

Robust Fundamental Frequency Estimation in Coloured Noise

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- ▶ **Fundamental frequency** (f_0) is important in several applications: music transcription, audio coding, and speech decomposition.
- ▶ Correlation-based methods (Ghahremani, 2014) (Talkin, 1995) suffer from octave errors, are not robust to the noise and have time-frequency resolution problems (Nielsen, 2017)
- ▶ Parametric methods (Christensen, 2009) (Shi, 2019) also present the problem of octave errors in coloured noise conditions, as they are typically derived under a **WGN** assumption.
- ▶ The spectral shape of the noise needs to be considered.
- ▶ In that case, the ML solution is not longer the same as the NLS one (Parra, 2001).
- ▶ We propose 2 approaches to estimate the speech signal and noise parameters in an **iterative** manner to improve accuracy.

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Vs. typical mixed spectrum estimation



- ▶ Jointly estimating the signal and noise parameters is a nonlinear multidimensional optimization problem (Kay, 1994).
- ▶ In the case of **independent** sinusoids:
 1. The sinusoidal and noise AR parameters are estimated separately in two different steps. This, in only one iteration (Li, 1996).
 2. In the first step, the different sinusoids are estimated using iterative algorithms such as matching pursuit and RELAX.
 3. Despite the noise spectral shape, the one-iteration solution still results in statistically efficient estimates (Stoica, 1997).
- ▶ This is not possible for **harmonically related** sinusoids. Exploiting the harmonic structure (Elvander, 2020) is necessary to allow for improved estimates.
- ▶ Ignoring the noise AR structure increases the risk of selecting a wrong peak (octave errors).

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Harmonic signal model:

$$x(n) = s(n) + e(n) = \sum_{l=-L}^L \alpha_l e^{j\omega_l n} + e(n), \quad (1)$$

where

- ▶ L : number of harmonics,
- ▶ $\alpha_0 = 0, \alpha_l = \alpha_{-l}^*$: complex amplitude of the l^{th} harmonic,
- ▶ $\omega_l = \omega_0 l$, i.e., sinusoids are harmonically related.

Coloured noise modelled as an **AR** process:

$$e(n) = - \sum_{i=1}^P a_i e(n-i) + w(n), \quad (2)$$

where

- ▶ $\{a_i\}_{i=1}^P$: noise AR parameters,
- ▶ $w(n)$ is a driving WGN process $\mathcal{N}(0, \sigma_w^2)$.

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A segment of N samples is expressed as

$$\mathbf{x} = \mathbf{s} + \mathbf{e} = \mathbf{Z}_L(\omega_0)\boldsymbol{\alpha} + \mathbf{e}, \quad (3)$$

where $\mathbf{x} = [x(0) \ \dots \ x(N-1)]^T$, and \mathbf{e} is similarly defined, and

$$\mathbf{Z}_L(\omega_0) = [\mathbf{z}(\omega_0) \ \mathbf{z}^*(\omega_0) \ \dots \ \mathbf{z}(\omega_0 L) \ \mathbf{z}^*(\omega_0 L)], \quad (4)$$

$$\mathbf{z}(\omega) = [1 \ e^{j\omega} \ \dots \ e^{j\omega(N-1)}]^T, \quad (5)$$

$$\boldsymbol{\alpha} = \frac{1}{2} [A_1 e^{j\psi_1} \ \dots \ A_L e^{j\psi_L} \ A_L e^{-j\psi_L}]^T. \quad (6)$$

When $L \neq 0$, the model describes a voiced speech segment, otherwise is a not-voiced one (unvoiced or speech pauses).

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Pre-whitening+NLS pitch estimates



- ▶ The signal is first **pre-whitened**. Otherwise, we observe a high number of gross errors (> 70%) and voicing detection errors (> 45%), especially at low SNRs (Esquivel, 2019).

