

Introducing Complex Functional Link Polynomial Filters

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Abstract

- We introduce a novel class of complex nonlinear filters, the **complex functional link polynomial (CFLiP)** filters.
- CFLiP filters are a sub-class of **linear-in-the-parameter** (LIP) nonlinear filters.
- They satisfy all conditions of **Stone-Weirstrass theorem** and thus are **universal approximators**.
- The CFLiP basis functions **separate** the **magnitude** and **phase** of the input signal.
- Moreover, CFLiP filters include many families of nonlinear filters with **orthogonal basis functions**.
- They are capable of modeling the nonlinearities of high power amplifiers (HPAs) of telecommunication systems with **better accuracy** than most of the filters currently used for this purpose.

Nonlinear system expansion

- Let us consider a causal, time-invariant, finite-memory, continuous nonlinear systems:

$$y(n) = f[x(n), x(n-1), \dots, x(n-N+1)].$$

with $x(n) \in \mathbb{C}_1 = \{x \in \mathbb{C}, \text{ and } |x| \leq 1\}$, $y(n) \in \mathbb{C}$.

- f is a **continuous function** from \mathbb{C}_1^N to \mathbb{C} .
- We can expand f with a series of basis functions $f_i(n)$:

$$y(n) = \sum_{i=1}^{+\infty} c_i f_i[x(n), \dots, x(n-N+1)],$$

- with $c_i \in \mathbb{C}$, and f_i a continuous function from \mathbb{C}_1^N to \mathbb{C} , $\forall i$.
- Any choice of the basis functions f_i defines a different class of filters.

The Stone-Weierstrass theorem

- We want to develop a class of complex nonlinear filters that can **arbitrarily well approximate** the system on the compact \mathbb{C}_1 according to the Stone-Weierstrass (S-W) theorem:

“Suppose \mathcal{A} is a self-adjoint algebra of complex continuous functions on the compact set K , \mathcal{A} separates points on K , and \mathcal{A} vanishes at no point on K , then the uniform closure \mathcal{B} of \mathcal{A} consists of all complex continuous functions on K ”.

- A family \mathcal{A} of complex functions is said to be an **algebra** if \mathcal{A} is closed under addition, multiplication, and scalar multiplication.
- An algebra \mathcal{A} is **self-adjoint** if for any complex function $f \in \mathcal{A}$ also the conjugate f^* belongs to \mathcal{A} .

CFLiP basis functions

- We separate module and phase of $x(n)$ using the polar form
$$x(n) = r(n)e^{j\phi(n)}, r(n) \in [0, 1] \text{ and } \phi(n) \in [0, 2\pi].$$
- Let $g_0[r(n)], g_1[r(n)], \dots$, be a set of **real** basis functions satisfying S-W for the continuous real functions defined in $[0, 1]$; $g_0[r(n)] = 1, g_i[r(n)]$ a polynomial of order i .
- The N -dimensional **basis functions** of CFLiP filters are defined as follows:

$$g_{k_0}[r(n)]e^{jp_0\phi(n)} \cdot \dots \cdot g_{k_{N-1}}[r(n - N + 1)]e^{jp_{N-1}\phi(n-N+1)}$$

- with $k_0, \dots, k_{N-1} \in \mathbb{N}$ and $p_0, \dots, p_{N-1} \in \mathbb{Z}$.
- We define the order of a basis function as $K = k_0 + \dots + k_{N-1}$, and the phase as $P = p_0 + \dots + p_{N-1}$.
 - The CFLiP basis functions and their linear combinations form a self-adjoint algebra that satisfies all conditions of S-W theorem.

Basis functions of order 3

For a CFiIP filter with memory N , order $K = 3$, and phase $P = 1$.

