



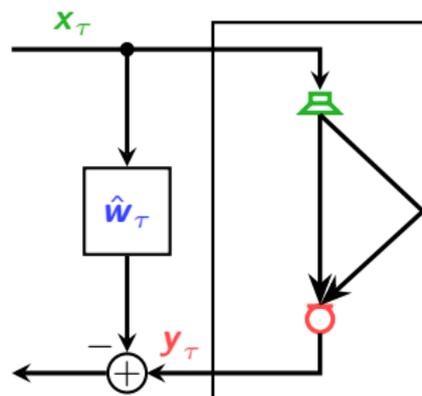
End-To-End Deep Learning-Based Adaptation Control for Frequency-Domain Adaptive System Identification

T. Haubner, A. Brendel, and W. Kellermann

Chair of Multimedia Communications and Signal Processing

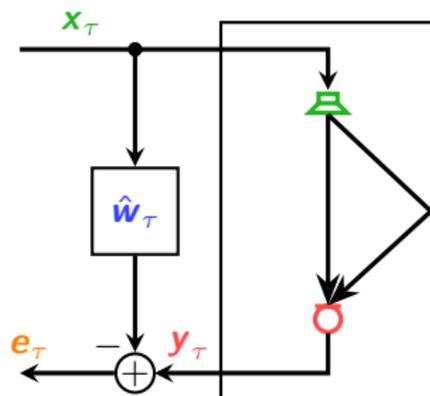
Motivation: Acoustic Echo Cancellation

- **Problem:** Identify **acoustic transfer function (ATF)** between **loudspeaker** and **microphone** signal



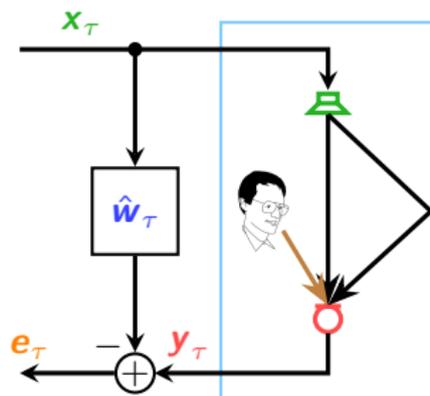
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- ▶ **Approach:**
 - ▶ Minimization of **error** signal power
 - ▶ Iterative update of ATF estimate:
$$\hat{w}_\tau = \hat{w}_{\tau-1} + \delta \hat{w}_\tau$$



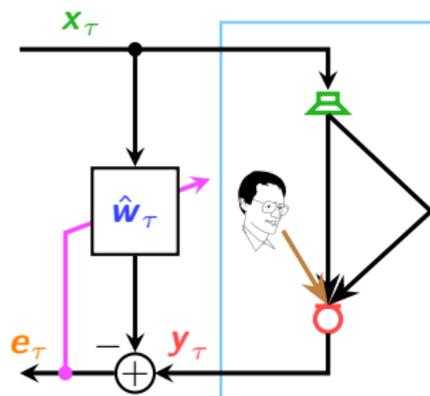
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 - ▶ **Interfering signals**, e.g., local speech or noise
 - ▶ Time-varying **acoustic environments**



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Robust **adaptation control** for improved convergence rate

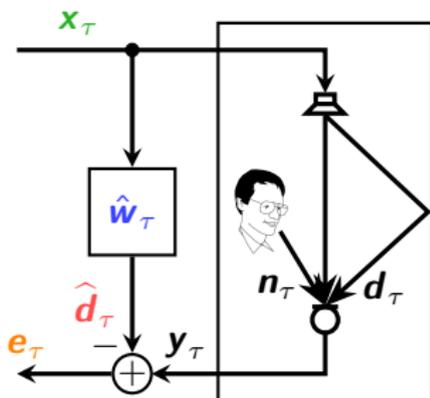
Outline

- ▶ Acoustic System Identification
- ▶ Proposed DNN-Based Adaptation Control
- ▶ Experimental Evaluation
- ▶ Conclusion

