

# Efficient Near Optimal Joint Modulation Classification and Detection for MU-MIMO Systems



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## Introduction

A joint maximum likelihood (ML) modulation classification (MC) of the co-scheduled user and data detection receiver is developed

#### Reference MU-MIMO Detection Schemes:

- Interference Ignoring
- Interference Rejection Combining (IRC) and MMSE Only make use of the channel estimate of the interferer
- Interference Assuming Say 16-QAM for example
- Interference Estimation Add a MC routine Feed estimate to a regular Interference Aware (IA) receiver

### **Modulation Classification Schemes:**

- Likelihood-based
- Feature-based

# **System Model**

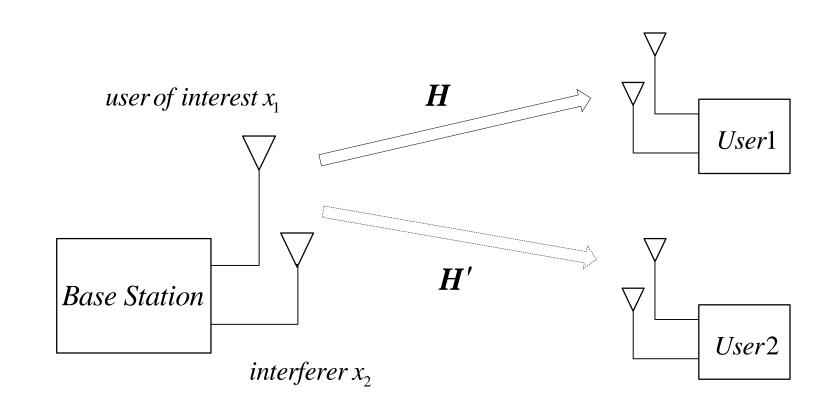
$$y = Hx + n$$

- $\mathbf{H} = N_r \times N_t$  channel matrix
- x transmitted QAM symbols
- n complex additive white Gaussian noise with zero mean and variance  $\sigma^2 = \frac{N_t}{SNR}$

We consider the case  $N_r = N_t = 2$  $\mathbf{y} = \mathbf{h}_1 x_1 + \mathbf{h}_2 x_2 + \mathbf{n}$ 

 $\overline{\Lambda}$  is the constellation of user of interest

 $\Lambda_0$  is Ø (no interference) -  $\Lambda_1$  is QPSK -  $\Lambda_2$  is 16-QAM -  $\Lambda_3$  is 64-QAM



# **Proposed Joint MC and Detection**

Bayesian formulation: 4-ary hypothesis testing

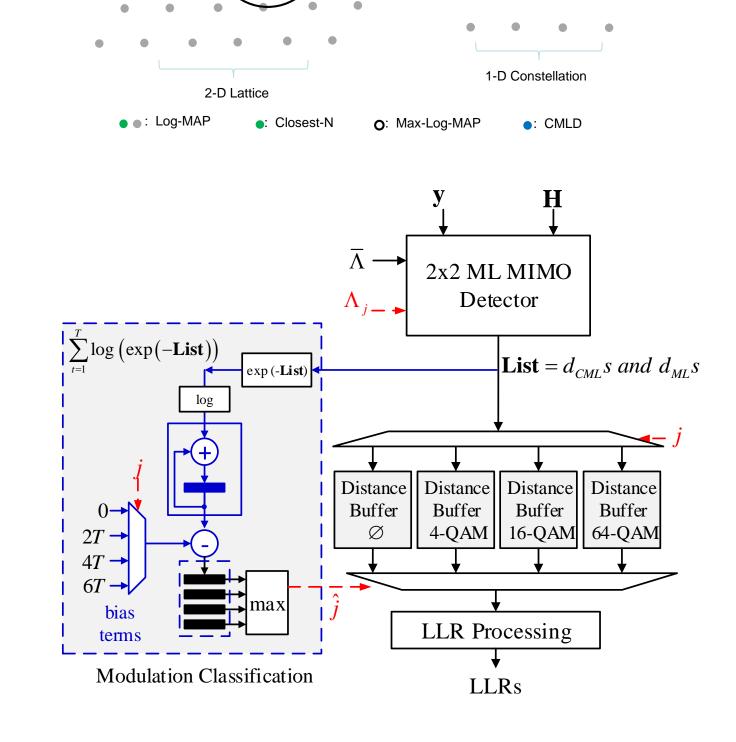
$$\begin{cases} \theta_{\mathbf{0}} \colon \mathbf{y} \sim P(\mathbf{y}; x_1 \in \overline{\Lambda}, x_2 \in \Lambda_0) \\ \theta_{\mathbf{1}} \colon \mathbf{y} \sim P(\mathbf{y}; x_1 \in \overline{\Lambda}, x_2 \in \Lambda_1) \\ \theta_{\mathbf{2}} \colon \mathbf{y} \sim P(\mathbf{y}; x_1 \in \overline{\Lambda}, x_2 \in \Lambda_2) \\ \theta_{\mathbf{3}} \colon \mathbf{y} \sim P(\mathbf{y}; x_1 \in \overline{\Lambda}, x_2 \in \Lambda_3) \end{cases}$$

