IEEE GLOBALSIP 2019

3-D MIMO-SAR IMAGING USING MULTI-CHIP CASCADED MMWAVE SENSORS

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November, 2019





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MOTIVATION

3-D holographic millimeter-wave (mmWave) imaging technology has a wide range of applications including security, packaging, automotive, medical, aerial imaging, etc.









Problems:

- Achieve high-resolution with as few antenna elements as possible
- Multiple-input multiple-output (MIMO) radars increase the degrees of freedom
- The complexity of extremely dense transceiver electronics limits the use of MIMO-only solutions

Proposed Solution:

- Hybrid concepts combining synthetic aperture radar (SAR) techniques and MIMO arrays:
 - Present a good compromise to achieve short data acquisition time and low-complexity.





RESEARCH OBJECTIVES

Use 4-chip cascaded radar board in near-field 3-D holographic MIMO-SAR imaging



Synthetic Aperture

Develop a novel and complete signal processing chain from data capture to image reconstruction



- Design a system-level imager prototype to demonstrate the proof-of-concept
 - Understand the challenges
 - Develop toolset and algorithms
 - Work towards real-world applications



DATA CAPTURE

Problem:

Develop a system to capture real MIMO-SAR data for demonstration



Tasks:

Develop familiarity with 4-chip cascaded mmWave board and data collection toolset



- Build a 2-D mechanical scanner to synthesize a larger aperture
- Integrate 4-chip cascaded board to the scanner
- Develop a complete software package in MATLAB to capture MIMO-SAR data

DESIGNING A NEW IMAGER PROTOTYPE

• A mechanical scanner has been designed, multi-chip cascaded sensors has been integrated



- Scanning area:
 - 1000 mm by 1000 mm
- Maximum speed:
 - 400 mm/s at both axes



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- The mmWave sensor hardware stack is integrated:
 - 1) 4-Chip cascaded board
 - 2) TSW14J56

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3) Mother Board + Adapter Board

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4-CHIP CASCADED BASED MIMO-SAR SYSTEM

• Data collection is performed by moving the **4-chip cascaded MIMO radar** along a trajectory in x - y plane



HIGH-LEVEL SYSTEM ARCHITECTURE

MATLAB GUI is developed to control the platform and the radar



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DATA CAPTURE GUI – SAR SCENARIO GENERATION SCREEN

v2.0.1

承 SAR Imaging Toolbox

SAR Imaging Toolbox

SAR configuration page:

Fully automated process



tialization SAR Configuration		
SAR Configuration (Con	tinuous Horizontal Scan)	SAR Co
Platform Speed (mm/s)	10 Horizontal Step Size (mm) 0.0000	# of Sam
# of Samples/Channel	0 Update Rectangular	Horizonta
Vertical Scan Size (mm)	0.0000 Params Scan Mode	Vertical S
Frame Periodicity (ms)	100 Raster	
Number of Frames	1 Recommend	Number
Time Before Capturing (s)	-1 ((-) if Radar Trigger First)	Horizont
Horizontal Scan Size (mm)	0	Number
Number of Steps at Vertical S	Scan 1	Vertical S
Vertical Step Size (mm)	83.859	
	Blank between values. Ex: 0.9 1.9	Wait Tim
Reconfigure Radar	Start Capturing Stop Capturing	
Total Movement	SAR Status	Total Mo

onfiguration (Discrete Horizontal Scan) Number of Frames ples/Channel 0 1 al Scan Size (mm) 0.0000 Update Params 0.0000 Scan Size (mm) * of Steps at Horizontal Scan al Step Size (mm) 0 * of Steps at Vertical Scan 0 Step Size (mm) Blank between values. Ex: 0.9 1.9 1 e Between Steps (s) Reconfigure Start Capturing Radar SAR Status vement

Emergency Stop Scanner





NEAR-FIELD MIMO ARRAY CALIBRATION

Problem:

- In a practical system, measurement errors in the MIMO array arise due to antenna gain and phase mismatches (because of imperfections in antenna layouts on the board: RF delays, etc.)
- Uncorrected phase error leads to unacceptable defocused blur and range shift in images



Tasks:

Propose a calibration method based on the ideal backscattered signal model from a reference point target