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- ▶ The signal is first **pre-whitened**. Otherwise, we observe a high number of gross errors ($> 70\%$) and voicing detection errors ($> 45\%$), especially at low SNRs (Esquivel, 2019).
- ▶ An AR spectrum is fitted to the estimated noise PSD $\Phi_e(k)$, $k = 0, 1, \dots, N - 1$ (e.g. MS, MMSE-SPP or parametric NMF).

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- ▶ The time-varying **AR pre-whitener** $A(\omega) = 1 + \sum_{i=1}^P a_i e^{-j\omega i}$ is applied (it does not modify ω_0).

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- ▶ The time-varying **AR pre-whitener** $A(\omega) = 1 + \sum_{i=1}^P a_i e^{-j\omega i}$ is applied (it does not modify ω_0).
- ▶ As the noise is now closer to white, an initial ω_0 is obtained from the **nonlinear least squares (NLS)** estimator

$$\hat{\omega}_0 = \arg \max_{\omega_0} \mathbf{x}^T \mathbf{Z}_L(\omega_0) \left[\mathbf{Z}_L^H(\omega_0) \mathbf{Z}_L(\omega_0) \right]^{-1} \mathbf{Z}_L^H(\omega_0) \mathbf{x}, \quad (7)$$

which can be solved in a rapid order-recursive manner (Nielsen, 2016). L is selected from model selection criteria (e.g BIC).

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First approach: Approximate Maximum Likelihood (ML)



- ▶ Using the harmonic structure, for a given $\hat{\omega}_0$, the LS amplitudes estimates are

$$\hat{\alpha} = [\mathbf{Z}_L^H(\hat{\omega}_0)\mathbf{Z}_L(\hat{\omega}_0)]^{-1}\mathbf{Z}_L^H(\hat{\omega}_0)\mathbf{x}. \quad (8)$$

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- ▶ The residual noise may be estimated as $\hat{\mathbf{e}} = \mathbf{x} - \mathbf{Z}_L(\hat{\omega}_0)\hat{\alpha}$.

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- ▶ The residual noise may be estimated as $\hat{\mathbf{e}} = \mathbf{x} - \mathbf{Z}_L(\hat{\omega}_0)\hat{\alpha}$.
- ▶ Reestimate $\{\hat{a}_i\}_{i=1}^P$ using the AR modeling *autocorrelation method* → form a new **pre-whitened** signal vector → **new** $\hat{\omega}_0$ and \hat{L} (rapid implementation (Nielsen, 2017)).

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- ▶ This is repeated until **convergence**, when either the NLS cost function (7) between 2 consecutive **iterations** is below a threshold value or a maximum number of **iterations** is reached.

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- ▶ This is repeated until **convergence**, when either the NLS cost function (7) between 2 consecutive **iterations** is below a threshold value or a maximum number of **iterations** is reached.
- ▶ When $\hat{L} = 0$, $\hat{\mathbf{e}} = \mathbf{x}$. If in the next iteration the segment is still detected as not-voiced, the process is stopped for that segment.

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Second approach: based on LCMV filtering



- ▶ The speech vector can be estimated by linear filtering as

$$\hat{\mathbf{s}} = \mathbf{H}^T \mathbf{x} = \mathbf{H}^T \mathbf{Z}_L(\hat{\omega}_0) \boldsymbol{\alpha} + \mathbf{H}^T \mathbf{e}. \quad (9)$$

- ▶ To minimize the residual noise power and avoid distortion,

$$\min_{\mathbf{H}} \text{Tr} \left\{ \mathbf{H}^T \mathbf{R}_e \mathbf{H} \right\} \quad \text{s.t.} \quad \mathbf{H}^T \mathbf{Z}_L(\hat{\omega}_0) = \mathbf{Z}_L(\hat{\omega}_0). \quad (10)$$

