

Constrained Clipping for PAPR Reduction in VLC Systems with Dimming Control

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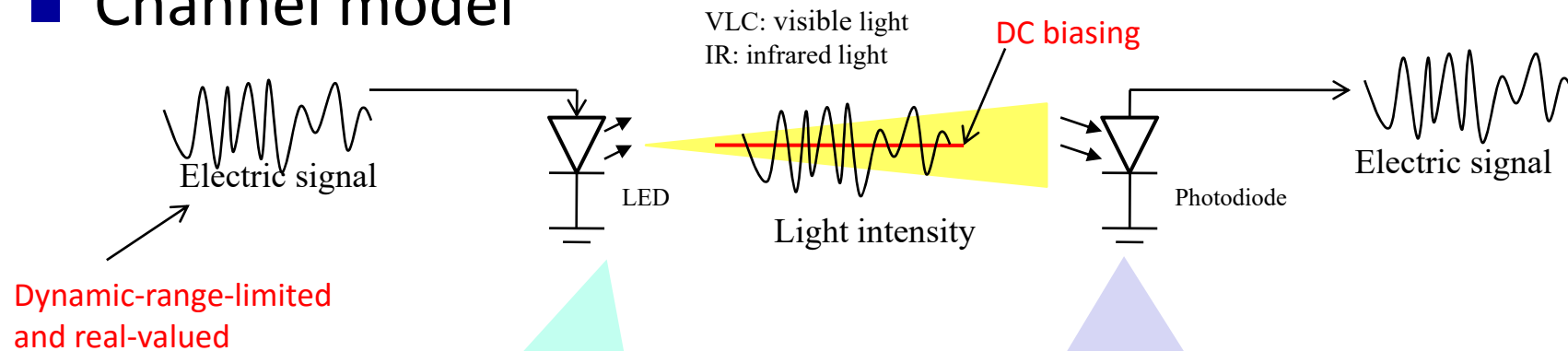
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- Introduction
- Double-sided Clipping
- Constrained Clipping
- Numerical Results

■ Channel model



Transmitter:

light emitting diode (LED) converts the amplitude of the electric signal into the intensity of the optical signal.

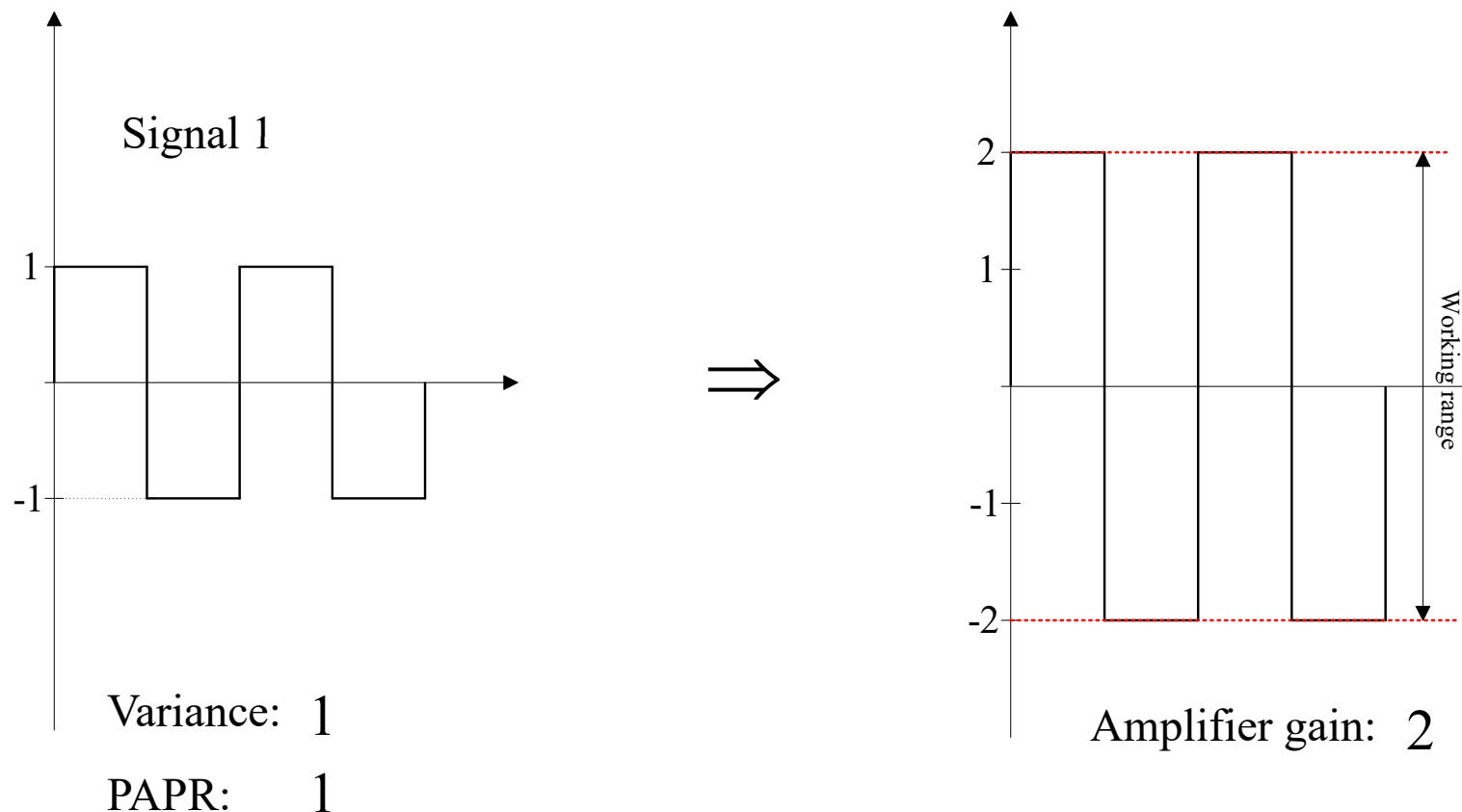
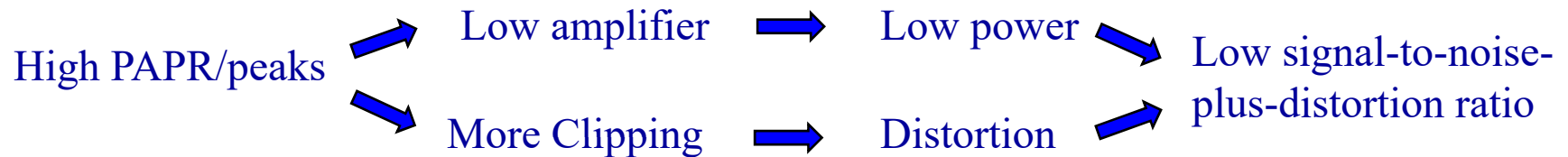
Receiver :

