

# GlobalSIP



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## Orbital Angular Momentum-Based Two-Dimensional Super-Resolution Targets Imaging

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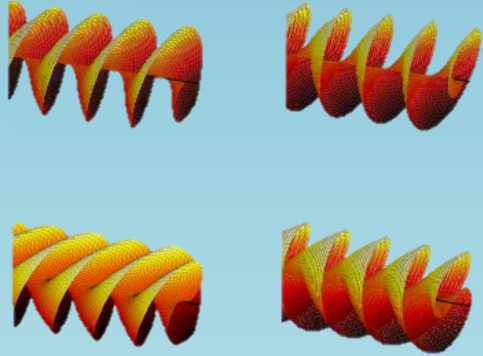
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# INTRODUCTION

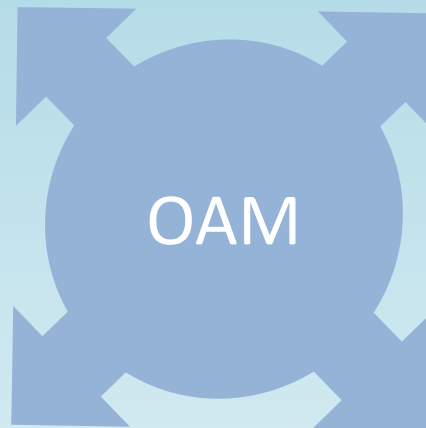


Since the discovery in 1992 that vortex light beams can carry orbital angular momentum (OAM) <sup>[1]</sup>, significant research effort has been focused on the OAM of vortex electromagnetic waves <sup>[2], [3]</sup>. The phase front of vortex electromagnetic wave rotates with azimuth exhibiting a helical structure  $e^{j\alpha\phi}$  in space as shown in the picture on the left.

Different OAM modes

Inherent orthogonality

A new degree of freedom



Wireless Communications

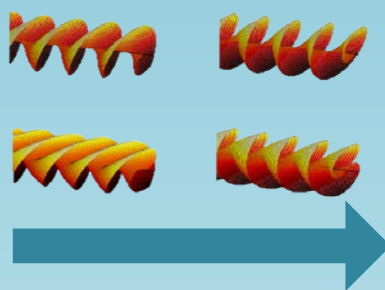
Radar

# INTRODUCTION



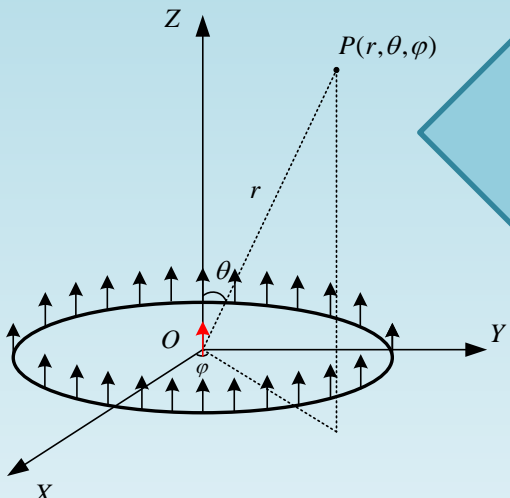
## Radar Application

The helical phase of vortex electromagnetic wave can be seen as multiple plane electromagnetic waves illuminating from continuous azimuth simultaneously