- ▶ The solution of this optimization problem is

$$\mathbf{H}_{\text{LCMV}} = \mathbf{R}_e^{-1} \mathbf{Z}_L(\hat{\omega}_0) \left(\mathbf{Z}_L^H(\hat{\omega}_0) \mathbf{R}_e^{-1} \mathbf{Z}_L(\hat{\omega}_0) \right)^{-1} \mathbf{Z}_L^H(\hat{\omega}_0) \quad (11)$$

- ▶ From the GS formula, $\mathbf{R}_e^{-1} = \frac{1}{\sigma_w^2} \{ \mathbf{L}_1^T \mathbf{L}_1 - \mathbf{L}_2^T \mathbf{L}_2 \}$, where \mathbf{L}_1 and \mathbf{L}_2 are upper triangular Toeplitz matrices whose first rows are $[1 \ a_1 \ \dots \ a_P \ 0]$ and $[0 \ a_P \ \dots \ a_1]$, respectively.
- ▶ Similar to the approximate ML, iteration between harmonic signal and AR parameter estimates until convergence.

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- ▶ Keele DB has ground truth, which is an **estimate** from an autocorrelation method for segments of length 26.5 ms.

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- ▶ Keele DB has ground truth, which is an **estimate** from an autocorrelation method for segments of length 26.5 ms.
- ▶ Comparison to non-parametric methods: **YIN, RAPT and Cepstrum** based, which have a final step of refinement (e.g. median filter or dynamic programming)



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- ▶ For the proposed, search in [60,400] Hz, with a maximum of $L = 27$ harmonics.



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- ▶ Added noise from NOISEX-92 DB: babble, factory and F-16, @iSNR of -5, 5 and 15 dB.

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- ▶ Added noise from NOISEX-92 DB: babble, factory and F-16, @iSNR of -5, 5 and 15 dB.
- ▶ 6 Monte Carlo simulations per Keele file (10 files) at each iSNR.

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- ▶ 6 Monte Carlo simulations per Keele file (10 files) at each iSNR.
- ▶ The initial applied pre-whitener is based on a **parametric NMF** noise PSD Φ_e estimate (Esquivel, 2019). AR order $P = 25$.

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- ▶ For this, a spectral basis matrix of typical speech and noise spectral envelopes was required and trained offline.

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$$GER = \frac{N_g}{N_{VV}} \times 100\%, \quad (12)$$

$$VDE = \frac{N_{VU} + N_{UV}}{N} \times 100\%, \quad (13)$$

$$FFE = \frac{N_{VU} + N_{UV} + N_g}{N} \times 100\%, \quad (14)$$

- ▶ N_{VV} : segments which are voiced by both the ground truth and the estimated f_0 .
- ▶ N_g : total of the N_{VV} frames with gross relative difference (>20%).
- ▶ N_{VU} : segments which are voiced but misdetected as not-voiced.
- ▶ N_{UV} : segments which are not-voiced but misdetected as voiced.
- ▶ N : total number of segments of an excerpt.

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Example of f_0 estimates of male excerpt @10 dB in F-16 noise