A photodetector (PD) or an image sensor generates an electric signal proportional to the intensity of the received optical signal.

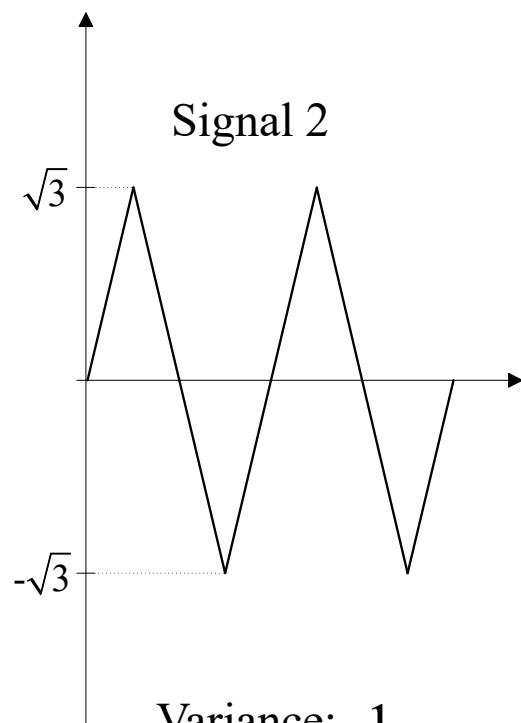
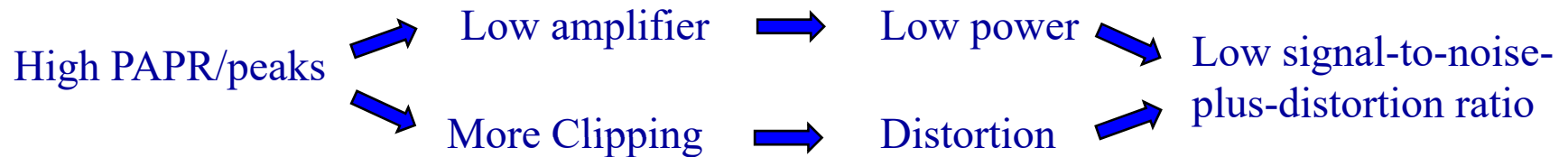
■ Differences from Radio Frequency (RF)

- Real-valued
- Dynamic-range-limited (turn-on/saturation)
- DC biasing required

- Motivation of peak-to-average power ratio (PAPR) reduction
 - *Considering signals with the same variance*

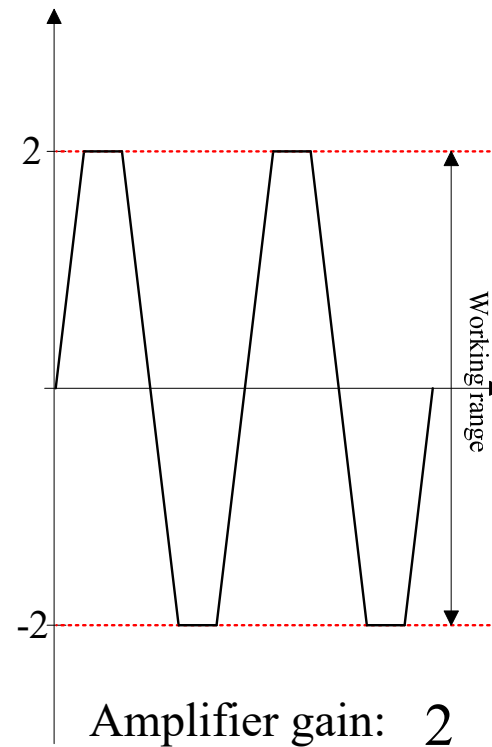


- Motivation of peak-to-average power ratio (PAPR) reduction
 - *Considering signals with the same variance*

