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Superimposed Pilots : An Alternative Pilot Structure to Mitigate Pilot Contamination in Massive MIMO

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Outline

• Channel Estimation and Pilot Contamination in Massive MIMO

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- Channel Estimation Using Superimposed Pilots
- Iterative Channel Estimation
- Simulation Results
- Some Comments on Performance

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Channel Estimation in Massive MIMO

• Uplink time slot is partitioned for transmitting uplink pilots and data



- Orthogonal pilots are used for estimating the channel
- Pilots are shared between neighboring cells

Channel Estimation in Massive MIMO

• Received pilot sequence at the $j^{\rm th}$ BS (implicitly assumed in equations) with M antennas and K users per cell

$$\mathbf{Y} = \sum_{\ell=0}^{L-1} \mathbf{H}_{\ell} \mathbf{\Phi}^{T} + \mathbf{W}$$

 \mathbf{H}_ℓ : Channel matrix between the j^{th} BS and users in the ℓ^{th} cell

$$\begin{split} & \Phi : \tau \times K \text{ Matrix of orthogonal pilot sequences } (\Phi^T \Phi^* = \tau \mathbf{I}) \\ & \mathbf{W} : \text{ Additive white Gaussian noise at } j^{\text{th}} \text{ BS} \end{split}$$

• LS channel estimate

$$\widehat{\mathbf{h}}_{j,m}^{\text{CP}} \triangleq \frac{1}{\tau} \mathbf{Y} \boldsymbol{\phi}_m^* = \mathbf{h}_{j,m} + \sum_{\substack{\ell=0\\ \ell \neq j}}^{L-1} \mathbf{h}_{\ell,m} + \frac{1}{\tau} \mathbf{W} \boldsymbol{\phi}_m^*$$

$$\begin{split} \mathbf{h}_{\ell,k} : \ k^{\mathrm{th}} \ \text{column of } \mathbf{H}_{\ell}, \ \mathbf{h}_{\ell,k} \sim \mathcal{CN}(\mathbf{0}, \mathcal{\beta}_{\ell,k}\mathbf{I}) \\ \boldsymbol{\phi}_m : \ m^{\mathrm{th}} \ \text{column of } \mathbf{\Phi} \end{split}$$

Iterative Channel Estimation

Pilot Contamination



Pilot Contamination

• SINR in the uplink and downlink

$$\operatorname{SINR}_{j,m}^{\operatorname{CP-ul}} = \operatorname{SINR}_{j,m}^{\operatorname{CP-dl}} = \frac{\beta_{j,m}^2}{\sum\limits_{\substack{M \to \infty}} \beta_{\ell,m}^2}$$

 $\beta_{\ell,k}$: Large-scale path-loss coefficient between $j^{\rm th}$ BS and the $k^{\rm th}$ user in the $\ell^{\rm th}$ cell

• Corresponding rates

$$\begin{aligned} \mathbf{R}_{j,m}^{\mathrm{CP-ul}} &= \frac{\left(\mathcal{C}_{u} - \tau\right)}{C} \log_{2} \left(1 + \mathrm{SINR}_{j,m}^{\mathrm{CP-ul}}\right) \\ \mathbf{R}_{j,m}^{\mathrm{CP-dl}} &= \frac{\mathcal{C}_{d}}{C} \log_{2} \left(1 + \mathrm{SINR}_{j,m}^{\mathrm{CP-dl}}\right) \end{aligned}$$

C: Coherence time of the channel. $C = C_u + C_d$

Superimposed Pilots

- Pilots of length C_u are transmitted simultaneously along with data
- If $KL \leq C_u$, each user is assigned a unique pilot
- Received signal at jth BS (implicitly assumed in equations) when employing superimposed pilots

$$\mathbf{Y} = \sum_{\ell=0}^{L-1} \sum_{k=0}^{K-1} \mathbf{h}_{\ell,k} \left(\rho_{\ell,k} \mathbf{x}_{\ell,k} + \lambda_{\ell,k} \mathbf{p}_{\ell,k} \right)^{T} + \mathbf{W}$$

 $\rho_{\ell,k}^2$ and $\lambda_{\ell,k}^2$: fractions of the transmit power reserved for pilot and data, respectively $\mathbf{p}_{\ell,k}$: $C_u \times 1$ orthogonal pilot transmitted by the k^{th} user of the ℓ^{th} cell

• The total transmitted power is $\mu_{\ell,k} = \lambda_{\ell,k}^2 + \rho_{\ell,k}^2$

