# CLOSED-FORM SINGLE SOURCE DIRECTION-OF-ARRIVAL ESTIMATOR USING FIRST-ORDER RELATIVE HARMONIC COEFFICIENTS

## **OBJECTIVES**

- All the existing relative harmonic coefficients (RHC) based direction-of-arrival (DOA) estimators suffer from resolution limitations, as they require searching over the DOA grid.
- This paper utilizes the first-order RHC to propose a closed-form DOA estimator by deriving a directional vector, which points towards to the desired source direction, thus circumventing the exhaustive search over directional space, while achieving equivalent localization accuracy.

# Spherical Harmonics DECOMPOSITION



Figure 1: DOA estimation using a spherical microphone array.

Decomposed into the spherical harmonics domain,  $P(\boldsymbol{x}_j, k) = \sum_{n=0}^{N} \sum_{m=-n}^{n} \alpha_{nm}(k) j_n(kr) Y_{nm}(\theta_j, \phi_j)$ 

 $\mathbf{1} \alpha_{nm}(\cdot)$ : spherical harmonic coefficient  $\mathbf{O}N = \lceil kr \rceil$ : truncated order of soundfield  $\mathfrak{S}_{j_n}(\cdot)$ : spherical Bessel function of the first kind •  $Y_{nm}(\cdot)$ : spherical harmonic function

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# Relative Harmonic COEFFICIENTS (RHC)

We define the RHC of order n and mode m as:

 $\beta_{nm}(k) = \frac{\alpha_{nm}(k)}{\alpha_{00}(k)}.$ 

Analytically, the feature can be expressed as:

 $\beta_{nm}(k) = 2\sqrt{\pi}i^n Y^*_{nm}(\vartheta_s, \varphi_s).$ 

(3)For the first-order microphone array, our feature vector is,

$$\begin{bmatrix} 2\sqrt{\pi}iY_{1,-1}^*(\vartheta_s,\varphi_s), 2\sqrt{\pi}iY_{1,0}^*(\vartheta_s,\varphi_s), 2\sqrt{\pi}iY_{1,1}^*(\vartheta_s,\varphi_s) \end{bmatrix}^I, \\ \text{Several advantages:} \tag{4}$$

Several advantages:

• It is easily estimated under noisy conditions.

- source-signal invariant, frequency-**2**It 1Sindependent and solely dependent on the source position (e.g., DOA).
- 3 It can be used to derive a closed-form DOA estimator (see below).

### A DIRECTIONAL VECTOR

Explicitly, the first-order RHC in (4) is,

$$\beta_{1,-1} = i\sqrt{3/2}\sin(\vartheta_s)e^{i\varphi_s}$$

$$= -\sqrt{3/2}\sin\vartheta_s\sin\varphi_s + i\sqrt{3/2}\sin\vartheta_s\cos\varphi_s$$

$$\beta_{1,0} = i\sqrt{3}\cos(\vartheta_s)$$

$$\beta_{1,1} = -i\sqrt{3/2}\sin(\vartheta_s)e^{-i\varphi_s}$$

$$= -\sqrt{3/2}\sin\vartheta_s\sin\varphi_s - i\sqrt{3/2}\sin\vartheta_s\cos\varphi_s.$$
(5)

*Theorem*: Denote the estimated first-order RHC as  $\beta_{1,-1}$ ,  $\beta_{1,0}$ , and  $\beta_{1,1}$ . Derive the *direction vector*:

$$\bar{\boldsymbol{I}} = \begin{bmatrix} \operatorname{Im}\{\bar{\beta}_{1,-1} - \bar{\beta}_{1,1}\} \\ \operatorname{Re}\{\bar{\beta}_{1,-1} + \bar{\beta}_{1,1}\} \\ \operatorname{Im}\{\bar{\beta}_{1,0}\} \end{bmatrix} \otimes \begin{bmatrix} \sqrt{1/6} \\ -\sqrt{1/6} \\ \sqrt{1/3} \end{bmatrix}$$
(6)

*Conclusion*: If the estimated first-order RHC coefficients are equal to their analytical values, the directional vector Ipoints towards the source direction, i.e.,  $(\vartheta_s, \varphi_s)$ . (Please see the proof in the paper)

# DIRECTIONAL VECTOR ESTIMATION PROCEDURE

The estimations of the directional vector comprise four steps:

• Measure the soundfield due to an unknown single sound source, and then transform the timedomain multichannel recordings into the shorttime Fourier transform domain.

• Decompose the multichannel STFT coefficients into the spherical harmonics domain and estimate the first-order spherical harmonic coefficients.

**3** Extract the first-order RHC and then apply frequency-smoothing over a wide frequency band, as the RHC may slightly differ from the frequency-independence property.

• Substitute the smoothed first-order RHC into the closed-form operation in (6), and normalize, as the estimated direction vector may deviate from the unit-norm property, and finally obtain a practical direction vector.

#### ALGORITHM PROPERTIES

• Usability with other microphone arrays: the algorithm is independent of the specific microphone constellation provided the array facilitates firstorder spherical harmonics decomposition.

• Computational-efficiency: (i) we applied time averaging and frequency smoothing, respectively, thus circumventing the need to localize the source for each time-frequency bin; and (ii) closed-form solution circumvents the tedious grid search.

Table 2: Various reverberation levels (SNR = 10 dB).

SR/MAEE	Reverberation time $(T_{60})$		
Methods	$150 \mathrm{ms}$	$350 \mathrm{ms}$	$550 \mathrm{ms}$
Decoupled	$100\%/1.05^{\circ}$	$99\%/3.56^{\circ}$	$88\%/4.69^{\circ}$
Proposed	$100\%/0.97^{\circ}$	$99\%/3.26^{\circ}$	$93\%/4.62^{\circ}$



 $|SR/\overline{N}|$ SNF 15 $\overline{25}$ 

**1** Baselines: (i) RHC-based decoupled DOA estimator [1] and (ii) the intensity-based method [2]. **2** Metrics: (i) success-ratio (SR/%) over the  $M_{\rm tot} = 100$  cases and (ii) average mean absolute estimated error  $(\overline{MAEE})$  over the successful cases. **3** Low-complexity: average time cost by the proposed method is 2.9 ms, while the decoupled approach takes 578 ms. See Figure 2 for tracking performance and more results in the paper. CONCLUSION

first-order relative harmonic coefficients. • Our algorithm achieves better localization accuracy and reduced complexity as compared with the baseline approaches.

**1**Y. Hu, et al. "Decoupled DOA Estimators using relative harmonic coefficients," in 2020 28th EUSIPCO, 246-250. **2**D. P. Jarrett, et al. "3D source localization in the spherical harmonic domain using a pseudointensity vector," in 2010 18th EUSIPCO, pp. 442-446.

EXPERIMENTAL RESULTS

Figure 2: DOA tracking using the proposed algorithm.

Table 1: Localization accuracy for various SNR levels.

MAEE	Localization methods			
a level	Intensity	Decoupled	Proposed	
dB	$82\%/3.54^{\circ}$	$100\%/0.35^{\circ}$	$100\%/0.33^{\circ}$	
dB	$98\%/1.21^{\circ}$	$100\%/0.17^{\circ}$	$100\%/0.12^{\circ}$	
dB	$100\%/0.33^{\circ}$	$100\%/0.15^{\circ}$	$100\%/0.07^{\circ}$	

**1** Proposed a closed-form DOA estimator using the

#### REFERENCE