$\forall n_1 = 0, \dots, N - 1, \quad n_2 = n_1 + 1, \dots, N - 1, \quad n_3 = n_2 + 1, \dots, N - 1$:

$$g_3[r(n - n_1)]e^{j\phi(n - n_1)}$$

$$g_2[r(n - n_1)]e^{j\phi(n - n_1)}g_1[r(n - n_2)]$$

$$g_2[r(n - n_1)]g_1[r(n - n_2)]e^{j\phi(n - n_2)}$$

$$g_1[r(n - n_1)]e^{j\phi(n - n_1)}g_2[r(n - n_2)]$$

$$g_1[r(n - n_1)]g_2[r(n - n_2)]e^{j\phi(n - n_2)}$$

$$g_2[r(n - n_1)]e^{2j\phi(n - n_1)}g_1[r(n - n_2)]e^{-j\phi(n - n_2)}$$

$$g_1[r(n - n_1)]e^{-j\phi(n - n_1)}g_2[r(n - n_2)]e^{2j\phi(n - n_2)}$$

$$g_1[r(n - n_1)]e^{j\phi(n - n_1)}g_1[r(n - n_2)]g_1[r(n - n_3)]$$

$$g_1[r(n - n_1)]g_1[r(n - n_2)]e^{j\phi(n - n_2)}g_1[r(n - n_3)]$$

$$g_1[r(n - n_1)]g_1[r(n - n_2)]g_1[r(n - n_3)]e^{j\phi(n - n_3)}$$

$$g_1[r(n - n_1)]e^{j\phi(n - n_1)}g_1[r(n - n_2)]e^{j\phi(n - n_2)}g_1[r(n - n_3)]e^{-j\phi(n - n_3)}$$

$$g_1[r(n - n_1)]e^{j\phi(n - n_1)}g_1[r(n - n_2)]e^{-j\phi(n - n_2)}g_1[r(n - n_3)]e^{j\phi(n - n_3)}$$

$$g_1[r(n - n_1)]e^{-j\phi(n - n_1)}g_1[r(n - n_2)]e^{j\phi(n - n_2)}g_1[r(n - n_3)]e^{j\phi(n - n_3)}$$

CFLiP filters

- A CFLiP filter of **memory** N , **order** K , and **phase** P , is a linear combination of basis functions with orders from 0 to K and phase P , satisfying the following **constraints**:
 - All basis functions of order K have phases p_i with $|p_i| \leq k_i$ for $i = 0, \dots, N - 1$.
 - All basis functions of order lower than K have phases p_0, \dots, p_{N-1} equal to one of the basis functions of order K and orders k_0, \dots, k_{N-1} , lower than or equal to those of this basis function.
- A CFLiP filter of **memory** N , **order** K , and **phases** P_0, \dots, P_R is the parallel of R filters having memory N , order K , phase P_i , with $i = 1, \dots, R$.
- A **single phase is often sufficient** to model the nonlinear systems of some applications.

Orthogonal CFLiP filters

- Some families of orthogonal FLiP filters have been proposed for the approximation of real nonlinear systems.
- For example, **even mirror Fourier nonlinear (EMFN)** [1] and **Legendre nonlinear (LN)** [2] filters are orthogonal for a white uniform distribution in $[-1, +1]$.
- We can define the corresponding CFLiP filters, whose basis functions are **orthogonal** for a white uniform distribution in \mathbb{C}_1 .
- The basis functions $\tilde{g}_k(\tilde{r})$ of EMFN and LN filters are defined in $[-1, +1]$, but can be shifted on $[0, 1]$ with $g_k(r) = \tilde{g}_k(2r - 1)$.
- In **Complex EMFN (CEMFN)** filters, we have $g_k(r) = \cos(k\pi r)$.
- In **Complex LN (CLN)** filters, the functions $g_k(r)$ are the Shifted Legendre polynomials,

$$g_k(r) = (-1)^k \sum_{s=0}^k \binom{k}{s} \binom{k+1}{s} (-r)^s.$$

Modelling RF High Power Amplifiers (HPAs)

- HPAs used in modern wireless systems are often operated close to **saturation** level to maximize the energy efficiency.
- While the HPA is typically a memory-less device, the filters that precede and follow the HPA introduce **memory effects**.
- If the HPA in the RF passband can be modeled as a Volterra filter, then its complex baseband representation is a **complex Volterra filter** composed only by **odd kernels** [3].
- Each kernel of order $2Q + 1$ is composed by products of $Q + 1$ direct samples and Q conjugate samples. E.g.,

$$\begin{aligned} &x(n)x(n - n_1)x^*(n - n_2), \\ &x(n)x^*(n - n_1)x(n - n_2), \\ &x^*(n)x(n - n_1)x(n - n_2). \end{aligned}$$

- Replacing the input samples with their polar form, the model is composed by basis functions with **phase** $P = 1$.

