



On Optimal Sensing and Capacity Trade-off in Cognitive Radio Systems with Directional Antennas

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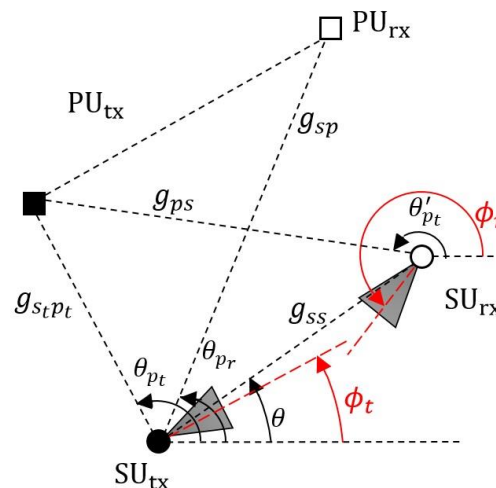
Outline

- ✓ **System Model**
- ✓ **Spectrum Sensing**
- ✓ **Data Communication Channel**
- ✓ **Sensing-Capacity Trade-off**
- ✓ **Constrained Optimization Problem**
- ✓ **Solution**
- ✓ **Simulation Results**

System Model

Geometry

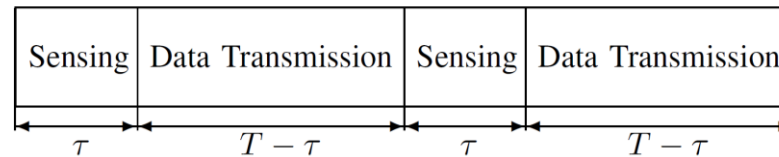
- Secondary users (SUs) and primary users (PUs) coexist.
- SUs are equipped with steerable directional antennas.
- The directional antennas can identify and enable transmission and reception across spatial domain and enhance spectrum utilization, compared with omni-directional antennas.
- SU_{tx} first senses the spectrum, and, then transmits data to SU_{rx} with power P if spectrum is sensed idle.
- SU_{tx} knows the geometry of CR network.
- SU_{tx} knows only the CSI of SU_{tx} - SU_{rx} link, and the statistics of the other links.
- θ , θ_{pr} and θ_{pt} are the orientations of SU_{rx} , PU_{rx} and PU_{tx} w.r.t. SU_{tx} .
- ϕ_t and ϕ_r are the boresight of SU_{tx} and SU_{rx} antennas (to be optimized).



System Model

Frame Structure of SUs

- SU_{tx} employs a frame with duration T seconds.
- SU_{tx} senses the spectrum for a duration of τ seconds, to decide whether it is busy or idle.
- The remaining frame of duration $T - \tau$ seconds is used for data transmission if the channel is sensed idle.



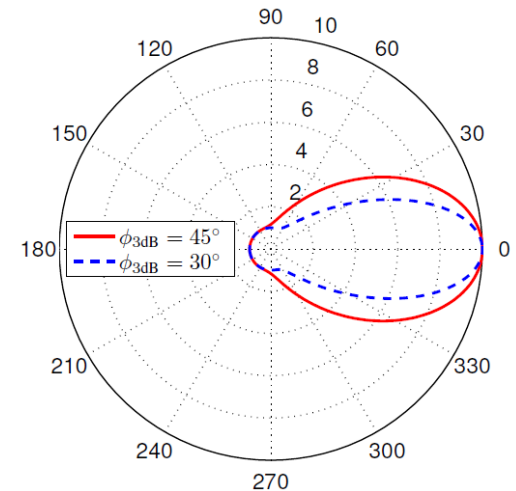
Antenna Model

Gaussian Pattern

$$A(\phi) = A_1 + A_0 \exp\left(-B \left(\frac{\phi}{\phi_{3dB}}\right)^2\right)$$

ϕ_{3dB} is the half-power beam-width

A_1 and A_0 are two constant parameters



Spectrum Sensing

- We formulate the spectrum sensing at SU_{tx} as a binary hypothesis testing problem.