Superimposed Pilots

• The least squares estimate of the channel

$$\begin{split} \widehat{\mathbf{h}}_{j,\ell,k} &= \mathbf{Y} \left(\lambda_{\ell,k}^2 \; \mathbf{p}_{\ell,k}^H \mathbf{p}_{\ell,k} \right)^{-1} \lambda_{\ell,k} \mathbf{p}_{\ell,k}^* \\ &= \mathbf{h}_{\ell,k} + \frac{1}{C_u \lambda_{\ell,k}} \sum_{n=0}^{L-1} \sum_{p=0}^{K-1} \rho_{n,p} \mathbf{h}_{n,p} \mathbf{x}_{n,p}^T \mathbf{p}_{\ell,k}^* + \frac{1}{C_u \lambda_{n,p}} \mathbf{W} \mathbf{p}_{\ell,k}^* \end{split}$$

• The data, estimated using a matched filter

$$\widetilde{\mathbf{x}}_{j,m}^{T} = \frac{1}{M\rho_{j,m}\beta_{j,m}} \widehat{\mathbf{h}}_{j,m}^{H} \left(\mathbf{Y} - \lambda_{j,m} \widehat{\mathbf{h}}_{j,m} \mathbf{p}_{j,m}^{T}\right)$$
$$\widehat{\mathbf{x}}_{j,m} = \eta \left(\widetilde{\mathbf{x}}_{j,m}\right)$$

 $\eta(\cdot)$: element-by-element hard-decision function that replaces each element of the input vector with the constellation point that is closest in Euclidean distance to that element

Performance

• The MSE of the superimposed pilot based channel estimate

$$\mathrm{MSE}_{j,m}^{\mathrm{SP}} = \frac{1}{C_u \lambda_{j,m}^2} \left(\sum_{k=0}^{K-1} \rho_{j,k}^2 \beta_{j,k} + \sum_{\substack{\ell=0\\ \ell \neq j}}^{L-1} \sum_{k=0}^{K-1} \rho_{\ell,k}^2 \beta_{\ell,k} \right) + \frac{\sigma^2}{C_u \lambda_{j,m}^2}$$

• The SINR after MF

$$\operatorname{SINR}_{j,m}^{\operatorname{SP-ul}} = \frac{\beta_{j,m}^2}{\frac{1}{C_u \lambda_{j,m}^2 \rho_{j,m}^2} \sum_{\ell=0}^{L-1} \sum_{k=0}^{K-1} \rho_{\ell,k}^2 \mu_{\ell,k} \beta_{\ell,k}^2}$$

• For comparison, recall the SINR of a system using conventional pilots

$$\operatorname{SINR}_{j,m}^{\operatorname{CP-ul}} = \frac{\beta_{j,m}^2}{\sum\limits_{\ell \neq j} \beta_{\ell,m}^2}$$

Implications

$$\operatorname{SINR}_{j,m}^{\operatorname{SP-ul}} = \frac{\beta_{j,m}^2}{\frac{1}{C_u \lambda_{j,m}^2 \rho_{j,m}^2} \sum_{\ell=0}^{L-1} \sum_{k=0}^{K-1} \rho_{\ell,k}^2 \mu_{\ell,k} \beta_{\ell,k}^2} \qquad \operatorname{SINR}_{j,m}^{\operatorname{CP-ul}} = \frac{\beta_{j,m}^2}{\sum_{\ell \neq j} \beta_{\ell,m}^2}$$

- Power control is essential for a system with superimposed pilots
- There is a coherence time beyond which a system with superimposed pilots performs better than a system with conventional pilots, i.e., if $\beta_{j,m} = 1$, $\forall m$, $\beta_{\ell,m} = \beta$, $\forall \ell \neq j, m$, and $\rho_{i,m}^2 = \lambda_{i,m}^2, \forall j, m$, then

$$\frac{\mathrm{SINR}_{j,m}^{\mathrm{SP-ul}} > \mathrm{SINR}_{j,m}^{\mathrm{CP-ul}}}{_{\substack{M \to \infty}}} \quad \text{if} \quad C_u > 2K \left(1 + \frac{1}{(L-1)\beta^2}\right)$$

• This dependence on C_u can be reduced by using an iterative data aided scheme.