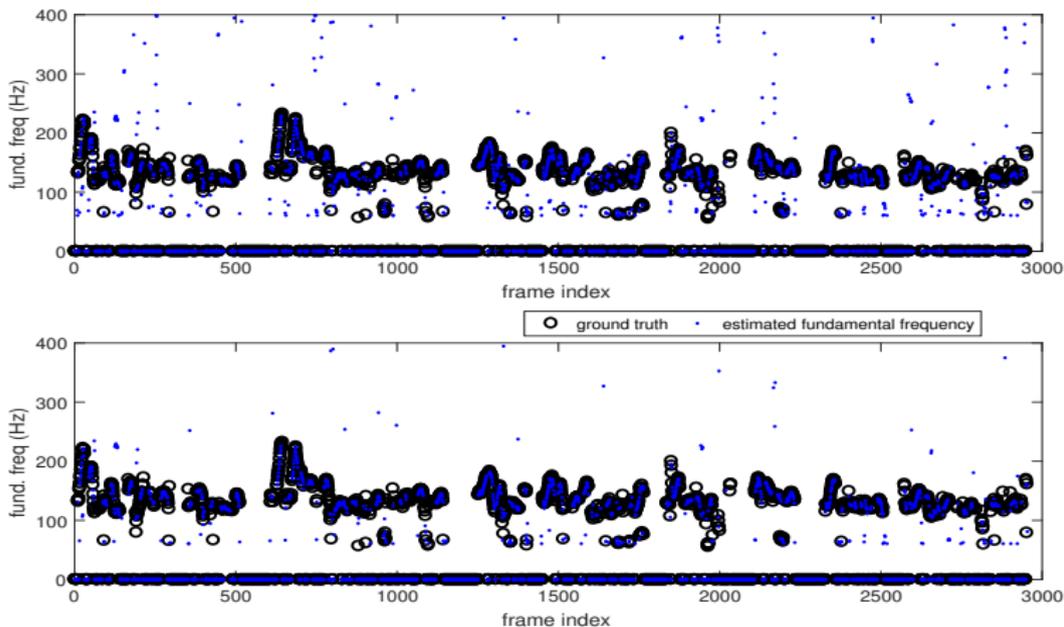


Figure: f_0 ground truth and estimates without iteration (top) and from approx-ML approach (bottom).

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Example of f_0 estimates of female excerpt @15 dB in factory noise

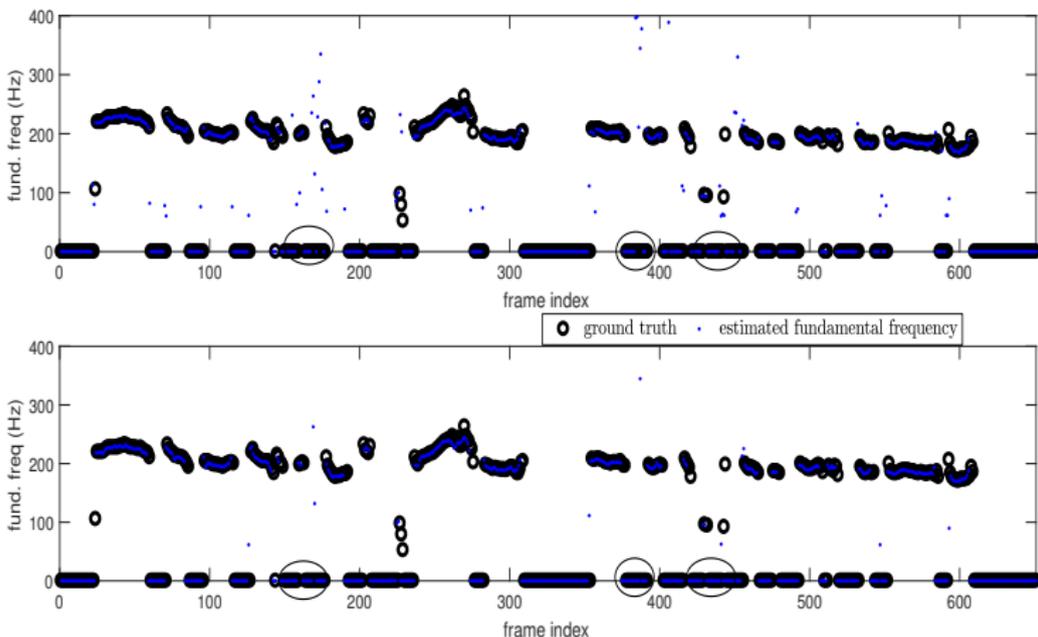


Figure: f_0 ground truth and estimates without iteration (top) and from iterative LCMV filtering approach (bottom).