Iterative ATF Estimation

$$\hat{d}_\tau = Q(x_\tau \odot \hat{w}_{\tau-1})$$

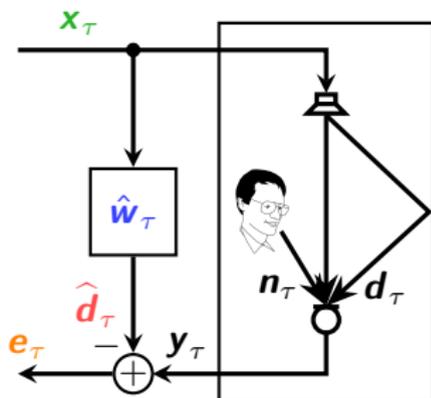
1. Estimate echo \hat{d}_τ by linear convolution of
 - ▶ loudspeaker signal block x_τ
 - ▶ with ATF estimate $\hat{w}_{\tau-1}$.
 - ▶ and linear convolution constraint matrix Q



Iterative ATF Estimation

$$\mathbf{e}_\tau = \mathbf{y}_\tau - \hat{\mathbf{d}}_\tau = \mathbf{y}_\tau - \mathbf{Q}(\mathbf{x}_\tau \odot \hat{\mathbf{w}}_{\tau-1})$$

1. Estimate echo $\hat{\mathbf{d}}_\tau$ by linear convolution of
 - ▶ loudspeaker signal block \mathbf{x}_τ
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2. Compute error signal block \mathbf{e}_τ

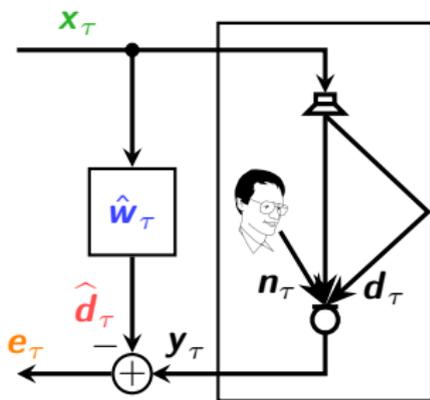


Iterative ATF Estimation

$$e_{\tau} = y_{\tau} - \hat{d}_{\tau} = y_{\tau} - \mathbf{Q}(\mathbf{x}_{\tau} \odot \hat{\mathbf{w}}_{\tau-1})$$

$$\hat{\mathbf{w}}_{\tau} = \hat{\mathbf{w}}_{\tau-1} + \mathbf{G} (\mathbf{x}_{\tau}^* \odot e_{\tau})$$

1. Estimate echo \hat{d}_{τ} by linear convolution of
 - ▶ loudspeaker signal block \mathbf{x}_{τ}
 - ▶ with ATF estimate $\hat{\mathbf{w}}_{\tau-1}$.
 - ▶ and linear convolution constraint matrix \mathbf{Q}
2. Compute error signal block e_{τ}
3. Compute stochastic gradient $\mathbf{G} (\mathbf{x}_{\tau}^* \odot e_{\tau})$ with FIR filter projection matrix \mathbf{G}
4. Perform gradient descent



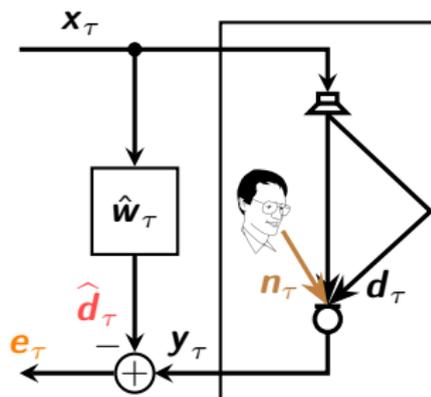
Adaptation Control

$$\mathbf{e}_\tau = \mathbf{y}_\tau - \hat{\mathbf{d}}_\tau = \mathbf{n}_\tau + (\mathbf{d}_\tau - \hat{\mathbf{d}}_\tau)$$

$$\hat{\mathbf{w}}_\tau = \hat{\mathbf{w}}_{\tau-1} + \mathbf{G} (\mathbf{x}_\tau^* \odot \mathbf{e}_\tau)$$