➤ The Binary Hypothesis Testing Problem

$$\begin{cases} \mathcal{H}_0: r[k] = w[k] & P(\mathcal{H}_0) = \pi_0 \\ \mathcal{H}_1: r[k] = \sqrt{g_{s_t p_t} A(\phi_t - \theta_{p_t}) L_{s_t p_t}} p[k] + w[k] & P(\mathcal{H}_1) = \pi_1 \end{cases}$$

- $E(p^2) = P_p$
- Considering energy detection, the decision rule at SU_{tx} is $Z = \frac{1}{N_s} \sum_{k=1}^{N_s} |r[k]|^2 \underset{\hat{\mathcal{H}}_0}{\overset{\hat{\mathcal{H}}_1}{\geq}} \xi$

$$P_f(\phi, \tau) = \Pr\{\hat{\mathcal{H}}_1 | \mathcal{H}_0\} = Q\left(\left(\frac{\xi}{\sigma_n^2} - 1\right) \sqrt{\tau f_s}\right)$$

$$P_d(\phi, \tau) = \Pr\{\hat{\mathcal{H}}_1 | \mathcal{H}_1\} = Q\left(\left(\frac{\xi}{\sigma_n^2} - \gamma - 1\right) \sqrt{\frac{\tau f_s}{2\gamma + 1}}\right)$$

- $\hat{\mathcal{H}}_1$ and $\hat{\mathcal{H}}_0$ with probabilities $\hat{\pi}_1$ and $\hat{\pi}_0$ denote that the result of spectrum sensing is busy and idle.

Data Communication Channel

- When the spectrum is sensed idle, the SU_{tx} uses power P to transmit signal to SU_{rx} .

$$y[m] = \sqrt{g_{ss}L_{ss}G(\theta, \phi_t, \phi_r)} s[m] + n[m]$$

$$G(\theta, \phi_t, \phi_r) = A(\phi_t - \theta)A(\phi_r - \pi - \theta) \quad s \sim \mathcal{N}(0, P) \quad n \sim \mathcal{N}(0, \sigma_n^2)$$

➤ Ergodic Capacity

- Spectrum sensing is imperfect and the ergodic capacity would depend on the true status of the PU and the spectrum sensing result.
- The false alarm and detection probabilities should be incorporated in the design and performance analysis.

$$C = D E\{\alpha_0 c_{0,0} + \beta_0 c_{1,0}\}$$

$$\alpha_0 = \Pr\{\mathcal{H}_0, \hat{\mathcal{H}}_0\} \quad \beta_0 = \Pr\{\mathcal{H}_1, \hat{\mathcal{H}}_0\} \quad D = \frac{T-\tau}{T}$$

$$c_{0,0} = \log_2 \left(1 + \frac{g_{ss}L_{ss}G P}{\sigma_n^2} \right)$$

$$c_{1,0} = \log_2 \left(1 + \frac{g_{ss}L_{ss}G P}{\sigma_n^2 + P_p g_{ps} L_{ps} A(\phi_r - \theta'_{pt})} \right)$$



Sensing-Capacity Trade-off

- If we increase the sensing time τ , the spectrum sensing becomes more accurate. On the other hand, data transmission duration decreases. Therefore, a trade-off exists between the sensing time and the transmission capacity of our CR network.
- To increase probability of detection $P_d(\phi, \tau)$, SU_{tx} 's antenna should be pointed to PU_{tx} 's direction to receive the maximum power. On the other hand, the SU_{tx} 's antenna should be pointed to SU_{rx} 's direction to maximize the transmission capacity. Thus, there is a sensing-capacity trade-off in terms of the SU_{tx} 's antenna orientation.
- There are trade-offs between sensing and capacity in terms of sensing duration τ and SU_{tx} 's antenna orientation.

