## Analysis of coprime arrays on moving platform

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## Outline

(1) Introduction
(2) Problem formulation
(3) Difference co-array
(4) Simulation results
(5) Conclusion

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(1) Introduction

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## DOA estiamtion based on a fixed coprime array


${ }^{1}$ Chun-Lin Liu and P. P. Vaidyanathan, "Robustness of coarrays of sparse arrays to sensor failures," in Proc. IEEE Int. Conf. Acoust., Speech, and Sig. Proc., Calgary, AB, Canada, April, 2017, pp. 3231 - 3235.

## Coprime array on a moving platform



[^0]
## Coprime array moving half-wavelength

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Difference coarray after moving half-wavelength
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- Advantages:
(1) Easy to be applied in practice
(2) Such motion can fill most, if not all of the holes, and as such significantly increases both the contiguous and unique lags


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## Passive synthetic aperture



The output of the receive array, at time $t$, is expressed as

$$
\mathbf{x}(t)=\sum_{q=1}^{Q} s_{q}(t) \exp \left(-j v t \kappa_{q}\right) \mathbf{a}\left(\theta_{q}\right)+\boldsymbol{\varepsilon}(t)=\mathbf{A} \mathbf{s}(t)+\boldsymbol{\varepsilon}(t)
$$

where $v$ is the velocity of platform, $\mathbf{a}\left(\theta_{q}\right)=\left[1, \exp \left(-j d_{2} \kappa_{q}\right), \cdots, \quad \exp \left(-j d_{L} \kappa_{q}\right]^{T}, \kappa_{q}=2 \pi \sin \left(\theta_{q}\right) / \lambda\right.$, $\mathbf{s}(t)=\left[s_{1}(t) \exp \left(-j v t \kappa_{1}\right), s_{2}(t) \exp \left(-j v t \kappa_{2}\right), \cdots, s_{Q}(t) \exp \left(-j v t \kappa_{Q}\right)\right]^{T}$, $\mathbf{A}=\left[\mathbf{a}\left(\theta_{1}\right), \mathbf{a}\left(\theta_{2}\right), \cdots, \mathbf{a}\left(\theta_{Q}\right)\right], \boldsymbol{\varepsilon}(t)$ is zero-mean complex additive white Gaussian noise vector.

## Passive synthetic aperture

At time $t+\tau$,

$$
\begin{aligned}
\mathbf{x}(t+\tau) & =\sum_{q=1}^{Q} s_{q}(t+\tau) \exp \left(-j v t \kappa_{q}\right) \exp \left(-j v \tau \kappa_{q}\right) \mathbf{a}\left(\theta_{q}\right)+\varepsilon(t+\tau) \\
& =\operatorname{Bs}(t+\tau)+\varepsilon(t+\tau) .
\end{aligned}
$$

where

$$
\begin{aligned}
\mathbf{B}= & {\left[\mathbf{b}\left(\theta_{1}\right), \mathbf{b}\left(\theta_{1}\right), \cdots, \mathbf{b}\left(\theta_{Q}\right)\right], } \\
\mathbf{b}\left(\theta_{q}\right)= & \exp \left(-j v \tau \kappa_{q}\right) \mathbf{a}\left(\theta_{q}\right), \\
= & {\left[\exp \left(-j v \tau \kappa_{q}\right), \exp \left(-j\left(v \tau+d_{2}\right) \kappa_{q}\right), \cdots, \exp \left(-j\left(v \tau+d_{L}\right) \kappa_{q}\right)\right]^{T}, } \\
\mathbf{s}(t+\tau)= & {\left[s_{1}(t+\tau) \exp \left(-j v t \kappa_{1}\right), s_{2}(t+\tau) \exp \left(-j v t \kappa_{2}\right), \cdots,\right.} \\
& \left.s_{Q}(t+\tau) \exp \left(-j v t \kappa_{Q}\right)\right]^{T} .
\end{aligned}
$$

