

Conjugate Augmented Spatial-Temporal Near-Field Sources Localization with Cross Array

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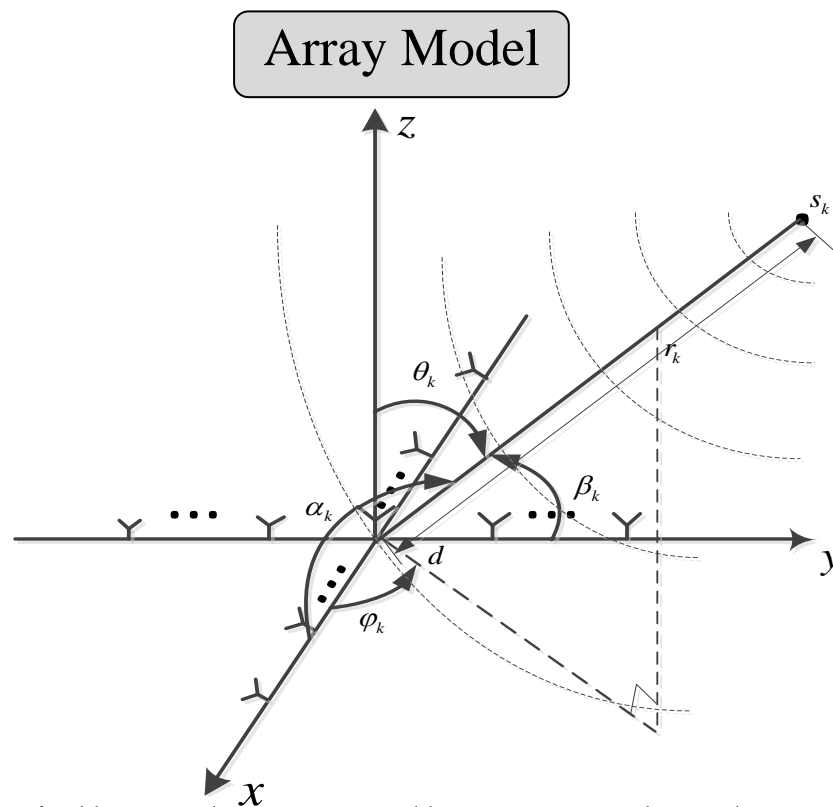
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Outline

1. **Introduction**
2. **Proposed Method**
3. **Simulation Results**
4. **Conclusions**

1.Introduction



- ◆ K narrowband, spatially and temporally uncorrelated sources S_k
- ◆ two uniform linear arrays (ULAs) : N_x sensors on x-axis and N_y sensors on y-axis
- ◆ the elevation and azimuth angles of the k -th signal : θ_k and φ_k
- ◆ the angles between the k -th signal and the x and y axes: α_k and β_k
- ◆ the range of the k -th signal: r_k
- ◆ If α_k and β_k are determined, θ_k and φ_k can be uniquely identified.

1.Introduction

Data Model

$$\begin{aligned}x_m(n) &= \sum_{k=1}^K s_k(n) e^{-j\tau_{m_x,k}} + n_{m_x}(n) \\y_m(n) &= \sum_{k=1}^K s_k(n) e^{-j\tau_{m_y,k}} + n_{m_y}(n)\end{aligned}\tag{1}$$

With the Fresnel approximation, $\tau_{m_x,k}$ and $\tau_{m_y,k}$ can be expressed as:

$$\begin{aligned}\tau_{m_x,k} &= \omega_{xk} m + \phi_{xk} m^2 \\ \tau_{m_y,k} &= \omega_{yk} m + \phi_{yk} m^2\end{aligned}\tag{2}$$

where

$$\begin{aligned}\omega_{xk} &= -\frac{2\pi d}{\lambda} \cos \alpha_k, & \phi_{xk} &= \frac{\pi d^2}{\lambda r_k} \sin^2 \alpha_k \\ \omega_{yk} &= -\frac{2\pi d}{\lambda} \cos \beta_k, & \phi_{yk} &= \frac{\pi d^2}{\lambda r_k} \sin^2 \beta_k\end{aligned}\tag{3}$$

1.Introduction

Data Model

$$\mathbf{x}(n) = \mathbf{A}_x \mathbf{s}(n) + \mathbf{n}_x(n) \quad (4)$$

$$\mathbf{y}(n) = \mathbf{A}_y \mathbf{s}(n) + \mathbf{n}_y(n) \quad (5)$$

where

$$\mathbf{A}_x = [\mathbf{a}(\omega_{x1}, \phi_{x1}), \dots, \mathbf{a}(\omega_{xk}, \phi_{xk})]$$