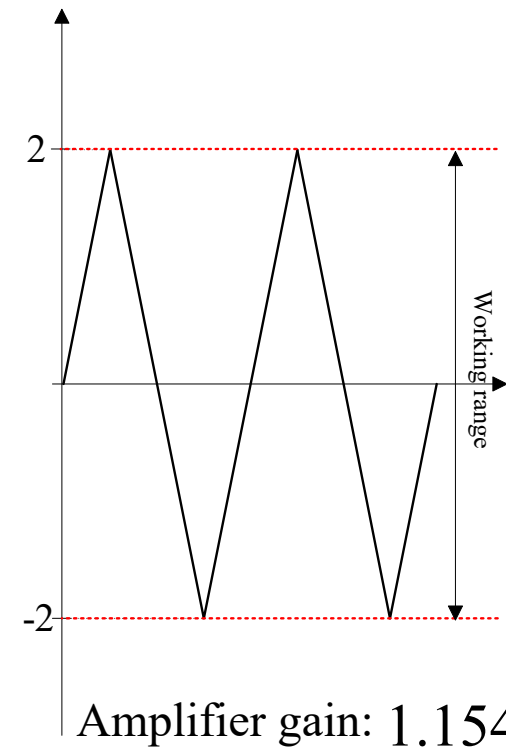


Variance: 1

PAPR: 3



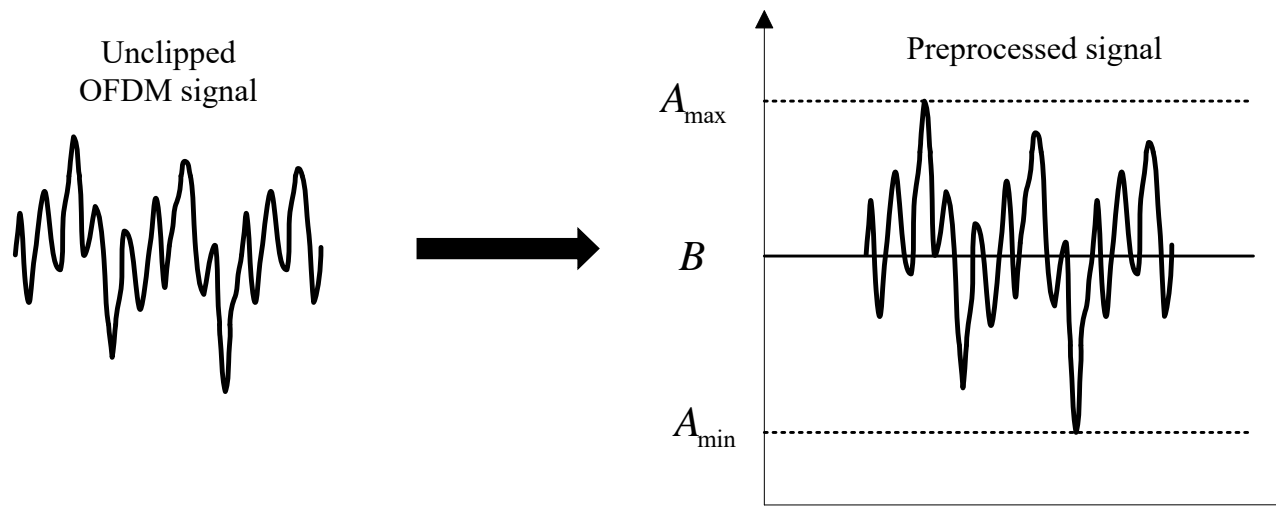
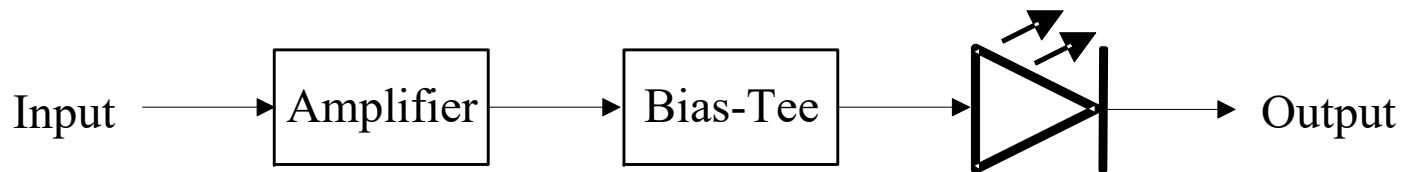
Clipping distortion