Large error powers $\|\mathbf{e}_\tau\|^2$ could result from

- ▶ system mismatch, i.e., imprecise echo estimates $\hat{\mathbf{d}}_\tau$
- ▶ interfering signals \mathbf{n}_τ



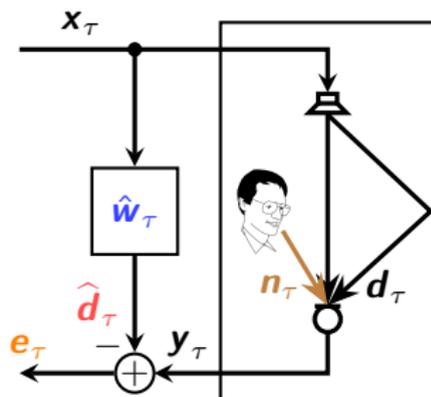
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Large error powers $\|\mathbf{e}_\tau\|^2$ could result from

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Adaptation Control

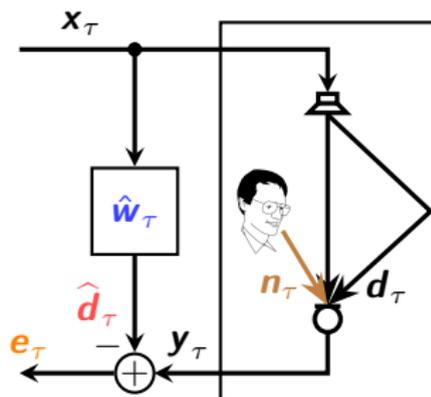
$$\mathbf{e}_\tau = \mathbf{y}_\tau - \hat{\mathbf{d}}_\tau = \mathbf{n}_\tau + (\mathbf{d}_\tau - \hat{\mathbf{d}}_\tau)$$

$$\hat{\mathbf{w}}_\tau = \hat{\mathbf{w}}_{\tau-1} + \mathbf{G}\Lambda_\tau(\mathbf{x}_\tau^* \odot \mathbf{e}_\tau)$$

Large error powers $\|\mathbf{e}_\tau\|^2$ could result from

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Solution: Control adaptation by time- and frequency-dependent step-sizes $[\Lambda_\tau]_{ff}$



Traditional Model-Based Adaptation Control

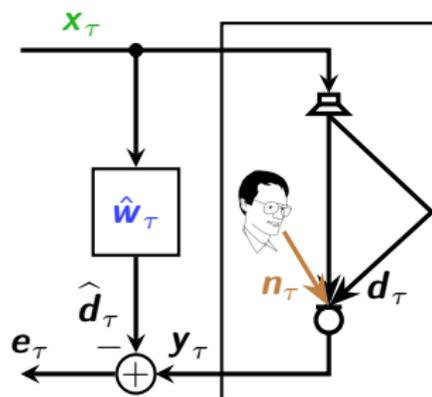
$$\Lambda_{\tau}^{\text{MB}} = f_{\text{MB}} \left(\psi_{\tau}^{\text{XX}}, \psi_{\tau}^{\text{NN}}, \psi_{\tau}^{\Delta W \Delta W}, \dots \right)$$

Compute step-size matrix $\Lambda_{\tau}^{\text{MB}}$ as a function of

- ▶ loudspeaker PSD ψ_{τ}^{XX}
- ▶ interference PSD ψ_{τ}^{NN}
- ▶ filter estimation uncertainty $\psi_{\tau}^{\Delta W \Delta W}$
- ▶ ...

Prominent examples:

FDAF, DFT-domain Kalman filter, ...

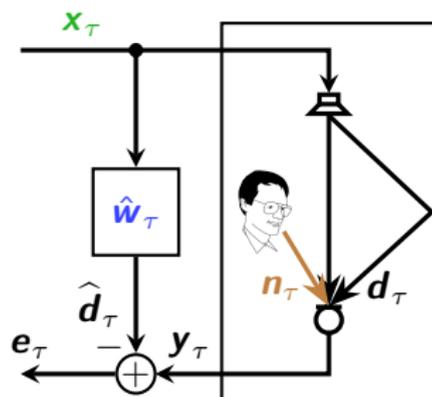


Traditional Model-Based Adaptation Control

$$\Lambda_{\tau}^{\text{MB}} = f_{\text{MB}} \left(\psi_{\tau}^{\text{XX}}, \psi_{\tau}^{\text{NN}}, \psi_{\tau}^{\Delta W \Delta W}, \dots \right)$$