Iterative Channel Estimation

 Replacing k, ℓ with a single index n such that 0 ≤ n ≤ N − 1, where N ≜ KL, received signal at the jth BS when employing superimposed pilots

$$\mathbf{Y} = \sum_{n=0}^{N-1} \mathbf{h}_n \left(\rho_n \mathbf{x}_n + \lambda_n \mathbf{p}_n \right)^T + \mathbf{W}$$

- Users are arranged in decreasing order of their path-loss coefficients, i.e., $\beta_0 > \beta_1 > \ldots > \beta_{N-1}$
- $U_m^{(i)}$ is the set of users whose channel and data vectors are used in feedback to estimate channel of $m^{\rm th}$ user in $i^{\rm th}$ iteration

Performance

Iterative Channel Estimation

• Set
$$\widehat{\mathbf{h}}_m^{(0)} = \mathbf{0}, \ \forall \ m \text{ and } \widehat{\mathbf{x}}_m^{(0)} = \mathbf{0}, \ \forall \ m = 0, \dots, N-1$$

• Channel estimation step

$$\widehat{\mathbf{h}}_{m}^{(i)} = \frac{1}{C_{u}\lambda_{m}} \left[\mathbf{Y} - \sum_{\substack{n=0\\n \in \mathcal{U}_{m}^{(i)}}}^{m-1} \rho_{n} \widehat{\mathbf{h}}_{n}^{(i)} \left(\widehat{\mathbf{x}}_{n}^{(i)} \right)^{T} - \sum_{\substack{n=m\\n \in \mathcal{U}_{m}^{(i)}}}^{N-1} \rho_{n} \widehat{\mathbf{h}}_{n}^{(i-1)} \left(\widehat{\mathbf{x}}_{n}^{(i-1)} \right)^{T} \right]$$

• Data estimation step

$$\begin{pmatrix} \widetilde{\mathbf{x}}_{m}^{(i)} \end{pmatrix}^{T} = \frac{1}{M\rho_{m}\beta_{m}} \begin{pmatrix} \widehat{\mathbf{h}}_{m}^{(i)} \end{pmatrix}^{H} \begin{pmatrix} \mathbf{Y} - \lambda_{m} \widehat{\mathbf{h}}_{m}^{(i)} \mathbf{p}_{m}^{T} \end{pmatrix}$$
$$\widehat{\mathbf{x}}_{m}^{(i)} = \eta \begin{pmatrix} \widetilde{\mathbf{x}}_{m}^{(i)} \end{pmatrix}$$

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Simulation Setup

- M = 300 antennas
- $C_u = 100$ symbols
- 6 interfering cells (L = 6)
- K = 5 users per cell
- Cell Radius 1km
- Users are equally spaced on a circle of radius 800m

• Path-loss exponent - 3

Iterative Channel Estimation

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SINR Performance



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Performance Benefits

- Superior uplink SINR performance in scenarios with high pilot contamination.
- As $M \to \infty$, optimal values of $\rho_{j,m} \to 0$ and $\lambda_{j,m} \to 1$, resulting in an interference-free downlink among the *KL* users
- Can augment systems with conventional pilots to increase the spectral efficiency

Caveats of Using Superimposed Pilots

- Number of users in all cells combined have to be less than uplink duration, i.e., $KL \leq C_u$
- Is superior to conventional pilots in uplink performance, in scenarios with high pilot contamination
- Uplink performance depends on $\frac{KL}{C}$
- $M \gg LK$ instead of $M \gg K$
- Power control is mandatory
- Coordination between cells is required for assigning pilots and for learning the path-loss coefficients $\beta_{j,\ell,k}$ of all users

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Thank You

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Backup Slides



Choice of $\mathcal{U}_m^{(i)}$ - Assumptions

- $\mathbf{e}_m^{(i)} \triangleq \mathbf{x}_m \widetilde{\mathbf{x}}_m^{(i)}$ is the error in the matched filtered data.
- $\Delta \mathbf{x}_m^{(i)} \triangleq \mathbf{x}_m \widehat{\mathbf{x}}_m^{(i)}$ is the error vector after slicing.
- $\Delta \mathbf{h}_m^{(i)} \triangleq \mathbf{h}_m \widehat{\mathbf{h}}_m^{(i)}$ is the error in the channel estimate.
- Assuming that the elements of e⁽ⁱ⁾_m are i.i.d. circular complex-Gaussian random variables with zero mean and variance I⁽ⁱ⁾_m.
- In deriving \$\mathcal{I}_m^{(i)}\$, the following simplifications have been made to obtain a closed form expression:
 - $\mathbf{e}_{m}^{(i)}$ is independent of \mathbf{x}_{k} and $\mathbf{W}, \forall k, i$.
 - **2** $\Delta \mathbf{x}_{m}^{(i)}$ is independent of \mathbf{x}_{k} , **W**, and \mathbf{h}_{k} , $\forall k, i$.
 - Solution 2 Δx⁽ⁱ⁾_m is independent of Δx^(p)_k, ∀p ≠ i, m ≠ k and the elements of Δx⁽ⁱ⁾_m are i.i.d.