Classic models of HPAs

- The memory polynomial filters

$$y(n) = \sum_{k=1}^K \sum_{i=0}^{N-1} a_{ki} x(n-i) |x(n-i)|^{k-1}$$

- The generalized memory polynomial filters [3]

$$y(n) = \sum_{k=1}^K \sum_{i=1}^{N-1} a_{ki} x(n-i) |x(n-i)|^{k-1} + \sum_{k=1}^K \sum_{i=0}^{N-1} \sum_{l=1}^{D_N} b_{kil} x(n-i) |x(n-i-l)|^k + \sum_{k=1}^K \sum_{i=0}^{N-1} \sum_{l=1}^{D_N} c_{kil} x(n-i) |x(n-i+l)|^k$$

- The orthogonal memory polynomial filters [4]

$$y(n) = \sum_{k=1}^K \sum_{i=0}^{N-1} a_{ki} \psi_k[x(n-i)] \quad (\psi_k \text{ shifted Legendre polynomials})$$

- These filters are special cases of the CFLiP filters, but they are **not universal approximators** since they do not satisfy S-W.

A simulation experiment

- CFLiP filters of phase $P = 1$ are interesting candidates for modeling and compensating HPAs.
- We provide some simulation results about the **identification of an HPA model**: the Wiener-Hammerstein model of [4].
- The model is composed by the cascade of a linear time-invariant (LTI) system $H(z)$, followed by a memory-less nonlinearity, in turn followed by a LTI system $G(z)$:

$$H(z) = \frac{1}{1.5} \frac{1+0.25z^{-2}}{1+0.4z^{-1}},$$

$$G(z) = \frac{1}{0.52} \frac{1-0.25z^{-1}}{1-0.2z^{-1}}$$

with **memoryless nonlinearity**:

$$y(n) = (\gamma_1 \tan^{-1}(\zeta_1 |z(n)|) + \gamma_2 \tan^{-1}(\zeta_2 |z(n)|)) e^{j\rho(n)}$$

with $\rho(n)$ the phase of $z(n)$.

- The input signal $x(n)$ is a white uniform noise in \mathbb{C}_1 .
- The signal-to-noise ration is 80 dB.

The HPA identification

- The model has been identified using different polynomial filters with input-output relationship

$$y(n) = [f_1(n), f_2(n), \dots, f_L(n)] \cdot \mathbf{h}$$

with $f_i(n)$ the basis functions, and \mathbf{h} the coefficient vector.

- The coefficients can be estimated with the LS approach:

$$\mathbf{h}_{\text{LS}} = (\mathbf{F}^H \mathbf{F})^{-1} \mathbf{F}^H \mathbf{d}.$$

with \mathbf{d} a column vector of output samples and \mathbf{F} a matrix whose n -th row is formed by the basis functions $[f_1(n), f_2(n), \dots, f_L(n)]$.

- CLN and CLN2 filters for $D_N \geq 2$ outperform by several dB most of the filters currently considered in the literature.
- The improved modeling accuracy is provided by the richer set of phase terms, i.e., from the use of the phase terms $e^{2j\phi(n-n_1)} e^{-j\phi(n-n_2)}, e^{j\phi(n-n_1)} e^{j\phi(n-n_2)} e^{-j\phi(n-n_3)}$.