Challenges

- ▶ Estimation of signal statistics of unknown quantities, e.g., ψ_{τ}^{NN}
- ▶ Mismatch of assumed model properties



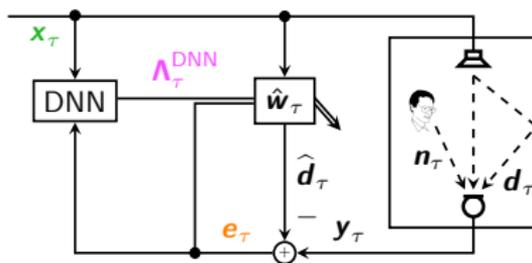
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Proposed General Concept

$$\Lambda_{\tau}^{\text{DNN}} = f_{\text{DNN}}(\mathbf{x}_1, \mathbf{e}_1, \dots, \mathbf{x}_{\tau}, \mathbf{e}_{\tau}; \theta)$$

- ▶ Learn mapping f_{DNN} of observed signal sequences to step-size matrices
- ▶ DNN parameters θ are learned from training data



Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = m_{f,\tau}^{\mu}$$

DNN provides raw step-size $m_{f,\tau}^{\mu} \in [0, \mu_{\text{MAX}}]$

Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = m_{f,\tau}^{\mu}$$

Challenges: DNN needs to model

- ▶ large numerical range due to non-whiteness of loudspeaker signals

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Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{\text{XX}}}$$

Challenges: DNN needs to model

- ▶ large numerical range due to non-whiteness of loudspeaker signals

Normalize raw step-size $m_{f,\tau}^{\mu}$ by **loudspeaker PSD** estimate $\hat{\Psi}_{f,\tau}^{\text{XX}}$

Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{XX}}$$

Challenges: DNN needs to model

- ▶ large numerical range due to non-whiteness of loudspeaker signals
- ▶ rapid changes due to non-stationarity of interfering signals

Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{\text{XX}} + \hat{\Psi}_{f,\tau}^{\text{II}}}$$

Challenges: DNN needs to model

- ▶ large numerical range due to non-whiteness of loudspeaker signals
- ▶ rapid changes due to non-stationarity of interfering signals

Traditional Approach: Normalization by interference PSD estimate $\hat{\Psi}_{f,\tau}^{\text{II}}$

Proposed Step-Size Structure

$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{\text{XX}} + \frac{M}{R} |m_{f,\tau}^e [\mathbf{e}_{\tau}]_f|^2}$$

Challenges: DNN needs to model

- ▶ large numerical range due to non-whiteness of loudspeaker signals
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Normalization by **masked error power** $|m_{f,\tau}^e [\mathbf{e}_{\tau}]_f|^2$

Proposed Step-Size Structure

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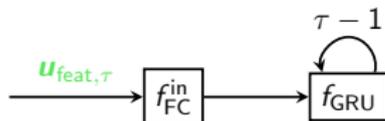
Exploitation of **domain knowledge** from traditional adaptation control

DNN Architecture



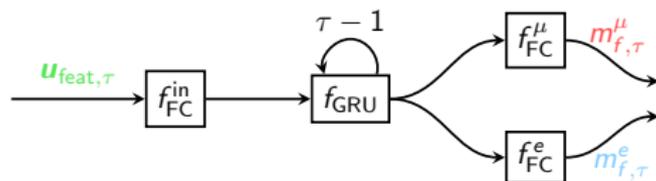
- ▶ Feature vector $u_{\text{feat}, \tau}$
 - ▶ Log. loudspeaker power spectrum
 - ▶ Log. error power spectrum
- ▶ Extract temporal information by GRU layers
- ▶ DNN provides masks $m_{f, \tau}^{\mu} \in [0, \mu_{\text{MAX}}]$ and $m_{f, \tau}^{\epsilon} \in [0, 1]$
- ▶ Time- and frequency-dependent step-sizes $[\Lambda_{\tau}^{\text{DNN}}]_{ff}$

