Blind Deconvolution of Sparse But Filtered Pulses With Linear State Space Models

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Goal: "Explain" observed signal $(y_1, \ldots, y_K) \in \mathbb{R}^K$ as output of a linear state space model (LSSM) of order n

$$\begin{aligned} X_k &= AX_{k-1} + B_k u_k + \epsilon_k \\ y_k &= CX_k + Z_k \end{aligned}$$

with (unknown)

- input signal $u = (u_1, \dots u_K) \in \mathbb{R}^K$
- input vectors $B_1, \ldots, B_K \in \mathbb{R}^{n \times 1}$
- observation noise $Z_k \stackrel{\text{\tiny iid}}{\sim} \mathcal{N}(0, \sigma_Z^2)$
- state noise $\epsilon_k \stackrel{\text{iid}}{\sim} \mathcal{N}(0, \sigma_{\epsilon}^2 I)$
- LSSM parameters: $A \in \mathbb{R}^{n \times n}$ and $C \in \mathbb{R}^{1 \times n}$

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 \Rightarrow require $u = (u_1, \dots u_K)$ to be sparse Tipping¹, automatic relevance determination (ARD) $U_k \sim \mathcal{N}(0, \sigma_{U_k}^2)$, with $\sigma_{U_k}^2$ estimated by maximum likelihood

$$p(y|\theta) = \int \underbrace{p(y|\theta, u)}_{\mathsf{LSSM}} \underbrace{p(u|\theta)}_{\mathsf{ARD \ prior}} \, \mathrm{d}u$$

¹M. E. Tipping, "Sparse Bayesian learning and the relevance vector machine," *Journal of Machine Learning Research*, vol. 1, p. 211–244, 2001

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Why is it sparse?

look at local maxima

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Factor graph representation

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Factor graph representation



- look at local maxima
- local optimality of $(\sigma_{U_k}^2, B_k)$

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- look at local maxima
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- Gaussian message passing In particular, on edge \tilde{U}_k :

$$\begin{split} &\overleftarrow{m}_{\tilde{U}_k} = \overleftarrow{m}_{X_k} - \overrightarrow{m}_{X_k} \\ &\overleftarrow{V}_{\tilde{U}_k} = A \overrightarrow{V}_{X'_k} A^\mathsf{T} + \overleftarrow{V}_{X_k} + \sigma_\epsilon^2 I \end{split}$$

Sparsity Behavior

Marginal log-likelihood with respect to $(\sigma_{U_k}^2, B_k)$

$$2\ln p(y|\hat{\theta}_k, B_k, \sigma_{U_k}^2) \propto \frac{\sigma_{U_k}^2 (B_k^\mathsf{T} \overleftarrow{V}_{\tilde{U}_k}^{-1} \overleftarrow{m}_{\tilde{U}_k})^2}{1 + \sigma_{U_k}^2 B_k^\mathsf{T} \overleftarrow{V}_{\tilde{U}_k}^{-1} B_k} - \ln(1 + \sigma_{U_k}^2 B_k^\mathsf{T} \overleftarrow{V}_{\tilde{U}_k}^{-1} B_k)$$

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Lemma : Local Optimality Condition

If
$$\widetilde{m}_{\widetilde{U}_{k}}^{\mathsf{T}} \widetilde{V}_{\widetilde{U}_{k}}^{-1} \widetilde{m}_{\widetilde{U}_{k}} > 1$$
, then

$$\hat{\sigma}_{U_{k}}^{2} = \left(1 - \frac{1}{\overleftarrow{m}_{\widetilde{U}_{k}}^{\mathsf{T}} \overleftarrow{V}_{\widetilde{U}_{k}}^{-1} \overleftarrow{m}_{\widetilde{U}_{k}}}\right) \|\overleftarrow{m}_{\widetilde{U}_{k}}\|^{2} > 0 \text{ and } \hat{B}_{k} = \frac{\overleftarrow{m}_{\widetilde{U}_{k}}}{\|\overleftarrow{m}_{\widetilde{U}_{k}}\|}$$

Else, $\hat{\sigma}_{U_k}^2 = 0$ and \hat{B}_k is any vector such that $\|\hat{B}_k\| = 1$.



An input (i.e., $\hat{\sigma}_{U_k}^2 > 0$) is introduced only to compensate a discrepancy $\overleftarrow{m}_{\tilde{U}_k} = \overrightarrow{m}_{X_k} - \overleftarrow{m}_{X_k}$ between \overleftarrow{m}_{X_k} and $\overrightarrow{m}_{X_k} \Rightarrow$ Awareness of input at k

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Likelihood Function

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$$\begin{cases} X_k = AX_{k-1} + B_k u_k + \epsilon_k \\ y_k = CX_k + Z_k \end{cases}$$

Likelihood Function: Integral form, X and U as hidden variables

$$p(y|\theta) = \int \int \underbrace{\prod_{k=1}^{K} p(y_k|x_k, \theta) p(x_k|x_{k-1}, u_k, \theta) p(u_k|\theta)}_{p(y, x, u|\theta)} \, \mathrm{d}x \, \mathrm{d}u$$

with $\theta = (C, A, \sigma_Z^2, \sigma_\epsilon^2, B_1, \dots, B_K, \sigma_{U_1}^2, \dots, \sigma_{U_K}^2).$

Introduction Signal Model Joint EM Gaussian Message Passing Simulation Results Conclusion Likelihood Function