Constrained Optimization Problem

➤ Outage Interference Probability Constraint

- We define the interference outage probability as the probability that the interference exceeds a maximum threshold I_{pk} be smaller than a maximum value ε .

$$Pr\{D\beta_0 g_{sp} L_{sp} P A(\phi_t - \theta_{pr}) > I_{pk} \mid g_{ss}\} \leq \varepsilon \quad (1)$$

➤ Peak Transmit Power Constraint

$$D \hat{\pi}_0 P \leq P_{pk} \quad (2)$$

➤ Constraints on Angles

$$|\phi_t - \theta| \leq \phi_{3dB} \quad (3a)$$

$$|\phi_r - \pi - \theta| \leq \phi_{3dB} \quad (3b)$$

➤ Optimization Problem

$$\text{Max}_{P, \tau, \phi_t, \phi_r} C = D E\{\alpha_0 c_{0,0} + \beta_0 c_{1,0}\}$$

s. t. : (1), (2) and (3) are satisfied.

Solution

- Taking the first derivative of C with respect to τ , we get

$$\lim_{\tau \rightarrow 0} \frac{\partial C}{\partial \tau} \rightarrow +\infty \quad \lim_{\tau \rightarrow T} \frac{\partial C}{\partial \tau} < 0$$

$$\xi \geq \sigma_n^2(1 + m\gamma) \quad m = \frac{\pi_1}{\pi_1 + \pi_0\sqrt{2\gamma + 1}} < 1$$

- Hence, C has a maximum point with respect to τ within the interval $(0, T)$.
- The capacity is concave with respect to P and ϕ_r . However, in general, it is not concave with respect to ϕ_t and τ .

$$P^{\text{opt}} = \min \left\{ \frac{P_{\text{pk}}}{D\hat{\pi}_0}, \frac{-I_{\text{pk}}}{D\bar{b}_0 \ln(\varepsilon)} \right\} \quad (14)$$

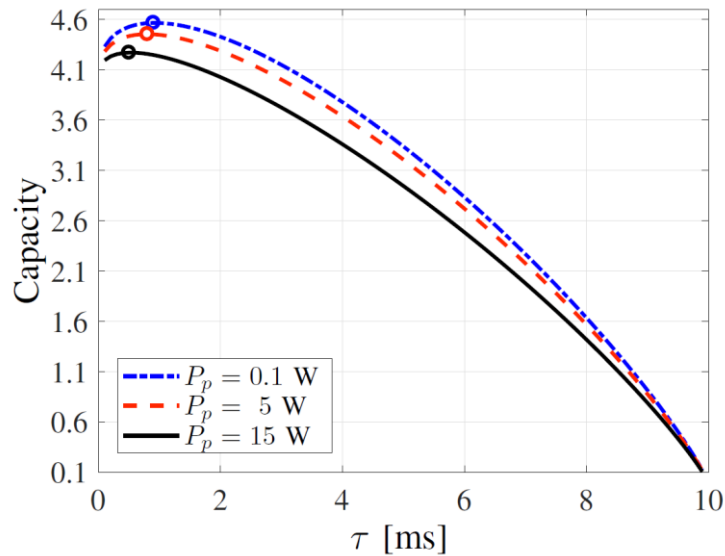
$$\bar{b}_0 = \beta_0 \gamma_{sp} L_{sp} A(\phi_t - \theta_{p_r})$$

Algorithm 1: Optimization Algorithm

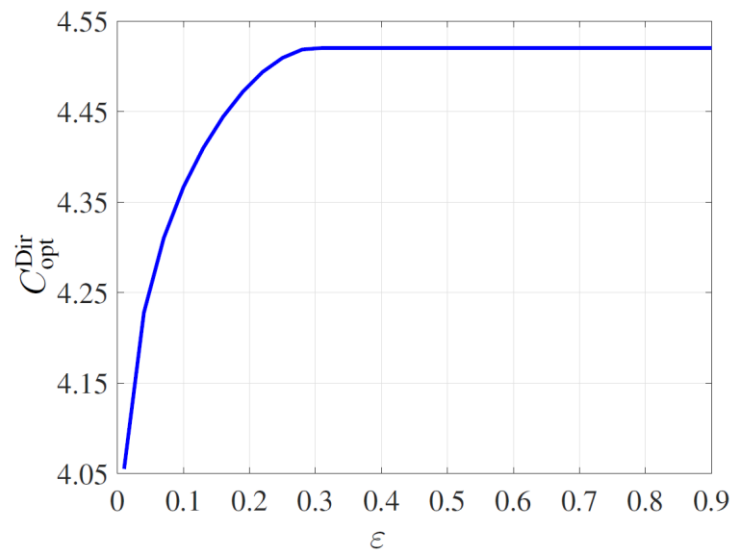
$\phi_t^{(0)} = \phi_{\text{init}}$
 $\tau^{(0)} = \tau_{\text{init}} \in (0, T)$
 calculate P using (14).
 solve $\partial C / \partial \phi_r = 0$ and obtain ϕ_r .
 $[\phi_t^{\text{opt}}, \tau^{\text{opt}}] = \text{argmax} \{C\}$ using bisection search
 $P^{\text{opt}} = [P]_{\phi_t = \phi_t^{\text{opt}}, \tau = \tau^{\text{opt}}}$
 $\phi_r^{\text{opt}} = [\phi_r]_{\phi_t = \phi_t^{\text{opt}}, \tau = \tau^{\text{opt}}$