Lower signal power

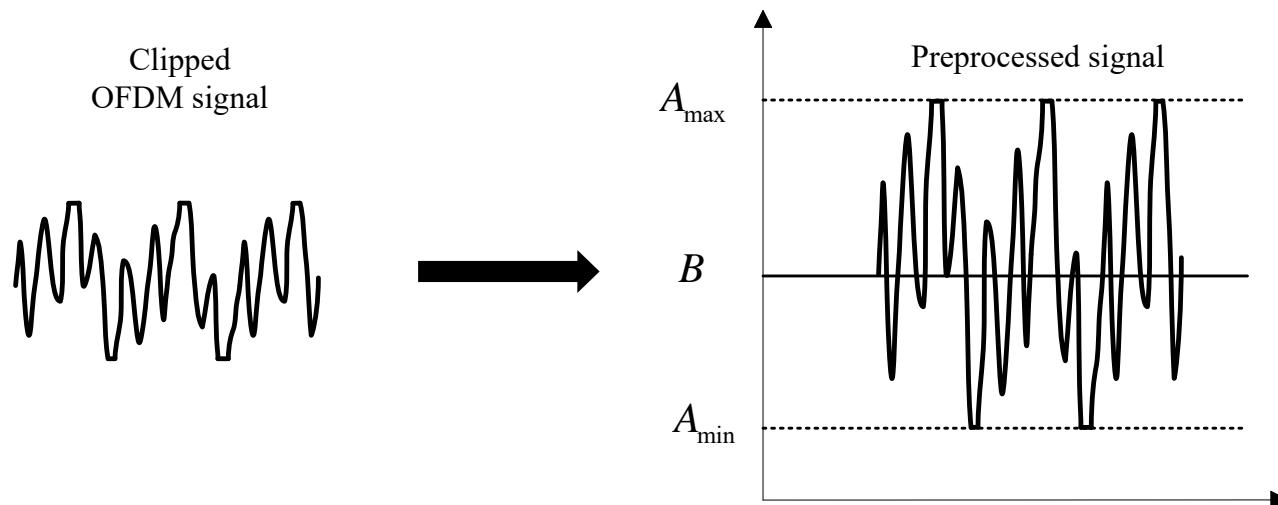
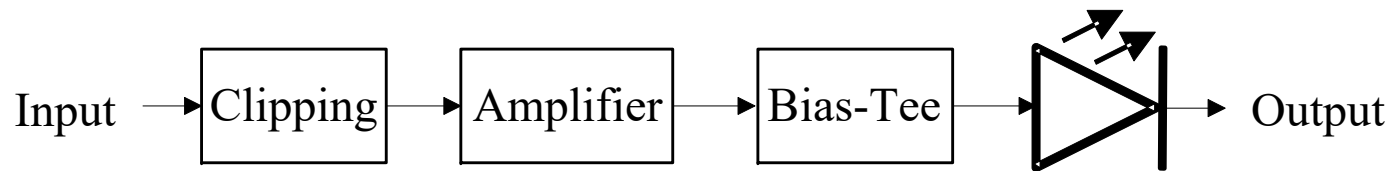
Double-sided Clipping

- Clipping is the simplest way to reduce PAPR
- Motivation of clipping
 - *Tradeoff between clipping distortion and signal power*



