbf{y}}(\mathbf{y}; \zeta) \\ [\hat{\zeta}_{\mathbf{r}}(\mathbf{r}) \hat{\alpha}_{\mathbf{r}}(\mathbf{r})]^T \triangleq \arg \max_{\zeta, \alpha \in \mathbb{R}} p_{\mathbf{r}}(\mathbf{r}; \zeta, \alpha)$$

### Hybrid - Estimator (MAP / JMAP-MLE)

$$\hat{\zeta}_{\mathbf{y}}(\mathbf{y}) \triangleq \arg \max_{\zeta \in \mathbb{R}} p_{\mathbf{y}, \zeta}(\mathbf{y}, \zeta) \\ [\hat{\zeta}_{\mathbf{r}}(\mathbf{r}) \hat{\alpha}_{\mathbf{r}}(\mathbf{r})]^T \triangleq \arg \max_{\zeta, \alpha \in \mathbb{R}} p_{\mathbf{r}, \zeta}(\mathbf{r}, \zeta; \alpha)$$

### Deterministic - Performance (MSE)

$$\text{MSE}_{\mathbf{y}}(\zeta) \triangleq \mathbf{E}_{\mathbf{y}; \zeta} [(\hat{\zeta}_{\mathbf{y}}(\mathbf{y}) - \zeta)^2] \\ \text{MSE}_{\mathbf{r}}(\zeta, \alpha) \triangleq \mathbf{E}_{\mathbf{r}; \zeta, \alpha} [(\hat{\zeta}_{\mathbf{r}}(\mathbf{r}) - \zeta)^2]$$

### Hybrid - Performance (MSE)

$$\text{MSE}_{\mathbf{y}} \triangleq \mathbf{E}_{\mathbf{y}, \zeta} [(\hat{\zeta}_{\mathbf{y}}(\mathbf{y}) - \zeta)^2] \\ \text{MSE}_{\mathbf{r}}(\alpha) \triangleq \mathbf{E}_{\mathbf{r}, \zeta; \alpha} [(\hat{\zeta}_{\mathbf{r}}(\mathbf{r}) - \zeta)^2]$$

## Deterministic Approach - Hard-limiting Loss (V)

### Cramér-Rao Lower Bound - Ideal System with $\infty$ -bit ADC

$$\text{MSE}_{\mathbf{y}}(\zeta) \triangleq F_{\mathbf{y}}^{-1}(\zeta) = \frac{1}{N}$$

### Cramér-Rao Lower Bound - Low-Complexity System with 1-bit ADC

$$\text{MSE}_{\mathbf{r}}^*(\zeta, \alpha) \triangleq F_{\mathbf{r}, \zeta \zeta}^{-1}(\zeta, \alpha) \\ \text{MSE}_{\mathbf{r}}(\zeta, \alpha) \triangleq \frac{F_{r, \alpha \alpha}(\zeta, \alpha)}{F_{r, \zeta \zeta}(\zeta, \alpha) F_{r, \alpha \alpha}(\zeta, \alpha) - F_{r, \zeta \alpha}^2(\zeta, \alpha)} \\ F_{r, \zeta \zeta}(\zeta, \alpha) = \mathbf{E}_{\mathbf{r}; \zeta, \alpha} \left[ \left( \frac{\partial \ln p_{\mathbf{r}}(\mathbf{r}; \zeta, \alpha)}{\partial \zeta} \right)^2 \right] = \frac{N}{2} (\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha)) \\ F_{r, \alpha \alpha}(\zeta, \alpha) = \mathbf{E}_{\mathbf{r}; \zeta, \alpha} \left[ \left( \frac{\partial \ln p_{\mathbf{r}}(\mathbf{r}; \zeta, \alpha)}{\partial \alpha} \right)^2 \right] = \frac{N}{2} (\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha)) \\ F_{r, \zeta \alpha}(\zeta, \alpha) = \mathbf{E}_{\mathbf{r}; \zeta, \alpha} \left[ \frac{\partial \ln p_{\mathbf{r}}(\mathbf{r}; \zeta, \alpha)}{\partial \zeta} \frac{\partial \ln p_{\mathbf{r}}(\mathbf{r}; \zeta, \alpha)}{\partial \alpha} \right] = \frac{N}{2} (\phi_+(\zeta, \alpha) - \phi_-(\zeta, \alpha)) \\ \phi_{\pm}(\zeta, \alpha) \triangleq \frac{\exp(-(\alpha \pm \zeta)^2)}{2\pi(Q(\alpha \pm \zeta) - Q^2(\alpha \pm \zeta))}$$

