

Iterative Estimation of Phase Using Complex Cepstrum Representation

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Introduction

- Goal: accurate estimation of phase from speech
- Importance of phase
 - 1. Speech parameterization for TTS
 - 2. Features for ASR
 - 3. Detection of speech pathologies
- Estimation of continuous phase spectrum
 - 1. Detection of glottal closure instants (GCI)
 - 2. Phase unwrapping
- Minimum MSE-based complex cepstrum analysis [Maia et al., 2013a]
 - Complex cepstrum analysis with no phase unwrapping
 - GCI are iteratively optimized in the process
- Phase estimation can be performed using the same concept!

Typical phase estimation issues

Speech segmentation

- Detection of glottal closure instants (GCIs)
- Influence of the shape and length of the windows

Phase unwrapping

- Discrete Fourier transform (DFT) gives phase modulo 2π
- Phase must be unwrapped





Complex cepstrum-based speech analysis and synthesis

Analysis

- 1. Determine pitch period onset times (or GCI): $\{p_0, \ldots, p_{Z-1}\}$
- 2. Complex cepstrum analysis

$$\begin{array}{lll} \hat{h}(n) & = & \frac{1}{2\pi} \int_{-\pi}^{\pi} \ln |S(e^{j\omega})| e^{j\omega n} d\omega & + & \frac{j}{2\pi} \int_{-\pi}^{\pi} \theta(\omega) e^{j\omega n} d\omega \\ & \downarrow & & \downarrow \\ \text{Cepstrum at } p_z & \text{Amplitude response at } p_z & \text{Phase response at } p_z \end{array}$$

Synthesis

1. Derive non-causal impulse responses

- 2. Make excitation e(n) with pulses located at p_z
- 3. Synthesize speech by making $\tilde{s}(n) = h(n) * e(n)$

Proposed phase estimation approach



- Phase iteratively estimated by minimizing the error between natural and reconstructed speech in the time domain
- Pitch period onsets jointly optimized
- No windowing: frame-based time-varying filtering
- No phase unwrapping: cepstral domain

Iterative estimation of phase: requirement

- Pulse positions $\{p_0, \ldots, p_{Z-1}\}$ must correctly indicate pitch periods but not necessarily GCI
- Because
 - 1. Smooth speech spectral envelope at p_z

$$|H_{z}(e^{j\omega})| = \left|\sum_{n=p_{z-1}}^{p_{z+1}} k(n-p_{z-1})s(n)e^{-j\omega n}\right|, \quad k(n): \text{ window}$$

2. Real cepstrum

$$\hat{h}_r(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \ln |H(e^{j\omega})| e^{j\omega n} d\omega, \quad n = -C, \dots, C$$

3. Minimum-phase cepstrum

$$\hat{h}_m(n) = \begin{cases} 0 & n < 0 \\ \hat{h}_r(n) & n = 0 \\ 2\hat{h}_r(n) & n = 1, \dots, C \end{cases}$$



All-pass/minimum-phase speech decomposition



$$\hat{\boldsymbol{h}}_{m} = \begin{bmatrix} \hat{h}_{m}(0) & \cdots & \hat{h}_{m}(C) \end{bmatrix}^{\top} \\ \boldsymbol{\phi} = \begin{bmatrix} \hat{h}_{a}(1) & \cdots & \hat{h}_{a}(C) \end{bmatrix}$$

Min.-phase cepstrum: amplitude and min. phase! Causal all pass cepstrum: residual phase!

Parameters of the model



- $p = \{p_0, \dots, p_{Z-1}\}$: pulse locations
- Non-causal synthesis filter parameters: variables to be determined

• $\{\phi_0, \ldots, \phi_{T-1}\}$: phase features at every frame

A two-step optimization process at the utterance level

Step 1 Estimation of the locations and amplitudes of the excitation signal

- \blacksquare Keep $\hat{h}(n)$ fixed
- Solution $\{p_0, \ldots, p_{Z-1}\}$ and $\{a_0, \ldots, a_{Z-1}\}$

Step 2 Phase estimation given the new pulse positions and amplitudes

- **Keep** e(n) fixed
- real Re-estimate $\phi(n)$ using a gradient method
 - → Non-linear relationship between h(n) and $\phi(n)$

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \exp\left\{\sum_{p=0}^{C} \hat{h}_m(p) e^{-j\omega p} - 2j \sum_{p=1}^{C} \phi(p-1) \sin(\omega p) + j\omega n\right\} d\omega$$