Reference beat signal

Solution:

• The uncalibrated measured beat signal: $\tilde{s}_{\ell}(t) = a_{\ell}e^{j2\pi(f_0+Kt)(\tau_i+\tau_{\ell})}$

• Rewrite: $\tilde{s}_{\ell}(t) = a_{\ell} e^{j\psi_i} e^{j2\pi f_i t} s_{\ell}(t)$

 η_{ℓ}

The total round-trip delay

• Estimate: $f_i = K \tau_i$ and η_ℓ



NEAR-FIELD MIMO ARRAY CALIBRATION

- The accuracy of the presented approach depends on the reference beat signal
- The target needs to be precisely positioned in calibrating process!



Given the measurements $\tilde{s}_{\ell}(t)$, the calibration error signal can be computed by:

 $w_{\ell}(t) = \tilde{s}_{\ell}(t)s_{\ell}^{*}(t) = \eta_{\ell}e^{j2\pi f_{i}t}$ Simple demodulation process

Parameter estimation problem of a single-frequency complex tone

• Estimate:
$$\hat{f}_i = \arg \max \sum_{\langle \ell \rangle} |W_\ell(f)|^2$$
 $W_\ell(f) = \int_0^T w_\ell(t) e^{-j2\pi ft} dt$

• Estimate η_{ℓ} by plugging the estimate \hat{f}_i into $w_{\ell}(t)$

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NEAR-FIELD MIMO ARRAY CALIBRATION

- Simulate the backscattered signal for each channel based on the exact target location
- Calculate complex calibration data η_{ℓ} and range (beat frequency) offset $K\tau_i$





3-D MIMO SAR IMAGE RECONSTRUCTION IN NEAR-FIELD

Problems:

- Near-field: the plane-wave assumption is invalid and the spherical wave model has to be used
- MIMO sampling: different trajectories of the incident and reflected electric fields has to be considered
- Proposed algorithm has to be computationally efficient



Tasks:

- Develop low-complexity and novel image reconstruction algorithms for MIMO-SAR configuration
- Validate the algorithms using simulation environment
- Demonstrate the real high-resolution 3-D imaging results using prototyped imager





CONVENTIONAL SAR SYSTEM CONFIGURATION

• The complex "beat" signal: $s(t) = \sigma \exp(j2\pi(f_0\tau + K\tau t - 0.5K\tau^2))$ Delay: $\tau = 2R/c$ Gain: $\sigma = p/R^2$



MULTISTATIC EFFECT IN NEAR-FIELD

- Midpoint approximation: $|\mathbf{r}_u \mathbf{r}_v| \le \sqrt{4\alpha\lambda R}$ for a small α doesn't hold in near field
- The total round-trip distance for the ℓ -th virtual channel



$$R_{u} + R_{v} = \sqrt{(x_{u} - x)^{2} + (y_{u} - y)^{2} + z_{0}^{2}} + \sqrt{(x_{v} - x)^{2} + (y_{v} - y)^{2} + z_{0}^{2}}$$

• Complex solution: $s_{\ell}(x', y', k) = \iiint p(x, y, z) \frac{e^{jk(R_{u} + R_{v})}}{R_{u}R_{v}} dxdydz$
• Taylor expansion
 $R_{\ell} \approx 2R + \frac{(d_{\ell}^{x})^{2} + (d_{\ell}^{y})^{2}}{4R} - \frac{((x - x')d_{\ell}^{x} + (y - y')d_{\ell}^{y})^{2}}{4R^{3}}$

- Multistatic to monostatic phase correction is proposed
- The received signal by the ℓ th hypothetical monostatic element:

$$s_{\ell}(k) \approx p \frac{\exp(jkR_{\ell})}{R^2} \approx s(k)e^{j\phi_{\ell}(k)} \quad \phi_{\ell}(k) \approx k \frac{(d_{\ell}^{x})^2 + (d_{\ell}^{y})^2}{4z_0}$$

 $4R^{3}$

3-D MONOSTATIC IMAGE RECONSTRUCTION

• Recall the beat signal: $s(t) = \sigma \exp(j2\pi(f_0\tau + K\tau t - (0.5K\tau^2))) \rightarrow \mathbb{R}$ Residual video phase (RVP) is negligible