### Hard-limiting Loss for the Deterministic Approach

$$\chi(\zeta, \alpha) \triangleq \frac{\text{MSE}_{\mathbf{y}}(\zeta)}{\text{MSE}_{\mathbf{r}}(\zeta, \alpha)} = 2 \frac{\phi_+(\zeta, \alpha) \phi_-(\zeta, \alpha)}{\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha)} \\ \chi^*(\zeta, \alpha) \triangleq \frac{\text{MSE}_{\mathbf{y}}(\zeta)}{\text{MSE}_{\mathbf{r}}^*(\zeta, \alpha)} = \frac{1}{2} (\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha))$$

## Hybrid Approach - Hard-limiting Loss (VI)

### Expected Cramér-Rao Lower Bound - Ideal System with $\infty$ -bit ADC

$$\text{MSE}_{\mathbf{y}} \triangleq \mathbf{E}_{\zeta} [F_{\mathbf{y}}^{-1}(\zeta)] = \frac{1}{N}$$

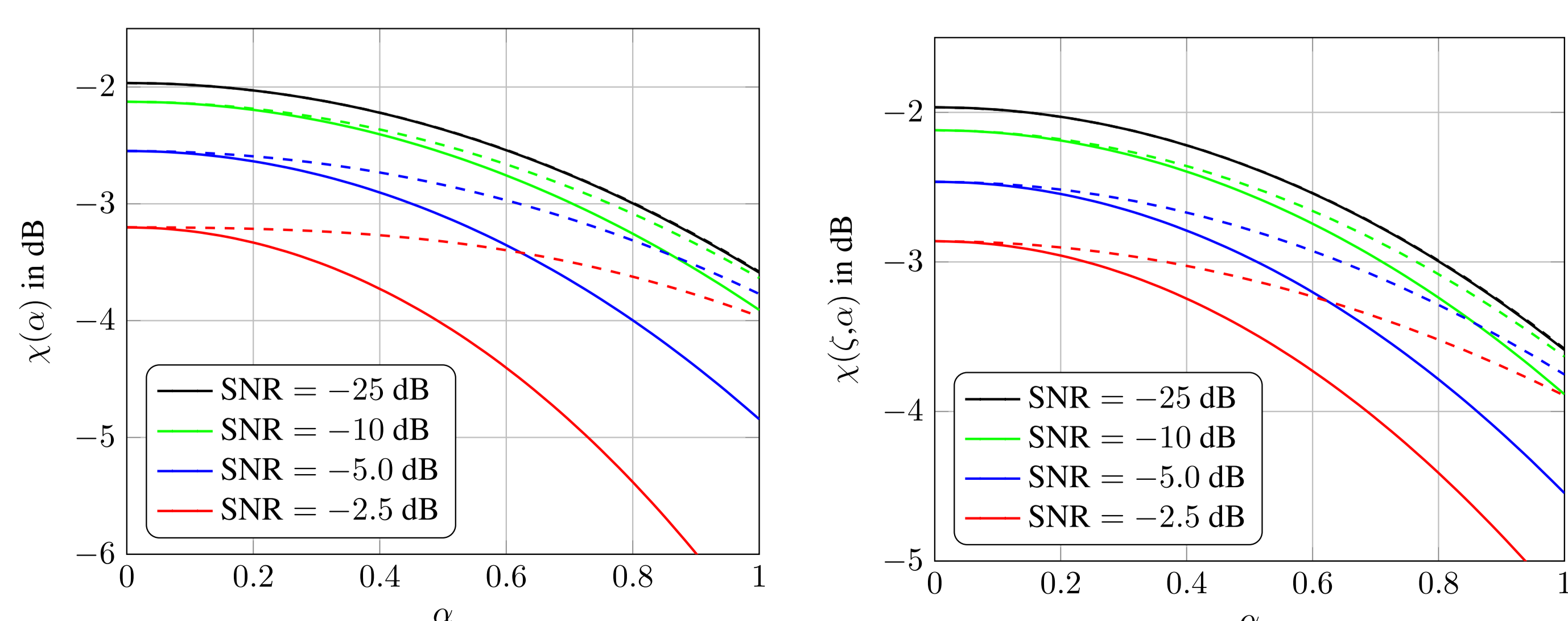
### Expected Cramér-Rao Lower Bound - Low-Complexity System with 1-bit ADC