## Passive synthetic aperture

For narrowband signals and $v \tau=d=\lambda / 2$,

$$
\begin{gathered}
\mathbf{x}(t+\tau)=\exp (j 2 \pi f \tau) \mathbf{B} \mathbf{s}(t)+\varepsilon(t+\tau) \\
\mathbf{b}\left(\theta_{q}\right)=\left[\exp \left(-j d \kappa_{q}\right), \exp \left(-j\left(d+d_{2}\right) \kappa_{q}\right), \cdots, \exp \left(-j\left(d+d_{L}\right) \kappa_{q}\right)\right]^{T}
\end{gathered}
$$

Compensating for the phase correction factor $\exp (j 2 \pi f \tau)$ (See Ref.[3]),

$$
\tilde{\mathbf{x}}(t+\tau)=\mathbf{x}(t+\tau) \exp (-j 2 \pi f \tau)=\mathbf{B} \mathbf{s}(t)+\tilde{\varepsilon}(t+\tau)
$$

where $\tilde{\varepsilon}(t+\tau)=\exp (-j 2 \pi f \tau) \varepsilon(t+\tau)$.
The output of the synthetic array is

$$
\mathbf{y}(t)=\left[\begin{array}{c}
\mathbf{x}(t) \\
\tilde{\mathbf{x}}(t+\tau)
\end{array}\right]=\mathbf{A}_{s} \mathbf{s}(t)+\left[\begin{array}{c}
\varepsilon(t) \\
\tilde{\varepsilon}(t+\tau)
\end{array}\right]
$$

where $\mathbf{A}_{s}=\left[\mathbf{a}_{s}\left(\theta_{1}\right), \mathbf{a}_{s}\left(\theta_{2}\right), \cdots, \mathbf{a}_{s}\left(\theta_{Q}\right)\right], \mathbf{a}_{s}\left(\theta_{q}\right)=\left[\mathbf{a}^{T}\left(\theta_{q}\right), \mathbf{b}^{T}\left(\theta_{q}\right)\right]^{T}$.

[^1]
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## Difference co-array



The subarrays of the original and synthetic coprime array $(M<N)$.

## Difference co-array

The set formed from the difference co-array corresponding to the synthetic array is given as

$$
\mathbb{S}_{\mathrm{c}}=\mathbb{S}_{12} \cup \mathbb{S}_{34} \cup \mathbb{S}_{13} \cup \mathbb{S}_{24} \cup \mathbb{S}_{14} \cup \mathbb{S}_{23}
$$

where $\mathbb{S}_{12}=\left\{M k_{1}-N k_{2}\right\} \cup\left\{N k_{2}-M k_{1}\right\}$.
Lemma 1: For sets $\mathbb{S}_{13}, \mathbb{S}_{24}, \mathbb{S}_{14}$ and $\mathbb{S}_{23}$ defined above, $\mathbb{S}_{13} \cup \mathbb{S}_{24} \cup \mathbb{S}_{14} \cup 23=\mathbb{S}_{14} \cup \mathbb{S}_{23}$.
proof: See Ref. [4].
Utilizing Lemma 1 and $\mathbb{S}_{12} \cup \mathbb{S}_{34}=\mathbb{S}_{12}, \mathbb{S}_{\mathrm{c}}$ is simplified as

$$
\mathbb{S}_{\mathrm{c}}=\mathbb{S}_{12} \cup \mathbb{S}_{14} \cup \mathbb{S}_{23}
$$

Because

$$
\mathbb{S}_{14} \cup \mathbb{S}_{23}=\tilde{\mathbb{S}}_{14} \cup \tilde{\mathbb{S}}_{23}
$$

[^2]
## Difference co-array

where

$$
\begin{aligned}
& \tilde{\mathbb{S}}_{14}=\left\{M k_{1}-N k_{4}-1\right\} \cup\left\{N k_{2}-M k_{3}-1\right\} \\
& \tilde{\mathbb{S}}_{23}=\left\{N k_{4}-M k_{1}+1\right\} \cup\left\{M k_{3}-N k_{2}+1\right\}
\end{aligned}
$$