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Rewrite in the wavenumber domain as:

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$$s(k) = p \frac{\exp(j2kR)}{R^2} \qquad R = \sqrt{(x'-x)^2 + (y'-y)^2 + z^2}$$

The received monostatic backscattered data from a 3-D target

$$s(x',y',k) = \iiint p(x,y,z) \frac{e^{j2kR}}{R} dxdydz \qquad R^{-2} \approx (z_0R)^{-1}$$

Representation of a spherical wave as a superposition of plane waves

$$\frac{e^{j2kR}}{R} = \frac{j}{2\pi} \iint \frac{e^{j(k_x(x'-x)+k_y(y'-y)+k_zz)}}{k_z} dk_x dk_y \qquad k_z = \sqrt{4k^2 - k_x^2 - k_y^2} k_x, k_y$$

- The final 3-D image in spectral domain: $P(k_x, k_y, k_z) = \hat{S}(k_x, k_y, k_z)$
- Interpolation step with k_z : $k_z S(k_x, k_y, k) \rightarrow \hat{S}(k_x, k_y, k_z)$

s(x, y, k)

FFT2

FFT

IFFT2

p(x, y, z)

x, y

 k_z

 $k \rightarrow k_z$

 $S(k_x, k_y, k)$

 $\hat{S}(k_x,k_y,z)$

| Interpolate | $\hat{S}(k_x, k_y, k_z)$

3-D image formation

PERFORMANCE PARAMETERS – MONOSTATIC ANALYSIS

 $\frac{\mathbf{r}_u + \mathbf{r}_v}{2}$

- Data bandwidth depends on:
 - Wavelength (λ)
 - Size of the radar aperture (D_x^S, D_y^S)
 - Size of the target aperture (D_x^T, D_y^T)
 - Distance between the target and the radar apertures (z_0)
- The total bandwidth:

$$\Delta k_x \approx 2k_x^{\rm bw} \approx 4k\sin\theta_x$$

The cross range resolution:

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$$\delta_{\chi} \approx \frac{2\pi}{\Delta k_{\chi}} = \approx \frac{\lambda z_0}{2D_{\chi}^{S}}$$

Phase center concept:

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Target Aperture
Data Spectrum
Data Spectrum

$$j_{x}$$

 j_{x}
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MIMO-SAR SIMULATIONS – MULTISTATIC EFFECT

- Simulated uniform array with 86 non-overlapped virtual elements, two captures in vertical axis
 - Multistatic data capture
 - Monostatic image reconstruction
- After multistatic to monostatic phase correction

 $\Delta_x = \lambda/4$ $D_x^S \approx 180 \text{ mm}$



Monostatic image reconstruction

MIMO-SAR EXPERIMENTS

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Image reconstruction using 4-Chip Cascaded Radar (86 virtual channels)



 $D_x^S \approx D_y^S \approx 500 \text{ mm}$ $\Delta_x = 1 \text{ mm} \approx \lambda/4$ $\Delta_{\gamma} = 83.859 \text{ mm} = 86\lambda/4$ $z_0 \approx 400 \text{ mm}$ ACE SRC Semiconductor Research

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IMAGING OF CONCEALED PHYSICAL OBJECTS

Experiment with scissors concealed in a cardboard box

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Image reconstruction using 4-Chip Cascaded Radar (86 virtual channels)



 $D_x^S \approx D_y^S \approx 500 \text{ mm}$ $\Delta_x = 1 \text{ mm} \approx \lambda/4$ $\Delta_y = 83.859 \text{ mm} = 86\lambda/4$ $z_0 \approx 250 \text{ mm}$

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SUMMARY

- Developed a near-field mmWave imaging prototype utilizing 4-chip cascaded board in SAR configuration
- Proposed a complete signal processing toolset from data capture to image formation
 - Developed a fully automated data capture software
 - Presented a calibration method for MIMO arrays
 - Proposed an efficient image reconstruction method based on monostatic reconstruction framework
- Investigated the design considerations including the spatial sampling criteria, and image resolution.
- Demonstrated the real imaging results using prototyped imager (1st image of the 4-chip cascaded board)
- Investigated the future improvements and challenges to work towards real-world applications









THANK YOU QUESTIONS?







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