Simulation Results

$$\pi_1 = 0.3, T = 10 \text{ ms}, f_s = 20 \text{ KHz},$$



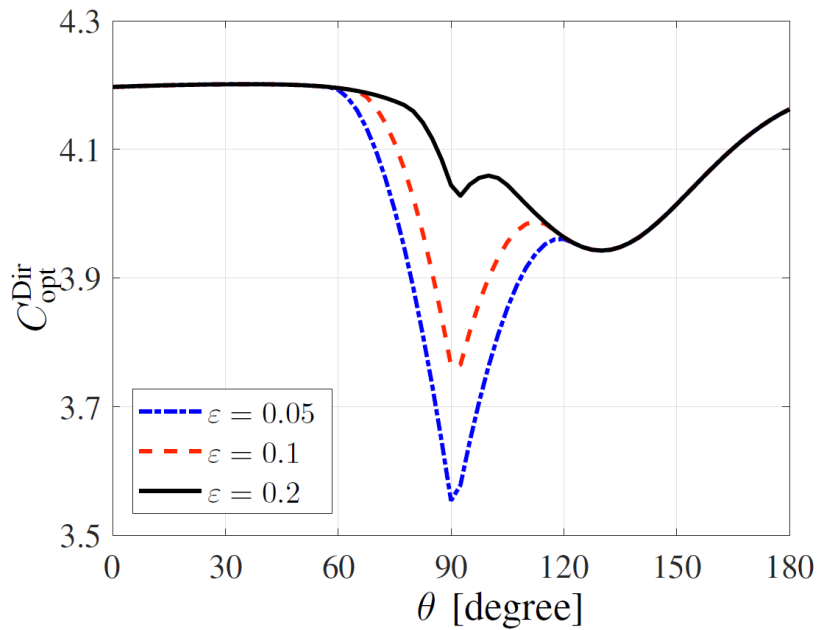
Variations of C versus τ .



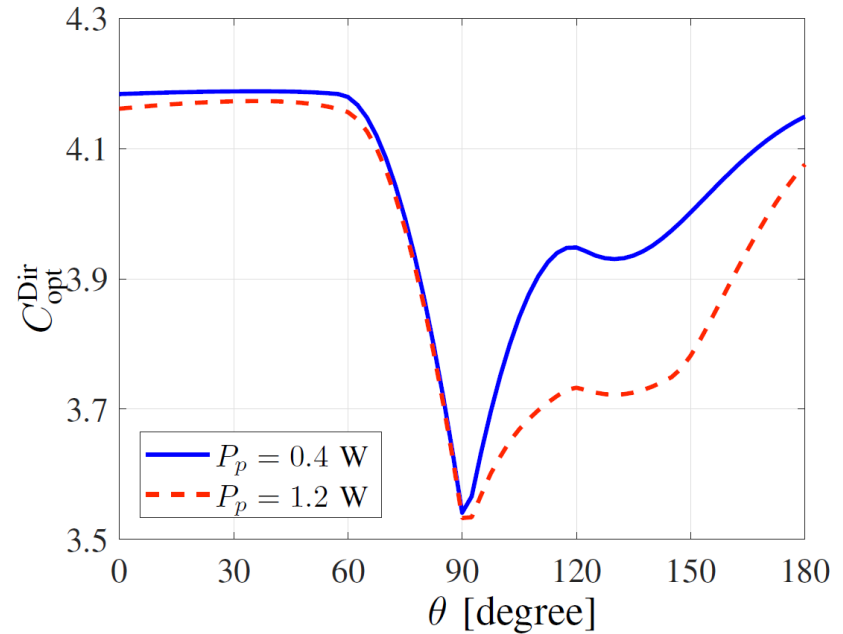
Variations of $C_{\text{opt}}^{\text{Dir}}$ versus ε .