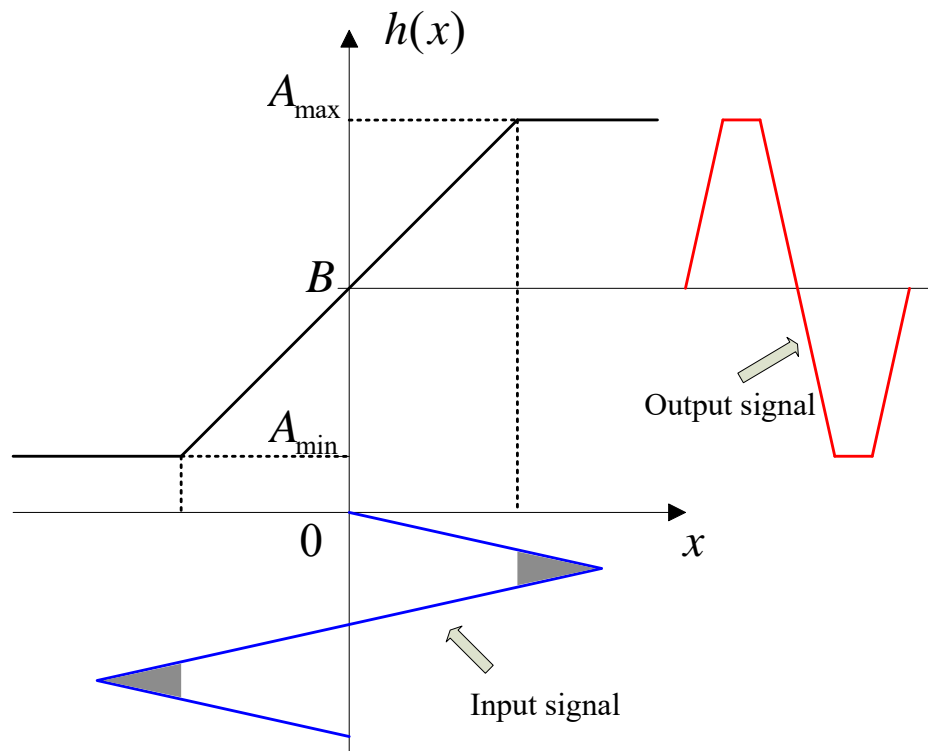
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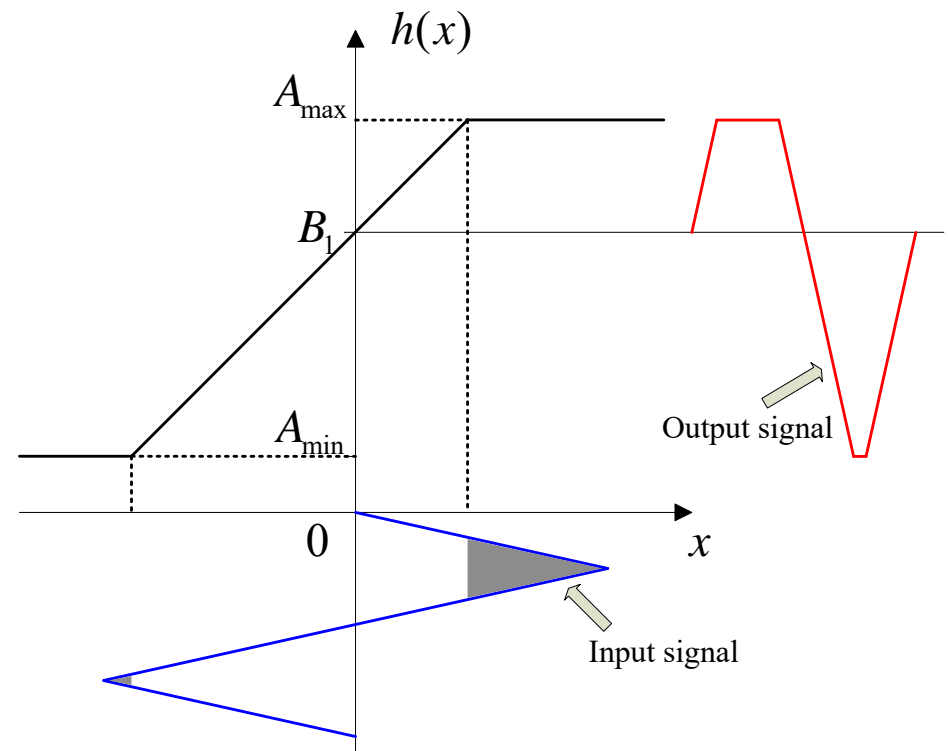


Double-sided Clipping

- Difference from amplitude-limited system
 - *Double-sided clipping*
 - *Clipping ratio is impacted by both gain and bias*