DNN Architecture



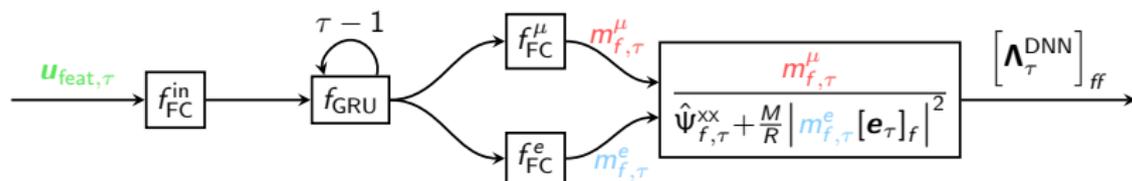
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DNN Architecture

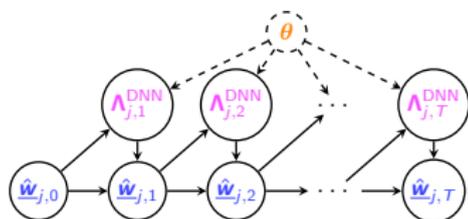


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DNN Training

Challenges:

- ▶ Choice of optimum target step-sizes
- ▶ Cost function design



Proposed Solution

End-to-end training of DNN parameters θ w.r.t. achieved system identification performance

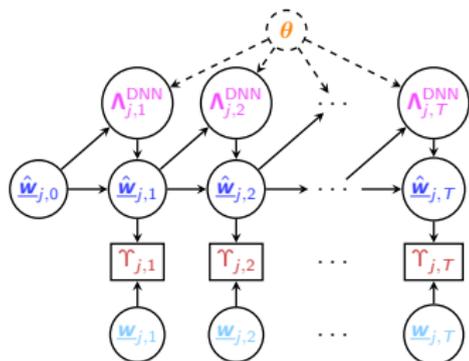
Cost function:

$$\mathcal{J}(\theta) = \frac{1}{TJ} \sum_{j=1}^J \sum_{\tau=1}^T 10 \log_{10} (\Upsilon_{j,\tau})$$

with normalized system distance

$$\Upsilon_{j,\tau} = \frac{\|\underline{w}_{j,\tau} - \hat{\underline{w}}_{j,\tau}\|_2^2}{\|\underline{w}_{j,\tau}\|_2^2}$$

j : sequence index, τ : block index

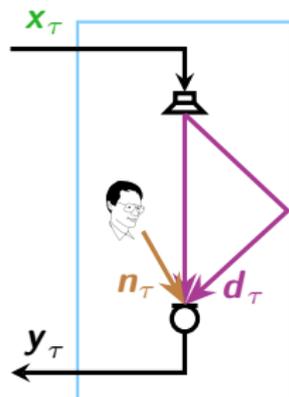


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- ▶ **Experimental Evaluation**
- ▶ Conclusion

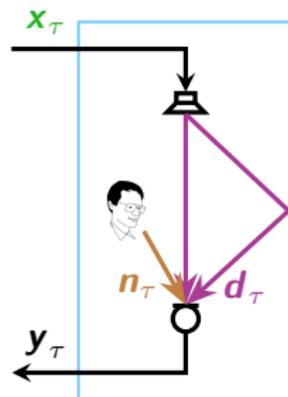
Experimental Evaluation: AEC Application

- ▶ **Loudspeaker signal:** 143 different speakers
- ▶ **Acoustic environment:** 201 measured AIRs
 - ▶ Sampling frequency: $f_s = 16$ kHz
 - ▶ Reverberation time $T_{60} \in [120\text{ms}, 780\text{ms}]$
 - ▶ Acoustic scene change in the interval $[7.2\text{s}, 8.8\text{s}]$
- ▶ **Interfering signal**
 - ▶ 145 different speakers (echo-to-near-end power ratio $\in [-10\text{dB}, 10\text{dB}]$)
 - ▶ White Gaussian noise (echo-to-noise power ratio $\in [25\text{dB}, 35\text{dB}]$)