$$I_{pk} = 2 \text{ dB}, \quad P_{pk} = 10 \text{ dB}, \quad \varphi_{3dB} = 30^\circ, \quad \varepsilon = 0.05$$

Simulation Results



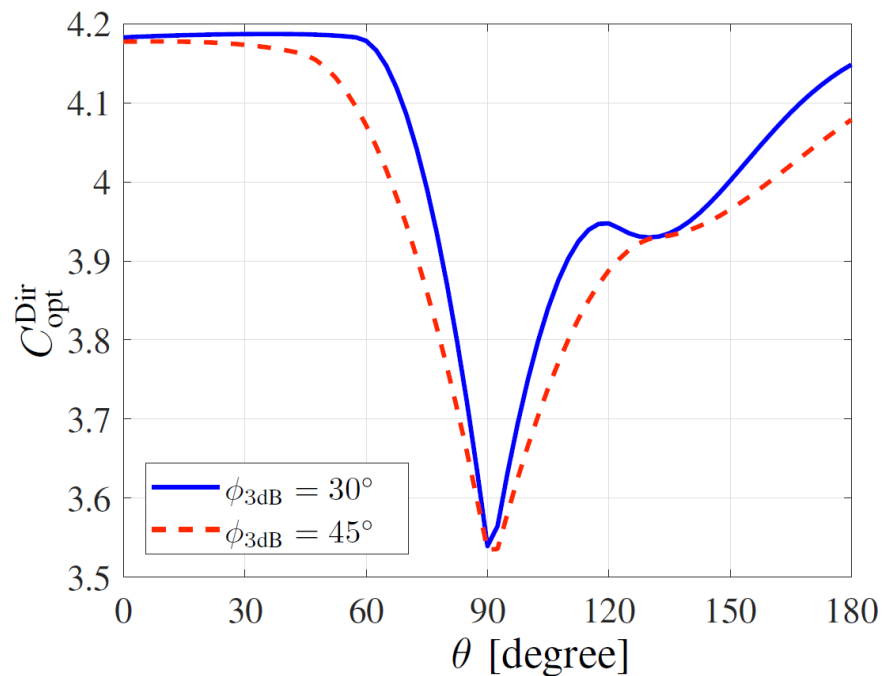
C_{opt}^{Dir} versus θ for different ϵ .



C_{opt}^{Dir} versus θ for different P_p .

$$\phi_{pr} = 90^\circ$$

Simulation Results



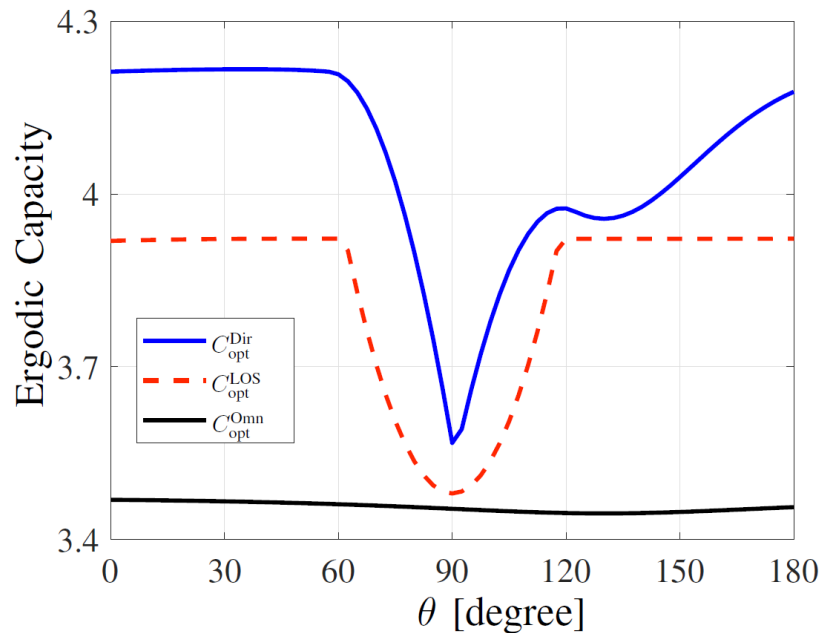
$C_{\text{opt}}^{\text{Dir}}$ versus θ for $\phi_{3\text{dB}} = 30^\circ, 45^\circ$.