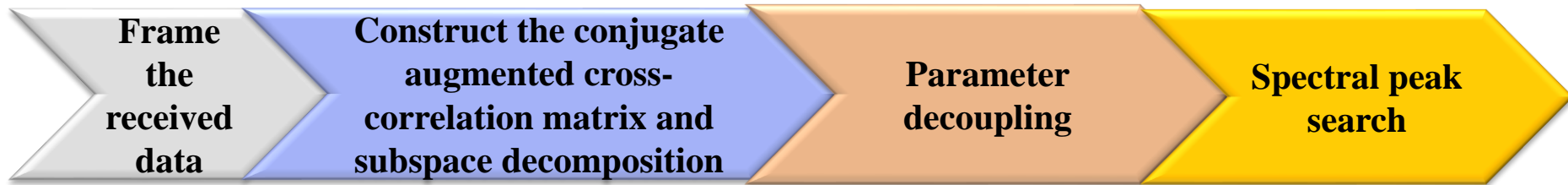
$$\mathbf{a}(\omega_{xk}, \phi_{xk}) = \left[e^{-j[\omega_{xk}(-M_x) + \phi_{xk}(-M_x)^2]} \quad \dots \quad e^{-j(\omega_{xk}M_x + \phi_{xk}M_x^2)} \right]^T$$

$$\mathbf{A}_y = [\mathbf{a}(\omega_{y1}, \phi_{y1}), \dots, \mathbf{a}(\omega_{yk}, \phi_{yk})]$$

$$\mathbf{a}(\omega_{yk}, \phi_{yk}) = \left[e^{-j[\omega_{yk}(-M_y) + \phi_{yk}(-M_y)^2]} \quad \dots \quad e^{-j(\omega_{yk}M_y + \phi_{yk}M_y^2)} \right]^T$$

$\mathbf{n}_x(n)$ and $\mathbf{n}_y(n)$ represent the additive Gaussian noise vectors for the two ULAs, respectively.

2. Proposed Method



1. Frame the received data

$\mathbf{x}(n)$ and $\mathbf{y}(n)$ are divided into L frames according to the principle of maximum overlap in the time domain. The l -th ($l = 1, 2, \dots, L$) frame data can be expressed as:

$$\begin{aligned} \mathbf{X}_l &= [\mathbf{x}(l), \mathbf{x}(l+1), \dots, \mathbf{x}(l+N-L)] \\ \mathbf{Y}_l &= [\mathbf{y}(l), \mathbf{y}(l+1), \dots, \mathbf{y}(l+N-L)] \end{aligned} \quad (6)$$

2. Proposed Method

2. Construct the conjugate augmented cross-correlation matrix and subspace decomposition

Select one array element from each of the two uniform arrays, then the delay cross-correlation information of the received data of these two array elements is:

$$\begin{aligned} r_{m_1, m_2}(l-1+L) &= E\{x_{m_1}(n+l-1)y_{m_2}^*(n)\} \\ &= \sum_{k_1=1}^K [a_{x, m_1}(\alpha_{k_1}, r_{k_1})a_{y, m_2}^*(\beta_{k_1}, r_{k_1})\mathbf{R}_{ss}(k_1, l-1+L)] \\ &\quad + \delta(m_1) \cdot \delta(m_2) \cdot \delta(l-1) \cdot \sigma_w^2 \end{aligned} \quad (7)$$

where \mathbf{R}_{ss} is the signal delay autocorrelation matrix.

By arranging $r_{m_1, m_2}(l_1)(l_1=1, 2, \dots, 2L-1)$, we have

$$\mathbf{R}_{xy} = [\mathbf{r}_{xy}(1), \mathbf{r}_{xy}(2), \dots, \mathbf{r}_{xy}(2L-1)] = (\mathbf{A}_y^*(\beta, r) \odot \mathbf{A}_x(\alpha, r))\mathbf{R}_{ss} + \sigma_w^2 \cdot \mathbf{R}_{w0} \quad (8)$$

where \mathbf{R}_{w0} only has one nonzero element, that is, $\mathbf{R}_{w0}(1, L) = 1$