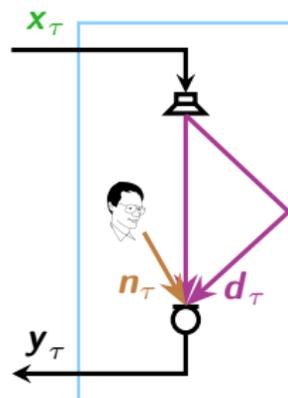
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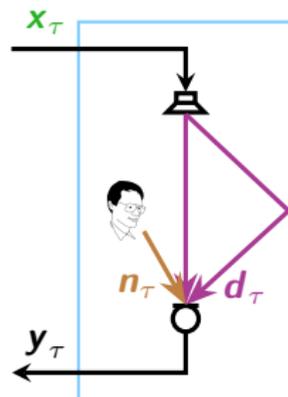
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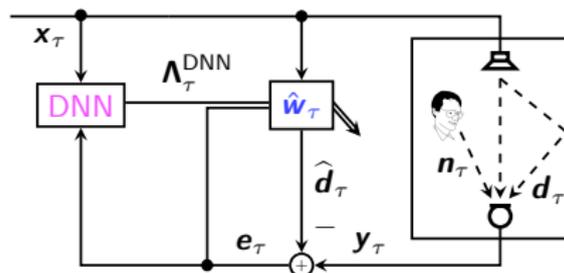
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Separation into disjoint training and testing data sets

Algorithmic Settings

- ▶ **FIR filter** length: 2048 samples
- ▶ Frame shift: 1024 samples
- ▶ **#DNN** parameters: 2.4 million
- ▶ Training optimizer: *Adam*



Performance Measures

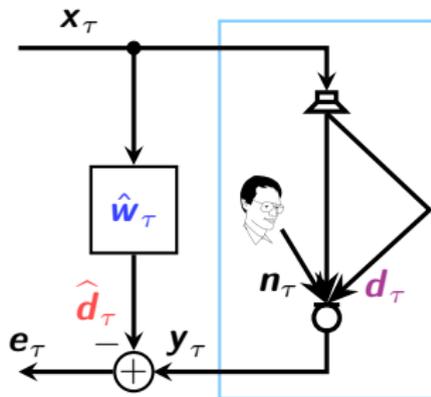
- ▶ **Normalized system distance** (the lower the better)

$$\Upsilon_{ZP,\tau} = 10 \log_{10} \frac{\|\tilde{\underline{w}}_{\tau} - \mathbf{V} \hat{\underline{w}}_{\tau}\|_2^2}{\|\tilde{\underline{w}}_{\tau}\|_2^2}$$

with \mathbf{V} being a zero-padding matrix

- ▶ **ERLE** (the higher the better)

$$\mathcal{E}_{\tau} = 10 \log_{10} \frac{\mathbb{E} [\|\underline{d}_{\tau}\|_2^2]}{\mathbb{E} [\|\underline{d}_{\tau} - \hat{\underline{d}}_{\tau}\|_2^2]}$$