Simulation Results

$$C_{\text{opt}}^{\text{Dir}} \rightarrow P, \tau, \phi_t, \phi_r$$

$$C_{\text{opt}}^{\text{Omn}} \rightarrow P, \tau$$

$$C_{\text{opt}}^{\text{LOS}} \rightarrow P, \tau \quad (\phi_t = \theta, \phi_r = \pi + \theta)$$



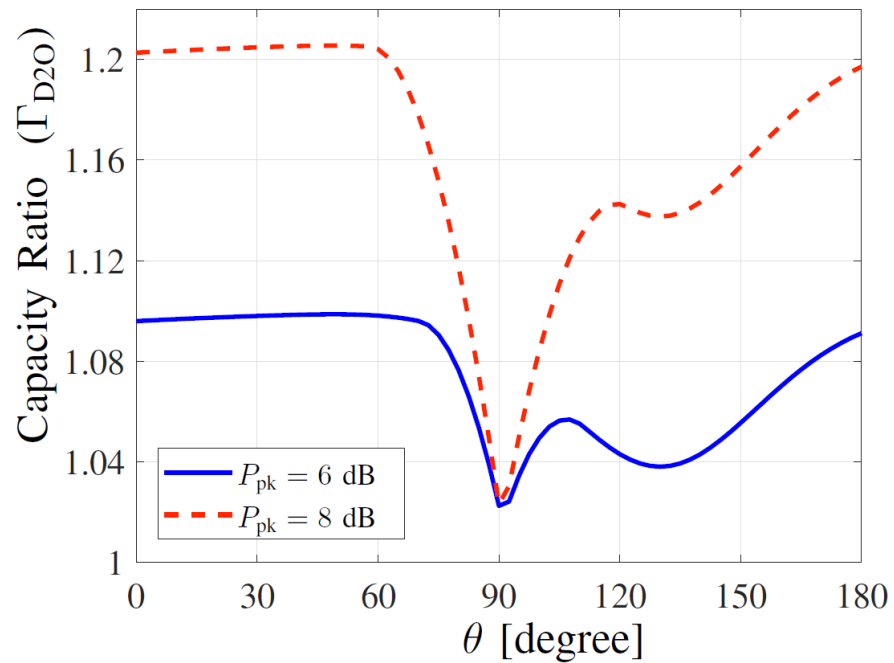
$C_{\text{opt}}^{\text{Dir}}$, $C_{\text{opt}}^{\text{LOS}}$ and $C_{\text{opt}}^{\text{Omn}}$ versus θ .

$$\varphi_{3dB} = 30^\circ$$

Simulation Results

capacity ratio

$$\Gamma_{D2O} = C_{opt}^{Dir} / C_{opt}^{Omn}$$



Γ_{D2O} versus θ for $P_{pk} = 6, 8$ dB.



Conclusion

- The directional antennas can identify and enable transmission and reception across spatial domain and enhance spectrum utilization, compared with omnidirectional antennas.
- There are trade-offs between sensing and capacity in terms of sensing duration τ and SU_{tx} 's antenna orientation.
- An antenna with narrower half-power beam-width always yields higher capacity since it can cancel more interference from PU_{tx} .
- The capacity ratio Γ_{D20} is higher for large P_{pk} .
- We showed the effectiveness of using directional antennas and as well as optimizing their orientation on the capacity of CR network.



Thank you for your attention