2. Proposed Method

Taking advantage of the conjugate symmetry of the autocorrelation function of the signal delay, we have

$$\begin{aligned}
 \tilde{\mathbf{R}}_{xy} &= \begin{bmatrix} \mathbf{R}_{xy} \\ \mathbf{R}_{xy}^* \mathbf{J}_{(2L-1)} \end{bmatrix} \\
 &= \begin{bmatrix} \mathbf{A}_y^*(\beta, r) \odot \mathbf{A}_x(\alpha, r) \\ \mathbf{A}_y(\beta, r) \odot \mathbf{A}_x^*(\alpha, r) \end{bmatrix} \mathbf{R}_{ss} + \sigma_w^2 \cdot \begin{bmatrix} \mathbf{R}_{w0} \\ \mathbf{R}_{w0} \mathbf{J}_{(2L-1)} \end{bmatrix} \\
 &= \tilde{\mathbf{A}}_{xy} \mathbf{R}_{ss} + \sigma_w^2 \cdot \tilde{\mathbf{R}}_{w0}
 \end{aligned} \tag{9}$$

Calculate the covariance matrix of as:

$$\tilde{\mathbf{R}} = E \left\{ \tilde{\mathbf{R}}_{xy} \tilde{\mathbf{R}}_{xy}^H \right\} = \tilde{\mathbf{A}}_{xy} \tilde{\mathbf{R}}_s \tilde{\mathbf{A}}_{xy}^H + \sigma_w^4 \tilde{\mathbf{R}}_w \tag{10}$$

Where $\tilde{\mathbf{R}}_s$ is the covariance matrix of signal delay autocorrelation matrix \mathbf{R}_{ss} , $\tilde{\mathbf{R}}_w$

is extremely sparse, with only $\tilde{\mathbf{R}}_w \left(\frac{N_x N_y + 1}{2}, \frac{N_x N_y + 1}{2} \right) = 1$.

2. Proposed Method

The eigenvalue decomposition of $\tilde{\mathbf{R}}$

$$\tilde{\mathbf{R}} = \mathbf{U}_s \boldsymbol{\Sigma}_s \mathbf{U}_s^H + \mathbf{U}_n \boldsymbol{\Sigma}_n \mathbf{U}_n^H \quad (11)$$

where

\mathbf{U}_n \longrightarrow noise subspace

\mathbf{U}_s \longrightarrow signal subspace

$\boldsymbol{\Sigma}_s$ \longrightarrow main eigenvalue matrix

$\boldsymbol{\Sigma}_n$ \longrightarrow small eigenvalue matrix

3. Parameter decoupling

According to the relationship between Khatri-Rao product and Kronecker product, the above-constructed delay cross-correlation matrix (formula (8)) can be rewritten as:

$$\mathbf{R}_{xy} = [\mathbf{a}_y^*(\beta_1, r_1) \otimes \mathbf{a}_x(\alpha_1, r_1), \dots, \mathbf{a}_y^*(\beta_K, r_K) \otimes \mathbf{a}_x(\alpha_K, r_K)] \mathbf{R}_{ss} + \sigma_w^2 \mathbf{R}_{w0} \quad (12)$$

2. Proposed Method

According to the symmetry of array structure, $\mathbf{a}_y^*(\beta, r)$ can be decomposed into the following form

$$\mathbf{a}_y^*(\beta, r) = \begin{bmatrix} e^{j(-M_y)\omega_y} & & & & & & & & & \\ & e^{j(-M_y+1)\omega_y} & & & & & & & & \\ & & \ddots & & & & & & & \\ & & & \ddots & & & & & & \\ & & & & 1 & & & & & \\ & & & & & \ddots & & & & \\ & & & & & & e^{j(-1)^2\phi_y} & & & \\ & & e^{j(M_y-1)\omega_y} & & & & & & & \\ e^{jM_y\omega_y} & & & & & & & & & \end{bmatrix} \begin{bmatrix} e^{j(-M_y)^2\phi_y} \\ e^{j(-M_y+1)^2\phi_y} \\ \vdots \\ e^{j(-1)^2\phi_y} \\ 1 \end{bmatrix} = \boldsymbol{\zeta}_y^*(\beta) \mathbf{v}_y^*(\beta, r) \quad (13)$$

Then, $\mathbf{a}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r)$ can be rewritten as:

$$\mathbf{a}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r) = (\boldsymbol{\zeta}_y^*(\beta) \mathbf{v}_y^*(\beta, r)) \otimes (\mathbf{I}_{N_x} \mathbf{a}_x(\alpha, r)) \quad (14)$$