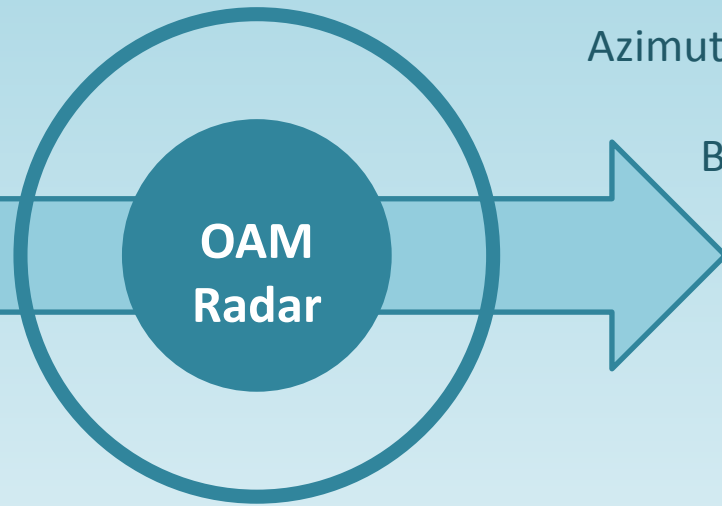


Achieves angular diversity without relative motion or beam scanning

Radar target detection model



Azimuth estimation algorithm



Back projection algorithms  
FFT algorithms  
2-D algorithms  
MUSIC algorithm

# SYSTEM MODEL



## MISO OAM radar targets detection mode

Electric field vector

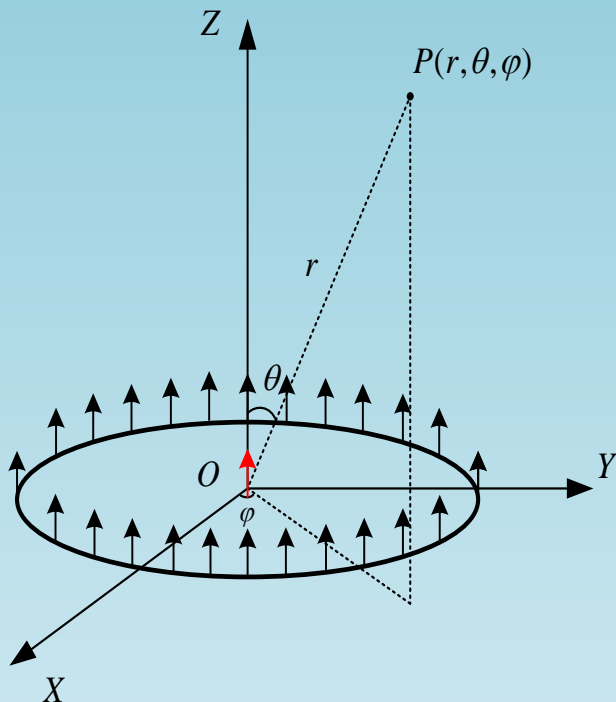
$$\mathbf{E}_s(\mathbf{r}) = -j \frac{\mu_0 \omega d N e^{i k r} e^{i \alpha \varphi}}{4 \pi r} i^{-\alpha} J_\alpha(k R \sin \theta)$$

When  $k R \sin \theta \gg 1$ ,

$$J_\alpha(k R \sin \theta) \approx \sqrt{\frac{2}{\pi k a \sin \theta}} \cos\left(k R \sin \theta - \frac{\alpha \pi}{2} - \frac{\pi}{4}\right)$$

Received echo signal

$$s(k, \alpha) = \sum_{m=1}^M \sigma'_m e^{i 2 k r_m} e^{i \alpha \varphi_m} J_\alpha(k R \sin \theta_m)$$



The  $N$  array elements are located uniformly along the perimeter of a circle and are fed with the same input signal but with successive phase shifts  $\phi_n = \alpha \varphi_n = \alpha \cdot \frac{2\pi n}{N}$ ,  $n = 0, 1, 2, \dots, N - 1$ . Thus, after a full turn the phase has the increment of  $2\pi\alpha$ .

# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Algorithm Principle

**ESPRIT**

Rotational  
Invariance

- Reduced sensitivity to array perturbations
- Improved performance
- Reduced computational load
- Freedom from array calibration



**2-D super-resolution  
targets detection  
based on  
ESPRIT algorithm**

## Estimation of Target Range

$$s(k_p) = \sum_{m=1}^M \sigma'_m J_\alpha(k_p R \sin \theta_0) e^{i2kr_m}$$

# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Estimation of Target Range

The echo signal samples

$$\mathbf{x}(k) = [s(k_1)/J_0(k_1 R \sin \theta_0), s(k_2)/J_0(k_2 R \sin \theta_0), \dots, s(k_p)/J_0(k_p R \sin \theta_0)]^T = \mathbf{A}_r \boldsymbol{\sigma} + \mathbf{n}$$

The covariance matrix

$$\mathbf{R}_x = \mathbf{E}[\mathbf{x}(k)\mathbf{x}^H(k)] = \mathbf{A}_r \mathbf{R}_\sigma \mathbf{A}_r^H + \rho_n \boldsymbol{\Lambda}_J$$