$$\text{MSE}_{\mathbf{r}}^*(\alpha) \triangleq \mathbf{E}_{\zeta} [F_{\mathbf{r}, \zeta \zeta}^{-1}(\zeta, \alpha)] \\ \text{MSE}_{\mathbf{r}}(\alpha) \triangleq \mathbf{E}_{\zeta} \left[ \frac{F_{r, \alpha \alpha}(\zeta, \alpha)}{F_{r, \zeta \zeta}(\zeta, \alpha) F_{r, \alpha \alpha}(\zeta, \alpha) - F_{r, \zeta \alpha}^2(\zeta, \alpha)} \right] \\ = \mathbf{E}_{\zeta} \left[ \frac{\frac{2}{N} (\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha))}{(\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha))^2 - (\phi_+(\zeta, \alpha) - \phi_-(\zeta, \alpha))^2} \right] \\ = \frac{1}{2N} \left( \mathbf{E}_{\zeta} \left[ \frac{1}{\phi_-(\zeta, \alpha)} \right] + \mathbf{E}_{\zeta} \left[ \frac{1}{\phi_+(\zeta, \alpha)} \right] \right) \\ = \frac{1}{N} \Psi_H \\ \Psi_H \triangleq \mathbf{E}_{\zeta} \left[ \frac{1}{\phi_-(\zeta, \alpha)} \right] = \mathbf{E}_{\zeta} \left[ \frac{1}{\phi_+(\zeta, \alpha)} \right]$$

### Hard-limiting Loss for the Hybrid Approach

$$\chi(\alpha) \triangleq \frac{\text{MSE}_{\mathbf{y}}}{\text{MSE}_{\mathbf{r}}(\alpha)} \stackrel{a}{=} \Psi_H^{-1} \\ \chi^*(\alpha) \triangleq \frac{\text{MSE}_{\mathbf{y}}}{\text{MSE}_{\mathbf{r}}^*(\alpha)} \stackrel{a}{=} \frac{1}{2 \mathbf{E}_{\zeta} \left[ \frac{1}{\phi_+(\zeta, \alpha) + \phi_-(\zeta, \alpha)} \right]}$$

## Performance Analysis - Results (VII)



## Summary and Insights (VIII)

- Characterization of channel parameter estimation performance under 1-bit ADC with unknown threshold
- Analysis of two different estimation approaches (deterministic/hybrid)
- Performance analysis through performance bounds for MLE and MAP/JMAP-ML
- Low SNR regime: additional performance loss (offset estimation) vanishes
- Medium SNR: loss vanishes for small offset values (requires careful 1-bit ADC hardware design)

## Further Work (IX)

- Analytic characterization of the performance loss in the low SNR regime
- Advanced channel estimation problems (e.g. ISI wireless channel)