Because $k_{1}, k_{3} \in[0, N-1], k_{2}, k_{4} \in[0, M-1], \tilde{\mathbb{S}}_{14}$ and $\tilde{\mathbb{S}}_{23}$ are equivalent to the following equations.

$$
\begin{aligned}
& \mathbb{S}_{12}^{\mathrm{L}}=\left\{M k_{1}-N k_{2}-1\right\} \cup\left\{N k_{2}-M k_{1}-1\right\}, \\
& \mathbb{S}_{12}^{\mathrm{R}}=\left\{N k_{2}-M k_{1}+1\right\} \cup\left\{M k_{1}-N k_{2}+1\right\}
\end{aligned}
$$

Then,

$$
\mathbb{S}_{\mathrm{c}}=\mathbb{S}_{12} \cup \mathbb{S}_{12}^{\mathrm{L}} \cup \mathbb{S}_{12}^{\mathrm{R}}
$$

The difference co-array of the synthetic array consists of the difference co-array of the original array and its unit-lag shifted versions in the direction and opposite to direction of motion.

## Difference co-array:Example

Assume $M=4, N=5$, then

$$
\mathbb{S}_{12}=\{0, \pm 1, \cdots, \pm 8, \pm 10, \pm 11, \pm 12, \pm 15, \pm 16\}
$$

The number of unique lags: 27, the number of holes: 6 After array motion,

$$
\begin{gathered}
\mathbb{S}_{12}^{\mathrm{L}}=\{0, \pm 1, \cdots, \pm 7,-8, \pm 9,10, \pm 11,-12,-13,14,15,-16,-17\} \\
\mathbb{S}_{12}^{\mathrm{R}}=\{0, \pm 1, \cdots, \pm 7,8, \pm 9,-10, \pm 11,12,13,-14,-15,16,17\} \\
\mathbb{S}_{\mathrm{c}}=\{0, \pm 1, \cdots, \pm 17\}
\end{gathered}
$$

The number of unique lags: 35 , the number of holes: 0

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## Simulation results



(b)


| -23 | -13 | -13 | -8 | -3 | 2 | 7 | 12 | 17 | 22 | 27 | 32 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| -33 | -28 | -23 | -15 |  |  |  |  |  |  |  |  |

Difference co-array for $M=4, N=9$. (a) $\mathbb{S}_{12}$; (b) $\mathbb{S}_{12}^{\mathrm{L}} ;$ (c) $\mathbb{S}_{12}^{\mathrm{R}}$; (d) $\mathbb{S}_{\mathrm{c}}$.(•: lags; $\times$ : holes.)

## Simulation results



Spatial spectra estimation using the LASSO. (a) $M=4, N=9, Q=26$ for the original array; (b) $M=4, N=9, Q=26$ for the synthetic array; (c) $M=6, N=7, Q=29$ for the original array; (d) $M=6, N=7, Q=29$ for the synthetic array; (e) $M=5, N=8, Q=31$ for the original array; (f) $M=5, N=8, Q=31$ for the synthetic array.

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## Conclusion

- Analysis of coprime arrays on moving platform
(1) Passive synthetic aperture for moving half-wavelength
(2) Composition of the difference co-array for synthetic array
(3) Camprison of difference co-array corresponding to the oringinal and synthetic array


## THANK YOU!


[^0]:    ${ }^{2}$ J. Ramirez and J. L. Krolik, "Synthetic aperture processing for passive co-prime linear sensor arrays," Digital Signal Process., vol. 61, pp. 62-75, 2017.

[^1]:    ${ }^{3}$ S. Stergios and E. J. Sullivan, "Extended towed array processing by an overlap correlator," J. Acoust. Soc.America, vol. 86, no. 1, pp. 158-171, 1989.

[^2]:    ${ }^{4}$ G. Qin, M. G. Amin, and Y. D. Zhang, "DOA estimation exploiting sparse array motions," IEEE Trans. Signal Process., vol. 67, no. 11, pp. 3013-3027, June 2019.