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3 $\Delta \mathbf{h}_m^{(i)}$ is independent of \mathbf{x}_k , \mathbf{W} , and $\Delta \mathbf{x}_k^{(p)}$, $\forall k, p$.

Choice of $\mathcal{U}_m^{(i)}$

• The user set
$$\mathcal{U}_m^{(i)}$$
 is selected as

$$\mathcal{U}_{m}^{\left(i
ight)}=rg\min_{\mathcal{U}\in\mathcal{P}\left(\mathcal{S}
ight)}\left\{ \mathcal{I}_{m}^{\left(i
ight)}\left(\mathcal{U}
ight)
ight\}$$

$${\mathcal I}_m^{(i)}({\mathcal U})$$
 is the interference power

$$\mathcal{I}_m^{(i)}(\mathcal{U}) \approx \frac{1}{\beta_m^2} \left(\frac{1}{M\rho_m^2} \sum_{\substack{n=0\\n\neq m}}^{N-1} \beta_n \beta_m + \frac{\sigma^2 \beta_m}{M\rho_m^2} + \frac{1}{M^2 \rho_m^2} \psi_m^{(i)}(\mathcal{U}) \right)$$

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Choice of $\mathcal{U}_m^{(i)}$

•
$$\psi_m^{(0)} = 0.$$
 For $i \ge 1$

$$\psi_m^{(i)} = \frac{M^2}{C_u \lambda_m^2} \left[\sum_{\substack{n=0\\n \notin \mathcal{U}_m^{(i)}}}^{N-1} \rho_n^2 \beta_n^2 + \sum_{\substack{n=0\\n \in \mathcal{U}_m^{(i)}}}^{m-1} \rho_n^2 \beta_n^2 \alpha_n^{(i)} + \sum_{\substack{n \in \mathcal{U}_m^{(i)}\\n \in \mathcal{U}_m^{(i)}}}^{N-1} \rho_n^2 \beta_n^2 \alpha_n^{(i-1)} + o\left(\frac{N}{M}\right) \right]$$

$$\alpha_k^{(i)} \triangleq \mathbb{E} \left\{ \left| \left[\Delta \mathbf{x}_k^{(i)} \right]_n \right|^2 \right\} \text{ is the variance of } \Delta \mathbf{x}_m^{(i)} \triangleq \mathbf{x}_m - \widehat{\mathbf{x}}_m^{(i)}$$
• For a P-QAM constellation

$$\alpha_k^{(i)} = \begin{cases} \frac{24}{\sqrt{P}(\sqrt{P}+1)} Q\left(\sqrt{\frac{i}{(P-1)}}\right), & i \ge 1\\ 1, & i = 0 \end{cases}$$

• The optimal set is given as

$$\begin{aligned} \mathcal{U}_m^{(i)} &= \left\{ k \in \mathbb{N} \left| \alpha_k^{(i)} < \gamma_k^{(i)} \text{ when } k < m \right. \right. \\ &\quad \text{and } \alpha_k^{(i-1)} < \gamma_k^{(i-1)} \text{ when } k \ge m \right\} \\ \gamma_k^{(i)} &\triangleq \frac{\left\{ \beta_k^2 + \frac{1}{M} \sum_{n=0}^{N-1} \beta_k \beta_n - \frac{\psi_k^{(i)} \Big|_{k \in \mathcal{U}_m^{(i)}}}{M^2} \right\}}{\left\{ \beta_k^2 + \frac{1}{M} \sum_{n=0}^{N-1} \beta_k \beta_n + \frac{\psi_k^{(i)} \Big|_{k \in \mathcal{U}_m^{(i)}}}{M^2} \right\}} \end{aligned}$$

• Alternatively, a fixed conservative threshold can be chosen as

$$\mathcal{U}_{ ext{fixed}} = \left\{ m \in \mathbb{N} \left| \mathcal{I}_m^{(2)}\left(\{m\}
ight) < \mathcal{I}_m^{(2)}\left(arnothing
ight) = \mathcal{I}_m^{(1)}\left(arnothing
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