Performance Measures

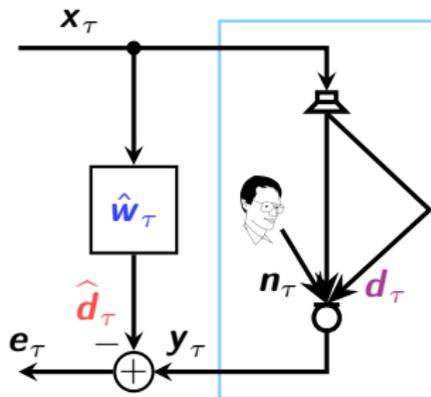
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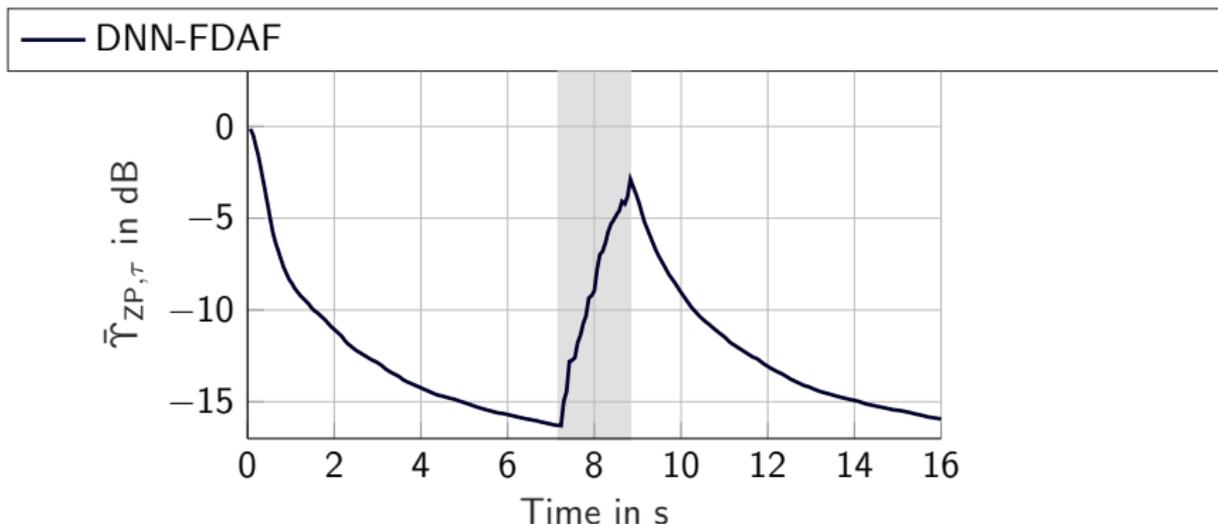
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Arithmetic average of **100** experiments with randomly-selected loudspeaker and interfering signals, AIRs and transition times.

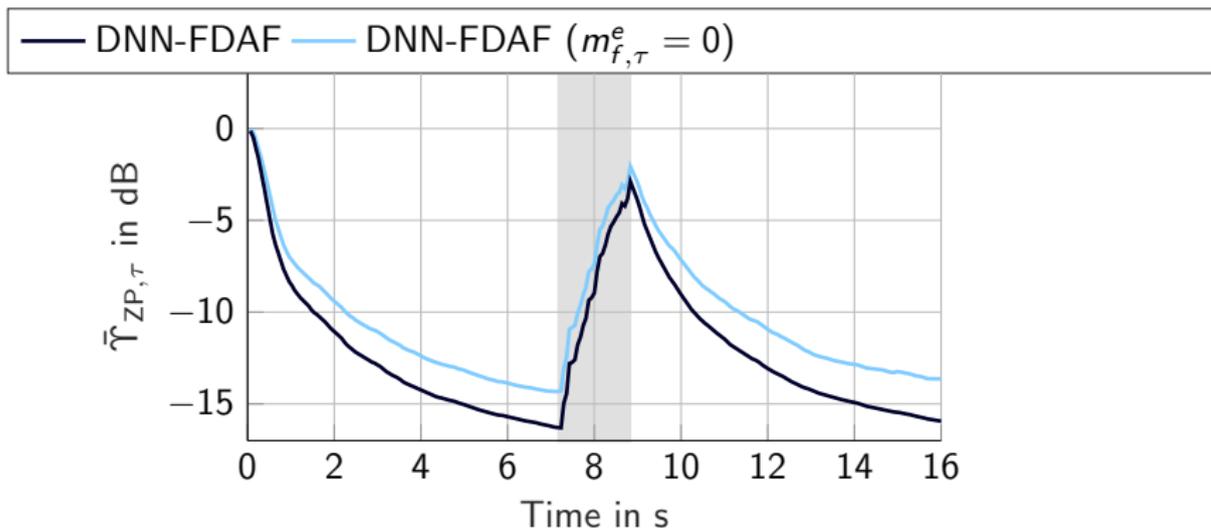
Analysis of Proposed DNN-FDAF



$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{XX} + \frac{M}{R} \left| m_{f,\tau}^e [\mathbf{e}_{\tau}]_f \right|^2}$$