We denote the columns of $\tilde{\mathbf{R}}_{xy}$ as $\tilde{\mathbf{a}}_{xy}$, then

$$\tilde{\mathbf{a}}_{xy} = \begin{bmatrix} (\boldsymbol{\zeta}_y^*(\beta) \mathbf{v}_y^*(\beta, r)) \otimes (\mathbf{I}_{N_x} \mathbf{a}_x(\alpha, r)) \\ (\boldsymbol{\zeta}_y(\beta) \mathbf{v}_y(\beta, r)) \otimes (\mathbf{I}_{N_x} \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \quad (15)$$

2. Proposed Method

According to the property of the Kronecker product, we have

$$(\mathbf{A} \otimes \mathbf{B})(\mathbf{C} \otimes \mathbf{D}) = \mathbf{AC} \otimes \mathbf{BD} \quad (15)$$

$\tilde{\mathbf{a}}_{xy}$ can be rewritten as follows:

$$\begin{aligned} \tilde{\mathbf{a}}_{xy} &= \begin{bmatrix} (\zeta_y^*(\beta) \mathbf{v}_y^*(\beta, r)) \otimes (\mathbf{I}_{N_x} \mathbf{a}_x(\alpha, r)) \\ (\zeta_y(\beta) \mathbf{v}_y(\beta, r)) \otimes (\mathbf{I}_{N_x} \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \\ &= \begin{bmatrix} (\zeta_y^*(\beta) \otimes \mathbf{I}_{N_x})(\mathbf{v}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r)) \\ (\zeta_y(\beta) \otimes \mathbf{I}_{N_x})(\mathbf{v}_y(\beta, r) \otimes \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \\ &= \begin{bmatrix} \mathbf{C}_1(\beta)(\mathbf{v}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r)) \\ \mathbf{C}_2(\beta)(\mathbf{v}_y(\beta, r) \otimes \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \\ &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} (\mathbf{v}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r)) \\ (\mathbf{v}_y(\beta, r) \otimes \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \end{aligned} \quad (16)$$

2. Proposed Method

We define $\mathbf{C}(\beta)$ that is only related to β :

$$\mathbf{C}(\beta) = \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \quad (17)$$

Based on the rank reduction principle ,we can construct the MUSIC spectral function related to angle β :

$$\mathbf{D}(\beta) = \mathbf{C}^H(\beta) \mathbf{U}_n \mathbf{U}_n^H \mathbf{C}(\beta) \quad (18)$$

2. Proposed Method

With the same operation as (16), we have

$$\begin{aligned}
 \tilde{\mathbf{a}}_{xy} &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} (\mathbf{v}_y^*(\beta, r) \otimes \mathbf{a}_x(\alpha, r)) \\ (\mathbf{v}_y(\beta, r) \otimes \mathbf{a}_x^*(\alpha, r)) \end{bmatrix} \\
 &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} (\mathbf{I}_{N_y} \mathbf{v}_y^*(\beta, r) \otimes \zeta_x(\alpha) \mathbf{v}_x(\alpha, r)) \\ (\mathbf{I}_{N_y} \mathbf{v}_y(\beta, r) \otimes \zeta_x^*(\alpha) \mathbf{v}_x^*(\alpha, r)) \end{bmatrix} \\
 &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} (\mathbf{I}_{N_y} \otimes \zeta_x(\alpha)) (\mathbf{v}_y^*(\beta, r) \otimes \mathbf{v}_x(\alpha, r)) \\ (\mathbf{I}_{N_y} \otimes \zeta_x^*(\alpha)) (\mathbf{v}_y(\beta, r) \otimes \mathbf{v}_x^*(\alpha, r)) \end{bmatrix} \\
 &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} \mathbf{E}_1(\alpha) (\mathbf{v}_y^*(\beta, r) \otimes \mathbf{v}_x(\alpha, r)) \\ \mathbf{E}_2(\alpha) (\mathbf{v}_y(\beta, r) \otimes \mathbf{v}_x^*(\alpha, r)) \end{bmatrix} \\
 &= \begin{bmatrix} \mathbf{C}_1(\beta) & \\ & \mathbf{C}_2(\beta) \end{bmatrix} \begin{bmatrix} \mathbf{E}_1(\alpha) & \\ & \mathbf{E}_2(\alpha) \end{bmatrix} \begin{bmatrix} (\mathbf{v}_y^*(\beta, r) \otimes \mathbf{v}_x(\alpha, r)) \\ (\mathbf{v}_y(\beta, r) \otimes \mathbf{v}_x^*(\alpha, r)) \end{bmatrix}
 \end{aligned} \tag{18}$$