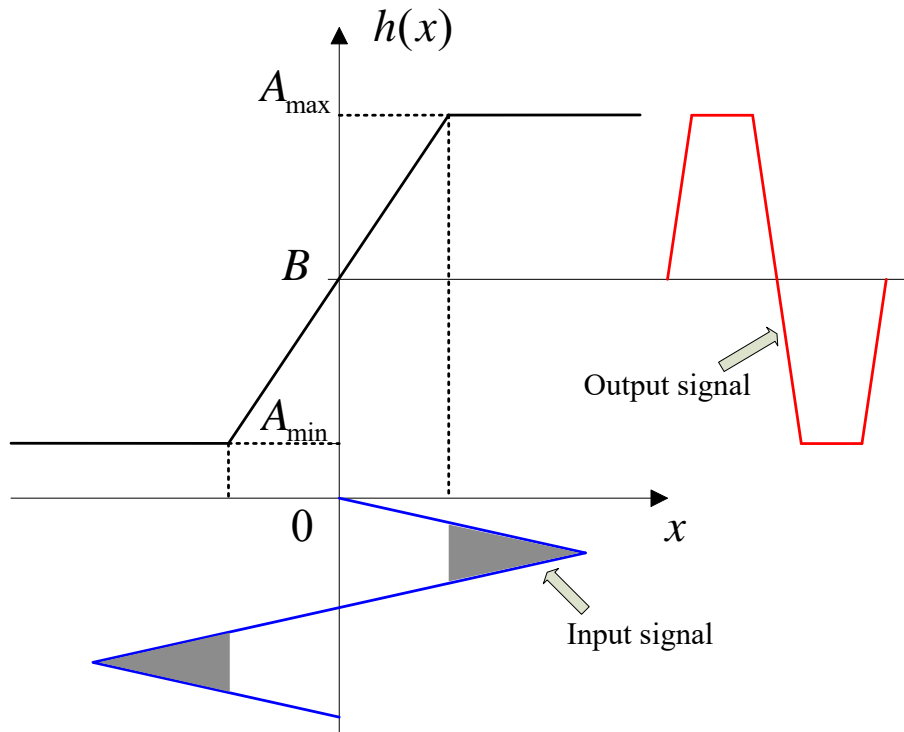
$$G=1, B = (A_{\max} + A_{\min}) / 2$$



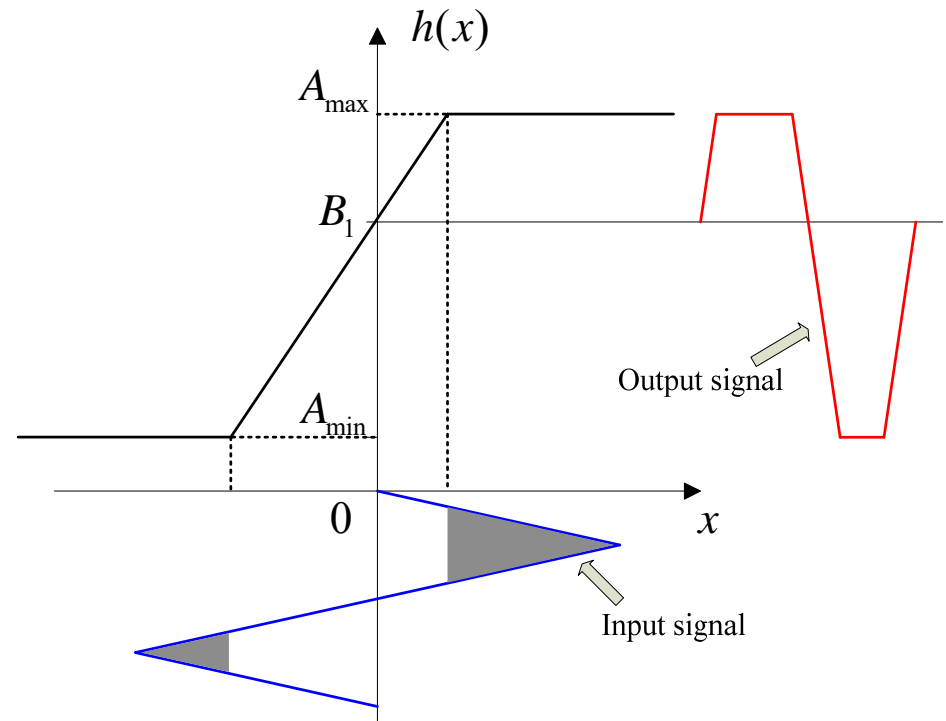
$$G=1, B_1 > B$$

Double-sided Clipping

- Difference from amplitude-limited system
 - *Double-sided clipping*
 - *Clipping ratio is impacted by both gain and bias*



$$G_1 > G, \quad B = (A_{\max} + A_{\min}) / 2$$



$$G_1 > G, \quad B_1 > B$$

Double-sided PAPR

■ Orthogonal frequency division multiplexing (OFDM)

■ *Time domain:* x_n

Real-valued

■ *Frequency domain:* $X_k = X_{-k}^*, 1 \leq k \leq N/2 - 1$

Hermitian symmetry

■ PAPR on both sides

■ *Upper PAPR (UPAPR):*

$$\mathcal{U}(x_n) \triangleq \frac{\left(\max_{0 \leq n \leq N-1} x_n \right)^2}{\sigma_x^2}$$

■ *Lower PAPR (LPAPR):*

$$\mathcal{L}(x_n) \triangleq \frac{\left(\min_{0 \leq n \leq N-1} x_n \right)^2}{\sigma_x^2}$$

■ Asymmetric factor

$$\rho \triangleq \frac{(A_{max} - B)^2}{(A_{min} - B)^2}$$

■ Joint CCDF of UPAPR and LPAPR

$$\text{CCDF}\{\mathcal{U}(x_n), \mathcal{L}(x_n), \gamma, \rho\}$$

$$\triangleq 1 - \text{Pr}\{\mathcal{U}(x_n) \leq \gamma, \mathcal{L}(x_n) \leq \gamma/\rho\}$$

Simple Clipping

■ Clipping in time domain

$$\bar{x}_n = \begin{cases} x_{max}, & x_n > x_{max}, \\ x_n, & x_{min} \leq x_n \leq x_{max}, \\ x_{min}, & x_n < x_{min}. \end{cases}$$

$$x_{max} = \frac{A_{max} - B}{G}$$

$$x_{min} = \frac{A_{min} - B}{G}$$

■ Distortion in frequency domain

$$E_k = \bar{X}_k - X_k$$

■ Error vector magnitude (EVM)

$$\text{EVM} \triangleq \frac{1}{\sigma_X} \sqrt{\frac{1}{N} \sum_{k \in \mathcal{I}} |E_k|^2}$$