 $h(n) = f(\phi(n))$



Step 1: estimation of $\{a_0, p_0, ..., a_{Z-1}, p_{Z-1}\}$



$$\varepsilon\left(\boldsymbol{p},\boldsymbol{a}\right) = \frac{1}{N} \left[\boldsymbol{s} - \sum_{z=0}^{Z-1} a_{z} \boldsymbol{g}_{p_{z}}\right]^{\top} \left[\boldsymbol{s} - \sum_{z=0}^{Z-1} a_{z} \boldsymbol{g}_{p_{z}}\right] \Rightarrow \begin{pmatrix} \hat{p}_{z} = \operatorname*{arg\,max} \\ p_{z} - \Delta p, \dots, p_{z} + \Delta p \\ p_{z} = \alpha \operatorname{arg\,max} \\ p_{z} = \Delta p, \dots, p_{z} + \Delta p \\ q_{p_{z}}^{\top} \left[\boldsymbol{s} - \sum_{\substack{i=0\\i\neq z}}^{Z-1} a_{i} \boldsymbol{g}_{p_{z}}\right] \\ \hat{a}_{z} = \frac{\boldsymbol{g}_{p_{z}}^{\top} \left[\boldsymbol{s} - \sum_{\substack{i=0\\i\neq z}}^{Z-1} a_{i} \boldsymbol{g}_{p_{z}}\right]}{\boldsymbol{g}_{p_{z}}^{\top} \boldsymbol{g}_{p_{z}}}$$



Step 2: estimation of $\{\phi_0, \ldots, \phi_{T-1}\}$



→ Matrix A_t contains samples of e(n) at frame t

→ Impulse response vector at frame t: $h_t = \left[h_t \left(-\frac{M}{2}\right) \cdots h_t \left(\frac{M}{2}\right)\right]^\top$

Step 2: estimation of $\{\phi_0, \ldots, \phi_{T-1}\}$

🖙 MSE

$$\varepsilon = \frac{1}{N} \left[\boldsymbol{s} - \sum_{t=0}^{T-1} \boldsymbol{A}_t \boldsymbol{h}_t \right]^{\top} \left[\boldsymbol{s} - \sum_{t=0}^{T-1} \boldsymbol{A}_t \boldsymbol{h}_t \right]$$

Cost function

$$\begin{split} \varepsilon\left(\boldsymbol{\phi}_{t}\right) &= \frac{1}{N} \left[\boldsymbol{r}_{t}^{\top} \boldsymbol{r}_{t} - 2\boldsymbol{r}_{t} \boldsymbol{A}_{t} f_{1}\left(\boldsymbol{\phi}_{t}\right) + \left\{ f_{1}\left(\boldsymbol{\phi}_{t}\right) \right\}^{\top} \boldsymbol{U}_{t} f_{1}\left(\boldsymbol{\phi}_{t}\right) \right] \\ & \left\{ \boldsymbol{r}_{t} = \boldsymbol{s} - \sum_{j=0, j \neq t}^{T-1} \boldsymbol{A}_{j} f_{1}\left(\boldsymbol{\phi}_{j}\right) \\ \boldsymbol{U}_{t} = \boldsymbol{A}_{t}^{\top} \boldsymbol{A}_{t} \end{split} \right.$$

Relationship between impulse response and residual phase

$$\begin{aligned} \boldsymbol{h}_{t} &= f_{1}\left(\boldsymbol{\phi}_{t}\right) = \frac{1}{2L}\boldsymbol{D}_{2}\exp\left(\boldsymbol{D}_{m,1}\hat{\boldsymbol{h}}_{m,t} + \boldsymbol{D}_{a,1}\boldsymbol{\phi}_{t}\right) \\ \begin{cases} D_{m,1}(i,j) &= e^{-j\omega_{i}j} & -L+1 \leq i \leq L, 0 \leq j \leq C \\ D_{a,1}(i,j) &= -2j\sin\left(\omega_{i}j\right) & -L+1 \leq i \leq L, 0 \leq j \leq C \\ D_{2}(i,j) &= e^{j\omega_{j}i} & -\frac{M}{2} \leq i \leq \frac{M}{2}, -L+1 \leq j \leq L \end{cases} \end{aligned}$$

 ϕ_t is determined by a gradient descent method!



Experiment

Conditions

- Female UK English speaker, 22.05 kHz, 50 sentences in each of the following styles: angry, fear, happy, neutral, sad and tender
- Methods evaluated
 - 1. GCI detection using DYPSA [Naylor et al., 2007] + phase unwrapping with a 8192-point DFT (g)
 - 2. MSE cepstrum analysis (MSE-CCEP) as in [Maia et al., 2013b] (b)
 - 3. Proposed (r)

Evaluation criterion





Results



 Proposed method performs better than GCI detection + unwrapping for all speech styles

 Proposed is similar to MSE-CCEP for all styles

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Analysis-synthesis comparison with MSE-CCEP



- MSE-CCEP performs better in terms of analysissynthesis
- MSE-CCEP optimizes not only phase but also amplitude
- In terms of phase estimation both perform the same
- The proposed method runs in average 3 times faster

- Method to estimate short-term residual phase using complex cepstrum-based analysis and synthesis of speech
- Accurate markings of the pitch periods are necessary
- Better performance than GCI detection followed by multi-resolution phase unwrapping
- Similar performance to the more computationally expensive MSE complex cepstrum analysis
- An automatic way to extract phase information from pitch marks, with no need of
- Future work: application to TTS and ASR exact GCI information



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