$$\begin{split} P(\mathbf{y}; \Lambda_n) &= \sum_{x_1 \in \overline{\Lambda}, x_2 \in \Lambda_n} P(\mathbf{y} | x_1, x_2) P(x_1, x_2) \\ \hat{n}_{\text{Log-MAP}} &= \underset{n = 0, 1, 2, 3}{\operatorname{argmax}} \left( \log \frac{1}{|\Lambda_n|} + \sum_{x_1 \in \overline{\Lambda}, x_2 \in \Lambda_n} \exp \left( -\frac{1}{\sigma^2} \|\mathbf{y} - \mathbf{H} \mathbf{x}\|^2 \right) \right) \\ d_{\text{ML},n} &= \min_{x_1 \in \overline{\Lambda}, x_2 \in \Lambda_n} \varphi(\mathbf{x}) \qquad \varphi(\mathbf{x}) = \frac{1}{\sigma^2} \|\mathbf{y} - \mathbf{H} \mathbf{x}\|^2 \\ \hat{n}_{\text{Max-Log-MAP}} &= \underset{n = 0, 1, 2, 3}{\operatorname{argmax}} \left( \log \frac{1}{|\Lambda_n|} - d_{\text{ML},n} \right) \end{split}$$

In general, for a group of distance metrics S, and after T observations:

$$\widehat{n} = \operatorname*{argmax}_{n=0,1,2,3} \sum_{t=1}^{T} \left( \log \frac{1}{|\Lambda_n|} + \sum_{\mathbf{x} \in S} \exp \left( -\frac{1}{\sigma^2} ||\mathbf{y} - \mathbf{H}\mathbf{x}||^2 \right) \right)$$