■ Clipping ratio / Normalized clipping levels

$$\text{Upper side: } \lambda_{upper} = \frac{x_{max}}{\sigma_x}$$

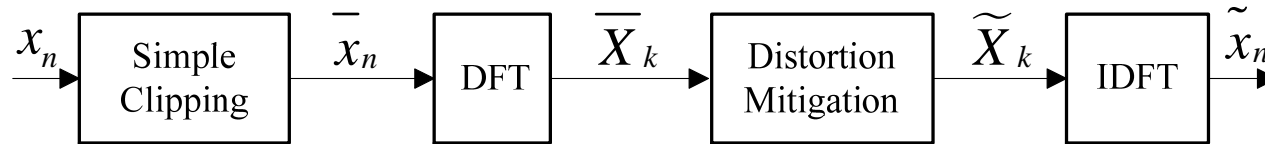
$$\text{Lower side: } \lambda_{lower} = -\frac{x_{min}}{\sigma_x}$$

$$\lambda_{upper} = \sqrt{\rho} \lambda_{lower}$$

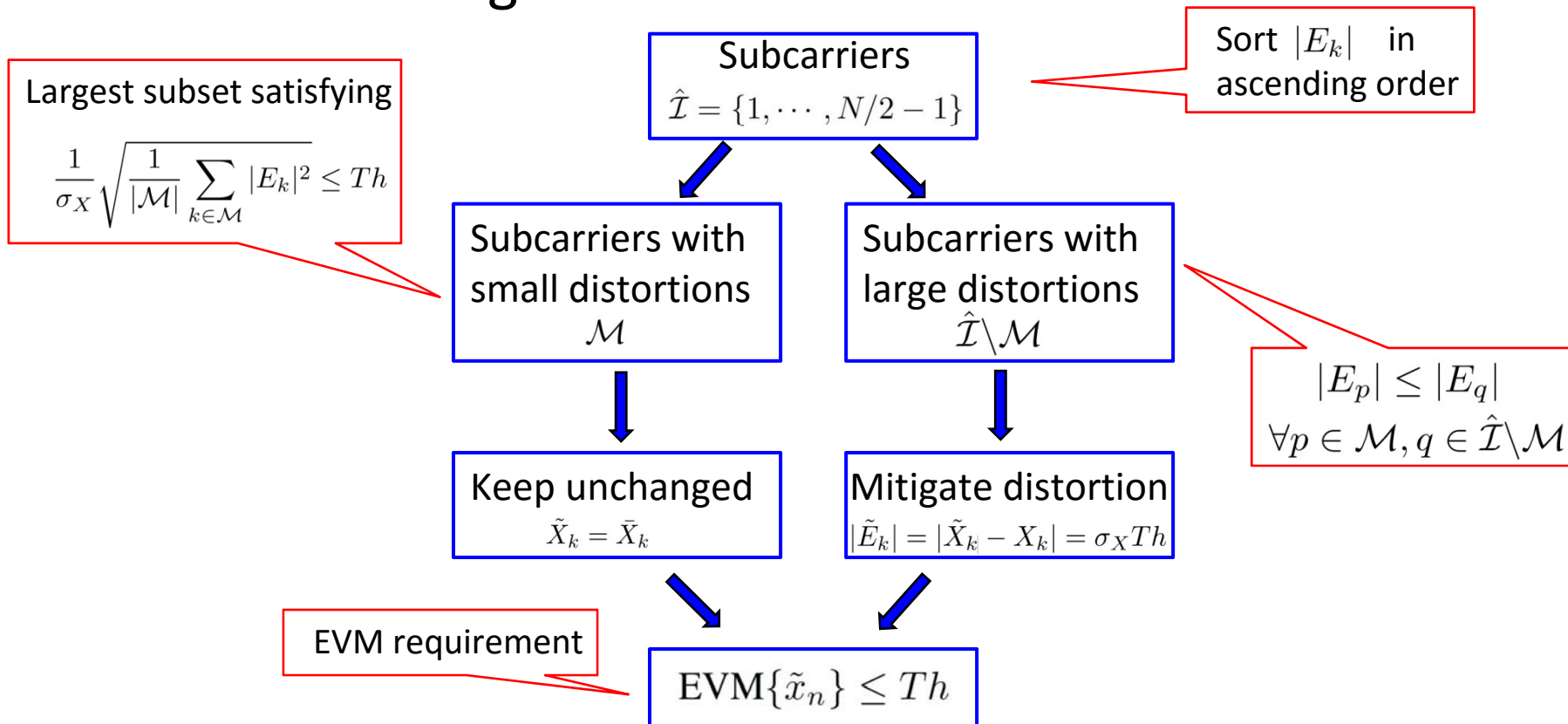
- To satisfy the EVM requirement, clipping levels cannot be very low.

Too conservative!