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Evaluation of parametric and non-parametric estimators

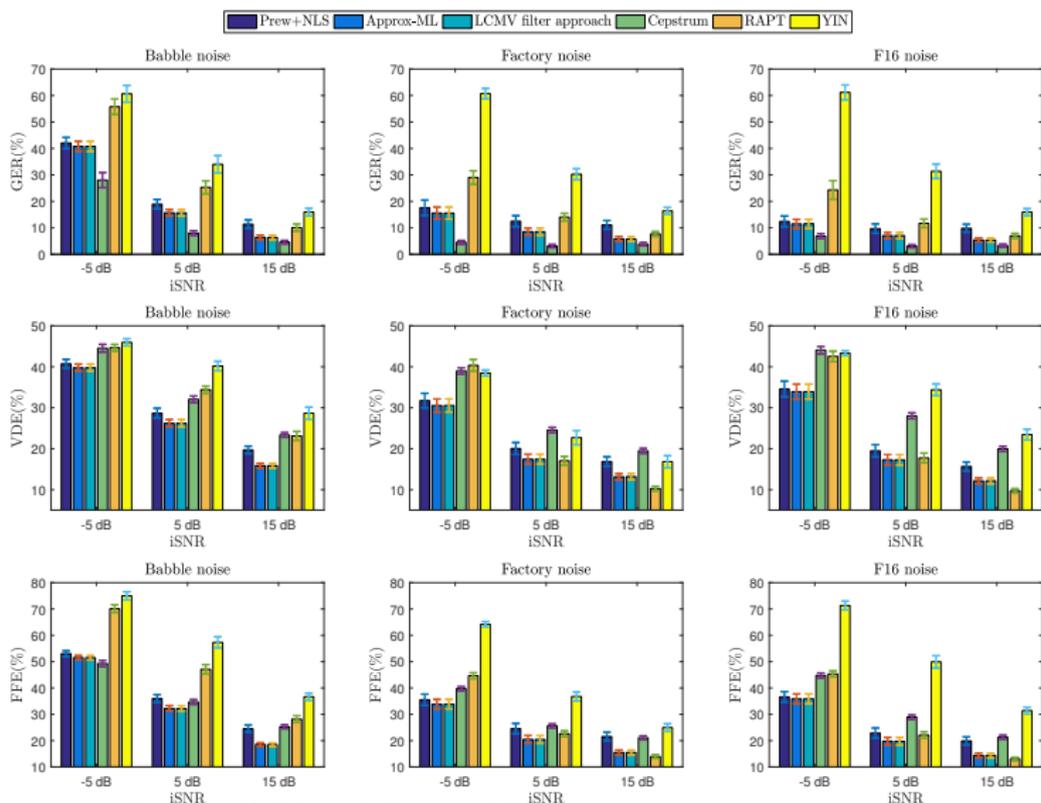


Figure: GER, VDE and FFE as a function of the iSNR.

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- ▶ Parametric estimators, as the NLS, are only statistically efficient in WGN conditions.

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- ▶ Parametric estimators, as the NLS, are only statistically efficient in WGN conditions.
- ▶ The NLS estimator (rapid implementation) is still useful as long as the additive noise shape is taken into account in forming a pre-whitening filter.

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- ▶ The NLS estimator (rapid implementation) is still useful as long as the additive noise shape is taken into account in forming a pre-whitening filter.
- ▶ f_0 along with model order and amplitudes, and AR noise parameters are estimated **iteratively**. This reduces the risk of octave errors and voicing detection errors noticeably.

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- ▶ f_0 along with model order and amplitudes, and AR noise parameters are estimated **iteratively**. This reduces the risk of octave errors and voicing detection errors noticeably.
- ▶ The proposed approach offers better performance than state-of-the-art f_0 methods only when the reiterations are applied.

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- ▶ f_0 along with model order and amplitudes, and AR noise parameters are estimated **iteratively**. This reduces the risk of octave errors and voicing detection errors noticeably.
- ▶ The proposed approach offers better performance than state-of-the-art f_0 methods only when the reiterations are applied.
- ▶ Even if the proposed does not take the correlation of consecutive estimates into account, is more robust to the noise.

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