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• maximize $p(y|\theta)$ for both input estimation and system identification • use expectation maximization (EM) for a joint estimation

$$\hat{\theta} = \operatorname*{argmax}_{\theta} \mathbb{E} \left[\ln p(y, U, X | \theta) \right] \; .$$

- for fixed $\hat{\theta}\text{, }p(y,x,u|\hat{\theta})$ is Gaussian
- Gaussian message passing for computing required expectations

Introduction Signal Model Joint EM Gaussian Message Passing Simulation Results Conclusion $\prod_{i=1}^{n} M_{i} = \prod_{i=1}^{n} M$

EM Update (X and U as hidden variables)

Maximization step: for each parameter,

simple closed-form expression or minimization of a quadratic form

•
$$\hat{\sigma}_{U_k}^2 = \mathbb{E}[U_k^2]$$
, for $k \in \{1, \dots, K\}$
• $\hat{B}_k = \frac{\mathbb{E}[U_k X_k] - \hat{A} \mathbb{E}[U_k X_{k-1}]}{\mathbb{E}[U_k^2]}$, for $k \in \{1, \dots, K\}$
• $\hat{A} = \operatorname*{argmin}_A \operatorname{Tr} \left(AV_A A^{\mathsf{T}} - 2A\xi_A\right)$

•
$$\hat{\sigma}_{\epsilon}^2$$
, \hat{C} , $\hat{\sigma}_Z^2$

Required computations:

 $\mathbb{E}[U_k^2]$, $\mathbb{E}[U_k X_{k-1}]$, $\mathbb{E}[U_k X_k]$, $\mathbb{E}[X_k X_k^{\mathsf{T}}]$, and $\mathbb{E}[X_{k-1} X_k^{\mathsf{T}}]$ \Rightarrow efficiently computed using Gaussian message passing



- \bullet the noise variance σ_Z^2 controls the sparsity level (should be fixed a priori)
- \bullet the state noise variance σ_{ϵ}^2 is the LSSM mismatch and should converge to a low value
- many elements of $(\sigma_{U_k} \cdot ||B_k||)_{k \in \{1,...,K\}}$ converge to zero but will typically not become exact zeros. To obtain exact zeros, use marginal likelihood update: $\overleftarrow{m}_{\tilde{U}_k}^{\mathsf{T}} \overleftarrow{V}_{\tilde{U}_k}^{-1} \overleftarrow{m}_{\tilde{U}_k} \leq 1 \Rightarrow \hat{\sigma}_{U_k}^2 = 0$

Joint EM

Gaussian Message Passing

Likelihood Function

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Special case: matrix A known

Likelihood Function: Integral form, X as hidden variable

$$p(y|\theta) = \int \underbrace{\prod_{k=1}^{K} p(y_k|x_k, \theta) p(x_k|x_{k-1}, \theta)}_{p(y, x|\theta)} dx$$

with $\theta = (C, \sigma_Z^2, \sigma_\epsilon^2, B_1, \dots, B_K, \sigma_{U_1}^2, \dots, \sigma_{U_K}^2).$

Likelihood Function

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Use expectation maximization (EM) algorithm for a joint estimation of heta

$$\hat{\theta} = \operatorname*{argmax}_{\theta} \mathbb{E} \left[\ln p(y, X | \theta) \right]$$

• \hat{B}_k : eigenvector of $\mathbb{E}\left[(X_k - AX_{k-1})(X_k - AX_{k-1})^{\mathsf{T}}\right]$ corresponding to the maximum eigenvalue λ_k

•
$$\hat{\sigma}_{U_k}^2 = \max\left(0, \lambda_k - \hat{\sigma}_{\epsilon}^2\right)$$
, for $k \in \{1, \dots, K\}$
 \Rightarrow can create exact zeros

•
$$\hat{\sigma}_{\epsilon}^2 = \operatorname*{argmin}_{\sigma_{\epsilon}} \frac{M_A}{\sigma_{\epsilon}^2} + nK \ln(\sigma_{\epsilon}^2) + \sum_{\lambda_k > \sigma_{\epsilon}^2} -\frac{\lambda_k - \sigma_{\epsilon}^2}{\sigma_{\epsilon}^2} + \ln\left(\frac{\lambda_k}{\sigma_{\epsilon}^2}\right)$$

Gaussian Message Passing

Quantities to be computed: $\mathbb{E}[U_k^2]$, $\mathbb{E}[U_k X_{k-1}]$, $\mathbb{E}[U_k X_k]$, $\mathbb{E}[X_k X_k^{\mathsf{T}}]$, $\mathbb{E}[X_{k-1} X_k^{\mathsf{T}}]$



- efficient (matrix multiplications)
- \bullet stable while σ_{ϵ}^2 and any $\sigma_{U_k}^2$ tend to zero
- suitable choice: modified Bryson-Frazier smoother¹
- avoid matrix inversions

¹L. Bruderer, H. Malmberg, and H.-A. Loeliger, "Deconvolution of weakly-sparse signals and dynamical-system identification by Gaussian message passing," *Proc. 2015 IEEE Int. Symp. on Information Theory*, Hong Kong, June 14–19 2015, pp. 326-330.

Overlapping Decaying Sines (high SNR)



Overlapping Decaying Sines (low SNR)



Joint EM

Gaussian Message Passing

Simulation Results

Conclusion

Lines with Jumps (A fixed)





- efficient algorithm for sparse input estimation and system identification
- ARD technique copes well with linear state space models
- robust joint estimation
- flexible framework