Block diagram



Distortion Mitigation

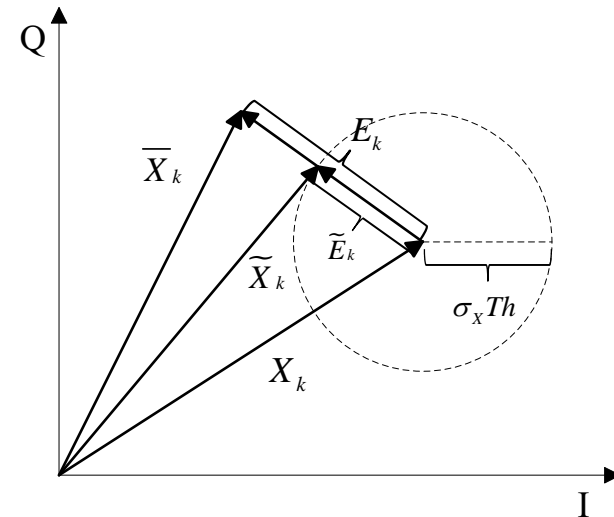


- How to revise subcarriers with large distortions
 - *Modifying frequency domain symbols will change the time domain signals. We make the change as small as possible.*
 - *Parseval's Theorem*

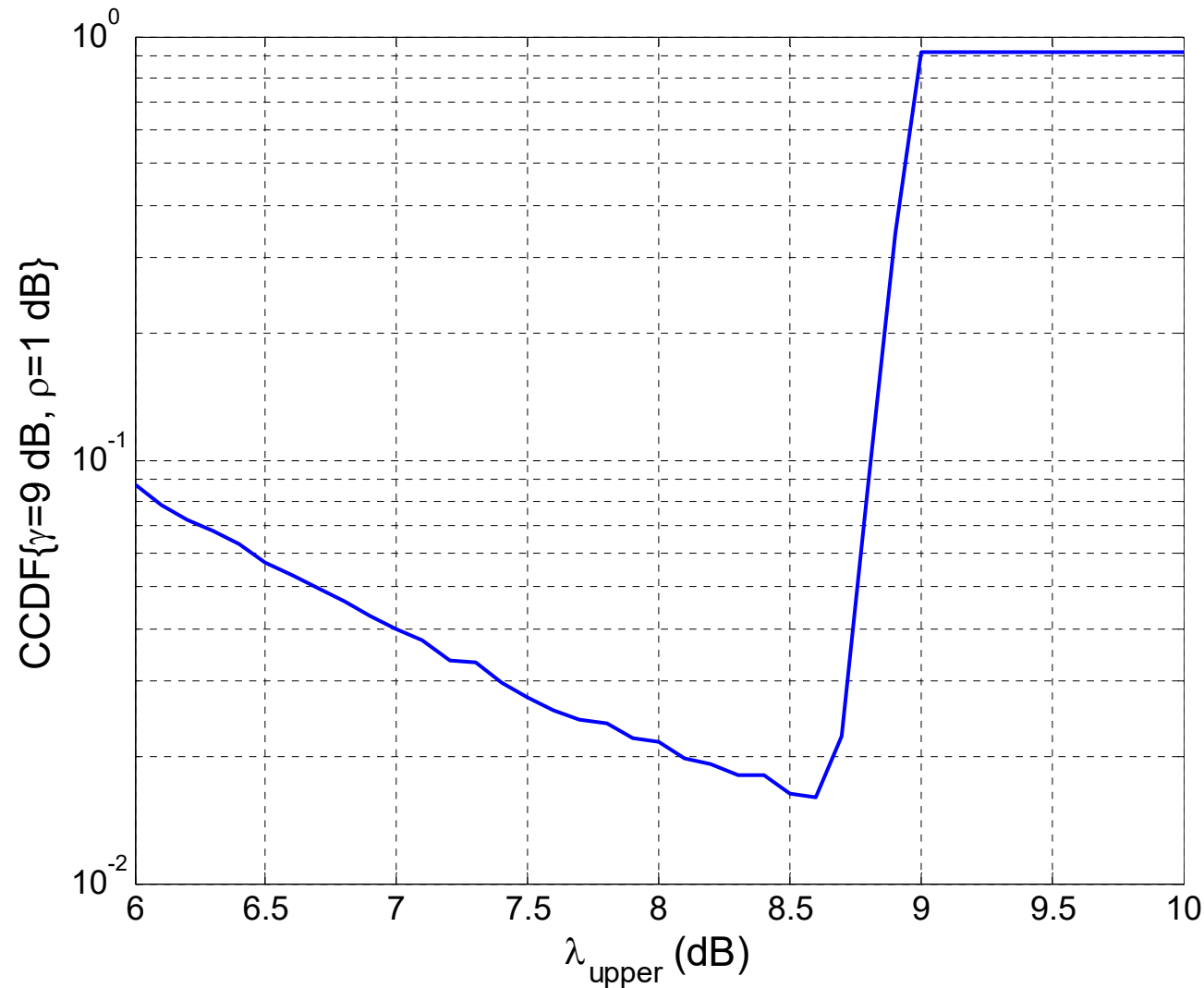
$$\sum_{n=0}^{N-1} |\tilde{x}_n - \bar{x}_n|^2 = \sum_{k \in \mathcal{I}} |\tilde{X}_k - \bar{X}_k|^2$$

- *Minimize the change in frequency domain*

$$\tilde{X}_k = X_k + Th \cdot \sigma_X e^{j\angle E_k}, \quad k \in \hat{\mathcal{I}} \setminus \mathcal{M}$$



- Probability that signal exceeds 9 dB in UPAPR or 8 dB in LPAPR



Thank you!

Q&A