Identification results for the HPA model

| Filter | N | K | N_D | L | NMSE(dB) | Cond.Num. |
|--------|-----|-----|-------|-----|--------------|-------------------|
| CLN | 5 | 3 | 0 | 20 | -24.7 | 7.78 |
| CLN | 5 | 3 | 2 | 257 | -38.5 | 36.6 |
| CLN | 5 | 3 | 3 | 419 | -41.6 | 43.3 |
| CLN | 5 | 3 | 4 | 530 | -42.3 | 47.4 |
| CLN2 | 5 | 3 | 0 | 15 | -24.1 | 2.54 |
| CLN2 | 5 | 3 | 2 | 103 | -37.8 | 785. |
| CLN2 | 5 | 3 | 3 | 147 | -40.2 | $1.12 \cdot 10^3$ |
| CLN2 | 5 | 3 | 4 | 175 | -40.9 | $1.32 \cdot 10^3$ |
| CEMFN | 5 | 3 | 0 | 20 | -24.7 | 3.38 |
| CEMFN | 5 | 3 | 2 | 257 | -32.0 | 10.5 |
| CEMFN | 5 | 3 | 3 | 419 | -32.4 | 12.4 |
| CEMFN | 5 | 3 | 4 | 530 | -32.4 | 13.6 |
| MP | 5 | 3 | 0 | 15 | -24.7 | $3.30 \cdot 10^3$ |
| MP | 5 | 5 | 0 | 25 | -24.7 | $4.93 \cdot 10^6$ |
| MP | 5 | 7 | 0 | 35 | -24.7 | $6.21 \cdot 10^9$ |

| Filter | N | K | N_D | L | NMSE(dB) | Cond.Num. |
|--------|-----|-----|-------|-----|--------------|-------------------|
| GMP | 5 | 3 | 2 | 39 | -24.7 | $4.85 \cdot 10^3$ |
| GMP | 5 | 3 | 3 | 49 | -24.7 | $7.42 \cdot 10^3$ |
| GMP | 5 | 3 | 4 | 55 | -24.7 | $9.33 \cdot 10^3$ |
| OP | 5 | 3 | 0 | 15 | -24.7 | 2.53 |
| OP | 5 | 5 | 0 | 25 | -24.7 | 3.99 |
| OP | 5 | 7 | 0 | 35 | -24.7 | 5.48 |
| GOP | 5 | 3 | 2 | 39 | -24.7 | 425. |
| GOP | 5 | 3 | 3 | 49 | -24.7 | 719. |
| GOP | 5 | 3 | 4 | 44 | -24.7 | $1.14 \cdot 10^3$ |
| Vp1 | 5 | 3 | 0 | 10 | -23.9 | 30.6 |
| Vp1 | 5 | 3 | 2 | 47 | -29.7 | 45.5 |
| Vp1 | 5 | 3 | 4 | 80 | -30.4 | 49.3 |
| Vp1 | 5 | 5 | 0 | 15 | -24.6 | 916 |
| Vp1 | 5 | 5 | 2 | 203 | -38.1 | $1.67 \cdot 10^3$ |
| Vp1 | 5 | 5 | 3 | 407 | -40.8 | $1.88 \cdot 10^3$ |
| Vp1 | 5 | 5 | 4 | 605 | -41.4 | $2.09 \cdot 10^3$ |

- The model has been identified with :
 CLN filter, CLN2 filter (using polynomials ψ_k of [4]), CEMFN filter, memory polynomial filter (MP), generalized MP filter (GMP) [3], orthogonal polynomial filter (OP) [4], generalized OP filters (GOP), complex Volterra filter composed only by phase 1 terms (Vp1).

Concluding Remarks

- CFLiP filters have been introduced in this paper.
- They belong to the class of **linear-in-the-parameter** nonlinear filters and are **universal approximators** according to the S-W theorem.
- CFLiP filters include many classes of nonlinear filters based on **orthogonal polynomials**.
- The **orthogonality** of the basis functions **improves** the **condition number** of the autocorrelation matrix used in LS identification and **increases** the **convergence speed** of gradient-descent identification algorithms.
- Future work will include the development of **Perfect Periodic Sequences** for CFLiP filters, which guarantee the orthogonality of the basis function on a finite period.

References

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