Define $\mathbf{E}(\alpha)$ that is only related to α :

$$\mathbf{E}(\alpha) = \begin{bmatrix} \mathbf{E}_1(\alpha) & \\ & \mathbf{E}_2(\alpha) \end{bmatrix} \tag{19}$$

Based on (19), we can construct another MUSIC spectral function for angle α :

$$\mathbf{F}(\hat{\beta}, \alpha) = \mathbf{E}^H(\alpha) \mathbf{C}^H(\hat{\beta}) \mathbf{U}_n \mathbf{U}_n^H \mathbf{C}(\hat{\beta}) \mathbf{E}(\alpha) \tag{20}$$

2. Proposed Method

4. Spectral peak search

We separately perform three 1-D searches to obtain estimates of the three parameters:

$$\text{For } \beta : \quad \hat{\beta} = \arg \max_{\beta} \frac{1}{\det[\mathbf{D}(\beta)]} \quad (21)$$

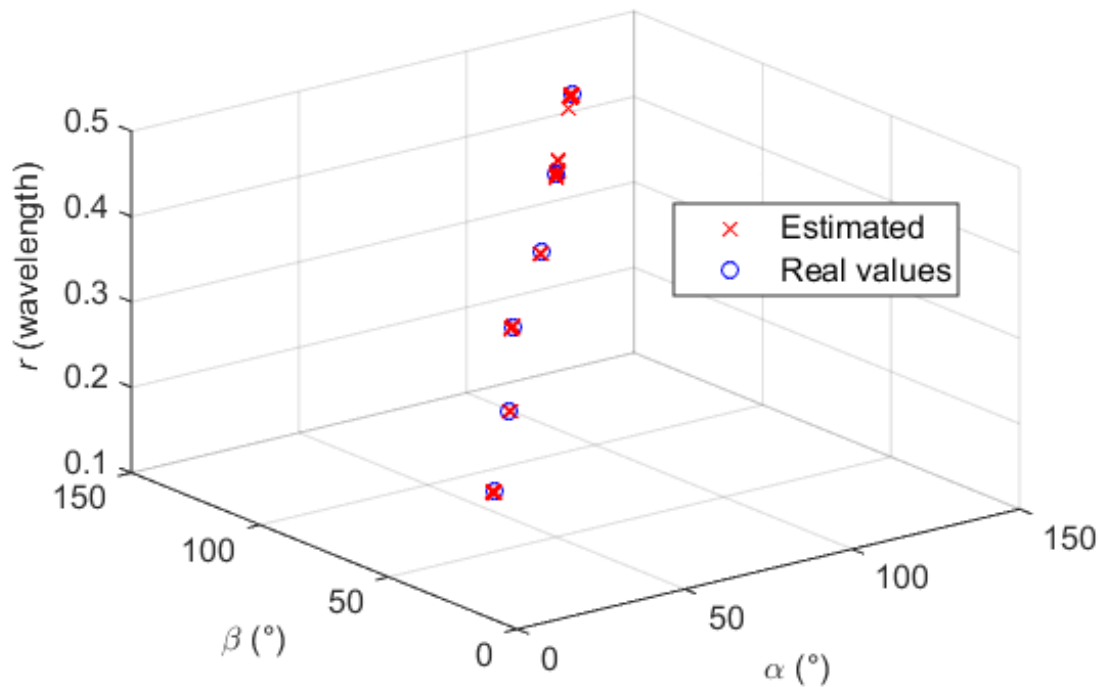
$$\text{For } \alpha : \quad \hat{\alpha} = \arg \max_{\alpha} \frac{1}{\det[\mathbf{F}(\hat{\beta}, \alpha)]} \quad (22)$$

$$\text{For } r : \quad \hat{r}_k = \arg \max_r f(\hat{\alpha}_k, \hat{\beta}_k, r) = \frac{1}{\tilde{\mathbf{a}}_{xy}^H(\hat{\alpha}, \hat{\beta}, r) \mathbf{U}_n \mathbf{U}_n^H \tilde{\mathbf{a}}_{xy}(\hat{\alpha}, \hat{\beta}, r)} \quad (23)$$