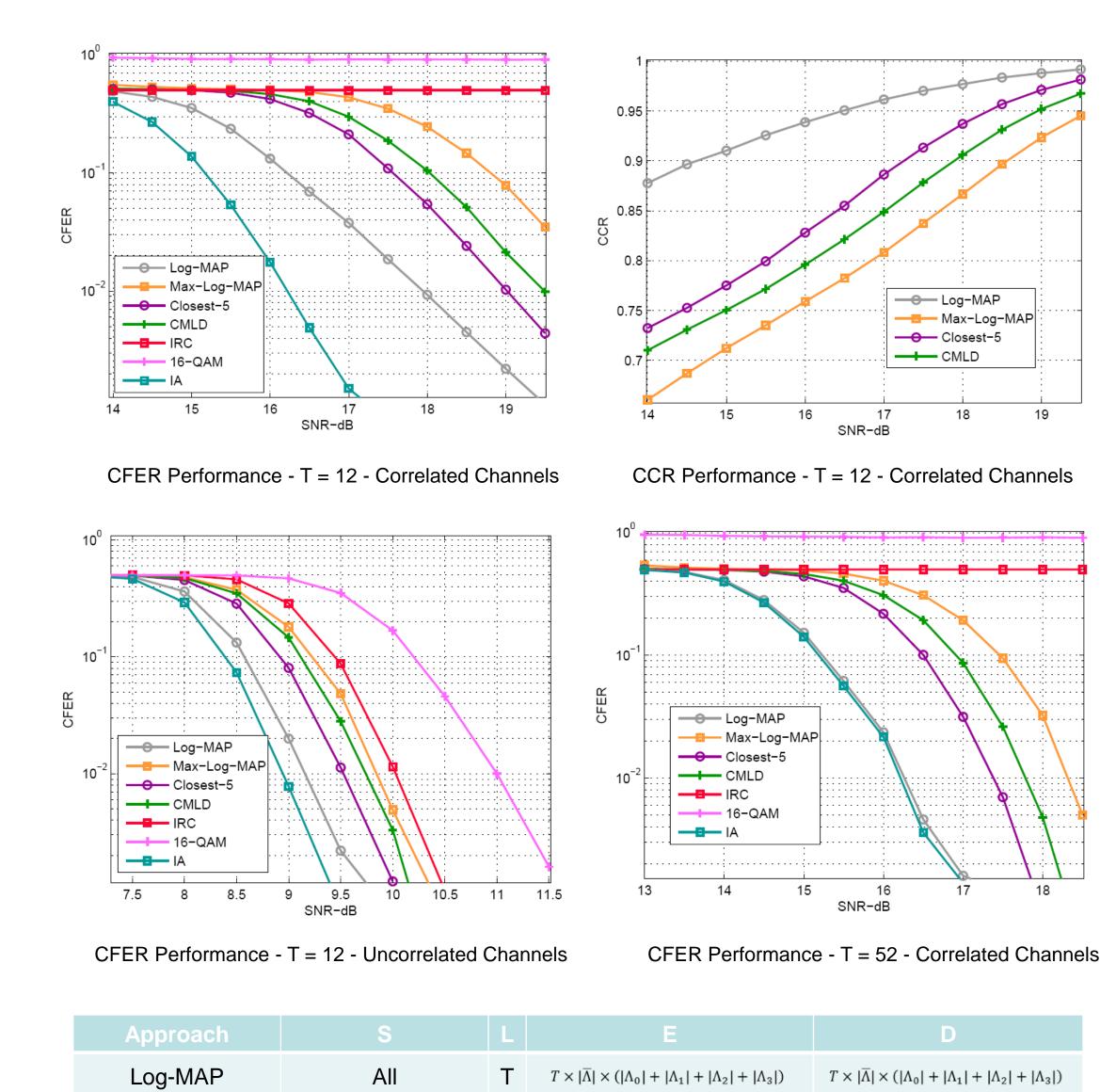
- Closest\_N: S consists of points corresponding to smallest N distance metrics.
- CMLD: S consists of points corresponding to ML and counter-ML distances.



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## Results

- Correct Classification Ratio (CCR) of classifiers
- Coded frame error rate (CFER) of detectors



# In the Context of WiFi

 $T \times 4 \times N$ 

 $4 \times T \times (K_1 + 1)$ 

 $T\left(K_2^{(0)}+K_2^{(1)}+K_2^{(2)}+K_2^{(3)}+4\right)$ 

 $4 \times T$ 

 $T \times |\overline{\Lambda}| \times (|\Lambda_0| + |\Lambda_1| + |\Lambda_2| + |\Lambda_3|)$ 

 $T \times |\overline{\Lambda}| \times (|\Lambda_0| + |\Lambda_1| + |\Lambda_2| + |\Lambda_3|)$ 

 $T \times |\overline{\Lambda}| \times (|\Lambda_0| + |\Lambda_1| + |\Lambda_2| + |\Lambda_3|)$ 

 $T \times |\overline{\Lambda}| \times (|\Lambda_0| + |\Lambda_1| + |\Lambda_2| + |\Lambda_3|)$ 

 $T \times |\overline{\Lambda}| \times (|\Lambda_0| + |\Lambda_1| + |\Lambda_2| + |\Lambda_3|)$ 

From 52 to 234 data tones

Closest\_N

CMLD

CMLD1

CMLD2

Max-Log-MAP

- large number of OFDM symbols (L)
- Constant interferer over T tones and L symbols

Closest N

ML+CMLs of x

ML+CMLs of  $x_1$  T

- MC executes on one OFDM symbol in the frame
- $\frac{3}{7}\%$  increase in distance computations
- L takes values from 8 to more than 1024