Where

$$\mathbf{A}_r = \begin{bmatrix} e^{i2k_1 r_1} & e^{i2k_1 r_2} & \dots & e^{i2k_1 r_M} \\ e^{i2k_2 r_1} & e^{i2k_2 r_2} & \dots & e^{i2k_2 r_M} \\ \vdots & \vdots & \ddots & \vdots \\ e^{i2k_p r_1} & e^{i2k_p r_2} & \dots & e^{i2k_p r_M} \end{bmatrix}$$

$$\boldsymbol{\sigma} = [\sigma'_1, \sigma'_2, \dots, \sigma'_M]^T$$

$$\mathbf{n} = [n_1/J_0(k_1 R \sin \theta_0), n_2/J_0(k_2 R \sin \theta_0), \dots, n_p/J_0(k_p R \sin \theta_0)]^T$$

$\boldsymbol{\Lambda}_J$

$$\mathbf{R}_\sigma = \mathbf{E}[\boldsymbol{\sigma}\boldsymbol{\sigma}^H] = \text{diag}\{J_0^{-2}(k_1 R \sin \theta_0), J_0^{-2}(k_2 R \sin \theta_0), \dots, J_0^{-2}(k_p R \sin \theta_0)\}$$

# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Estimation of Target Range

### Singular Value Decomposition (SVD) Algorithm

Eigenvalue Decomposition:

$$\mathbf{R}_X = \mathbf{Q}\mathbf{\Lambda}\mathbf{Q}^H$$

Where

$$\mathbf{\Lambda} = \text{diag}\{\lambda_1, \lambda_2, \dots, \lambda_M\}$$

Eigenvalue Decomposition:

$$\mathbf{R}_X \mathbf{q} = \lambda_{max} \mathbf{q}$$

Where

$$\mathbf{q} = [q_1, q_2, \dots, q_P]^T$$

Reconstruct the Covariance Matrix

$$\mathbf{R}^f = \begin{matrix} & q_1 & q_2 & q_3 & \cdots & q_{P-i+1} \\ & q_2 & q_3 & q_4 & \cdots & q_{P-i+2} \\ \mathbf{R}^f = & q_3 & q_4 & q_5 & \cdots & q_{P-i+3} \\ & \vdots & \vdots & \vdots & \ddots & \vdots \\ & q_i & q_{i+1} & q_{i+2} & \cdots & q_P \end{matrix}$$



# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Estimation of Target Range

### The steps of target range estimation

- (i) Calculate the eigenvector  $\mathbf{q}$  corresponding to the largest eigenvalue of the covariance matrix  $\mathbf{R}_X$ .
- (ii) Reconstruct  $\mathbf{R}^f$  based on  $\mathbf{q}$ .
- (iii) Perform SVD on  $\mathbf{R}^f$  as  $\mathbf{R}^f = \hat{\mathbf{U}}\Sigma\hat{\mathbf{V}}^H$

Where  $\hat{\mathbf{U}}$  is an  $i \times i$  unitary matrix and  $\hat{\mathbf{V}}$  is a  $(P - i + 1) \times (P - i + 1)$  unitary matrix. Then, we can obtain signal subspace  $\hat{\mathbf{U}}_s$  that is the part of  $\hat{\mathbf{U}}$  corresponding to  $M$  large singular values containing distance information of  $M$  targets.

- (iv) Construct new matrices  $\hat{\mathbf{U}}_1$  taking the first  $i - 1$  rows of  $\hat{\mathbf{U}}_s$  and  $\hat{\mathbf{U}}_2$  taking the last  $i - 1$  rows of  $\hat{\mathbf{U}}_s$ .
- (v) Calculate  $\Psi = (\hat{\mathbf{U}}_1^H \hat{\mathbf{U}}_1)^{-1} \hat{\mathbf{U}}_1^H \hat{\mathbf{U}}_2$
- (vi) Perform eigenvalue decomposition on  $\Psi$  as  $\Psi = \hat{\mathbf{Q}}\Lambda_r\hat{\mathbf{Q}}^{-1}$ .