3. Simulation Results

Performance test

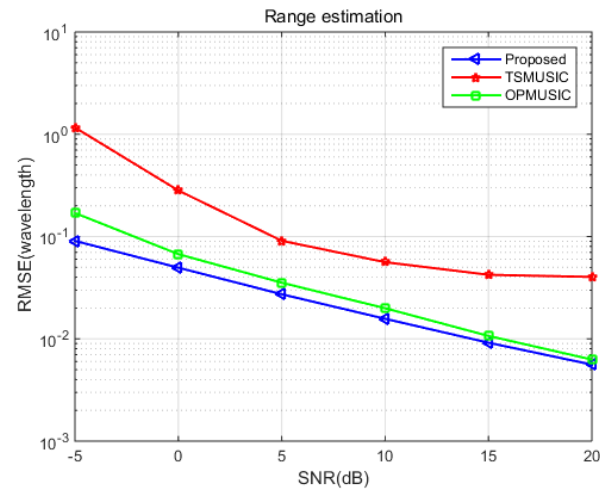
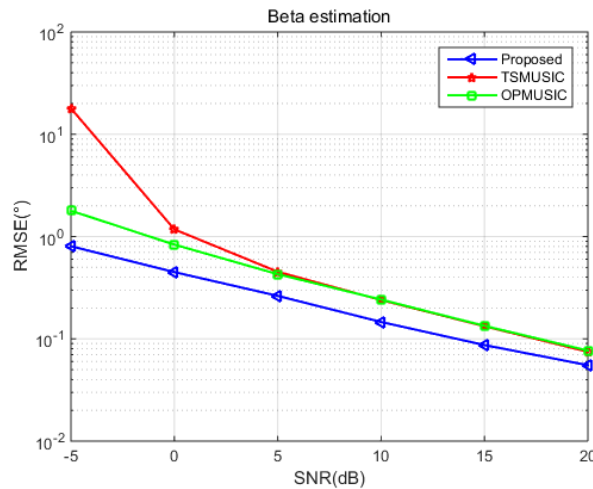
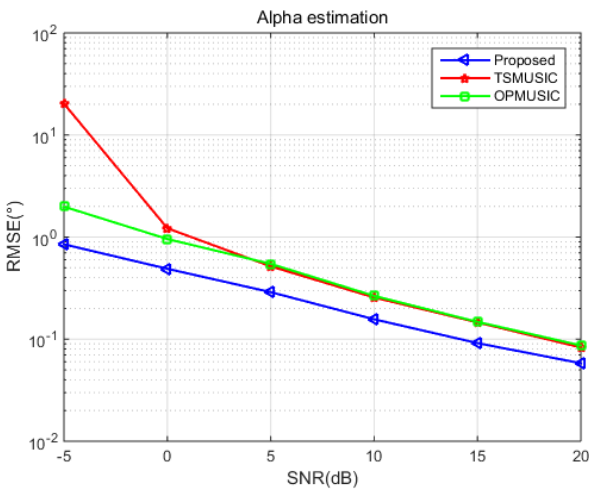
There are six uncorrelated NF sources from $\{20^\circ, 35^\circ, 0.2\lambda\}$, $\{40^\circ, 55^\circ, 0.25\lambda\}$, $\{60^\circ, 80^\circ, 0.3\lambda\}$, $\{80^\circ, 95^\circ, 0.35\lambda\}$, $\{100^\circ, 115^\circ, 0.4\lambda\}$ and $\{120^\circ, 135^\circ, 0.45\lambda\}$ impinging onto a symmetric cross array with $N_x = N_y = 3$, i.e., the total number of elements of the cross array is 5.



3. Simulation Results

Performance versus SNR

There are three uncorrelated NF sources impinging onto a symmetric cross array with $N_x = N_y = 5$, the number of snapshots, frames and Monte Carlo trials are set to be 1000, 100 and 500, respectively.



4. Conclusions

- **A new localization method for NF sources has been proposed using a cross array in this paper.**
 - **Make full use of the space-time two-dimensional characteristics of the signal to improve the degree of freedom of the algorithm, and use mathematical knowledge and dimensionality reduction technology to reduce the complexity of the algorithm.**
 - **This algorithm can realize direction finding under the condition of underdetermination, and can automatically match, achieving a good compromise between accuracy and complexity.**
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Thanks for your attention!