Where  $\Lambda_r = \text{diag}\{e^{i2r_1}, e^{i2r_2}, \dots, e^{i2r_M}\}$

# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Estimation of Target Azimuth

The echo signal samples

$$\mathbf{y}(\alpha) = [s(\alpha_1)/J_{\alpha_1}(k_p R \sin \theta_0), s(\alpha_2)/J_{\alpha_2}(k_p R \sin \theta_0), \dots, s(\alpha_Q)/J_Q(k_p R \sin \theta_0)]^T = \mathbf{A}_\varphi \mathbf{a} + \mathbf{m}$$

Where

$$\mathbf{A}_\varphi = \begin{bmatrix} e^{i\alpha_1\varphi_1} & e^{i\alpha_1\varphi_2} & \dots & e^{i\alpha_1\varphi_M} \\ e^{i\alpha_2\varphi_1} & e^{i\alpha_2\varphi_2} & \dots & e^{i\alpha_2\varphi_M} \\ \vdots & \vdots & \ddots & \vdots \\ e^{i\alpha_Q\varphi_1} & e^{i\alpha_Q\varphi_2} & \dots & e^{i\alpha_Q\varphi_M} \end{bmatrix}$$

$$\mathbf{a} = [a_1(k_p), a_2(k_p), \dots, a_M(k_p)]^T, a_m(k_p) = \sigma'_m e^{i2k_p r_m} J_\alpha(k_p R \sin \theta_m)$$

# 2-D SUPER-RESOLUTION TARGETS IMAGING



## Numerical Simulation and Results

### Simulation Data

$$r_1 = 40m, r_2 = 45m, r_3 = 50m$$

$$(\varphi_1, \theta_1) = (10^\circ, 70^\circ), (\varphi_2, \theta_2) = (15^\circ, 70^\circ), (\varphi_3, \theta_3) = (20^\circ, 70^\circ)$$

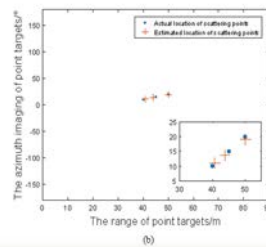
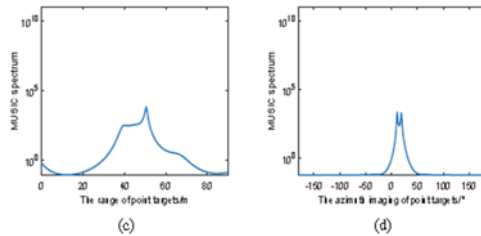
*Discrete frequencies: 9.024GHz to 9.979GHz*

*Wave number:  $k = 189, 190, \dots, 209$*

*OAM modes:  $\alpha = -16, -15, \dots, 15$*

# 2-D SUPER-RESOLUTION TARGETS IMAGING

## Numerical Simulation and Results Simulation Result

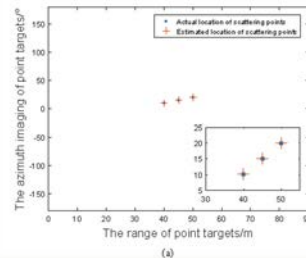
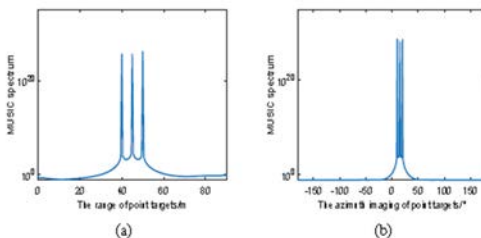


MUSIC  
16.5dB

ESPRIT  
16.5

MUSIC  
25dB

ESPRIT  
25dB



Both MUSIC and ESPRIT algorithms-based methods can fulfill 2-D imaging of three close point targets at higher SNR, however, ESPRIT algorithms-based methods can provide the positions of targets directly instead of searching spectrum peaks. Furthermore, at lower SNR ESPRIT method can still distinguish three point targets while only one or two blunt peaks could be seen by MUSIC method. It follows that the proposed ESPRIT algorithm-based method is more robust to the MUSIC algorithm-based method.

# CONCLUSIONS



ESPRIT algorithm-based 2-D OAM radar targets detection method shows the 2-D super-resolution capability of OAM-based radar. In contrast to conventional algorithm, both MUSIC algorithm-based radar targets detection and ESPRIT algorithm-based OAM radar targets detection belong to super-resolution imaging methods. Meanwhile, our proposed ESPRIT algorithm-based 2-D OAM radar targets detection method further outperforms MUSIC.

# THANKS!

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