Proposed step-size control

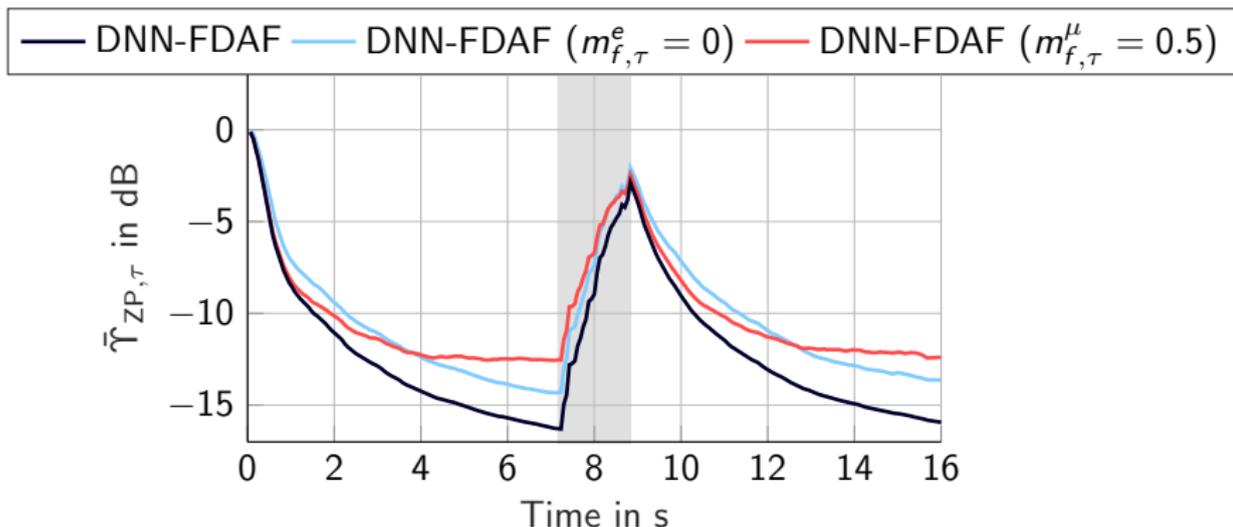
Analysis of Proposed DNN-FDAF



$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{m_{f,\tau}^{\mu}}{\hat{\Psi}_{f,\tau}^{XX} + 0}$$

Discarding error power normalization

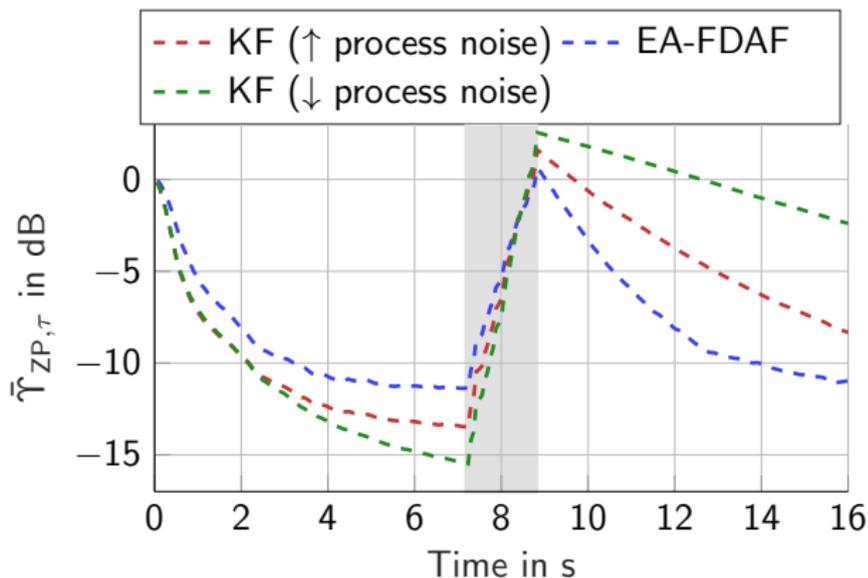
Analysis of Proposed DNN-FDAF



$$\left[\Lambda_{\tau}^{\text{DNN}} \right]_{ff} = \frac{0.5}{\hat{\Psi}_{f,\tau}^{XX} + \frac{M}{R} \left| m_{f,\tau}^e [\mathbf{e}_{\tau}]_f \right|^2}$$

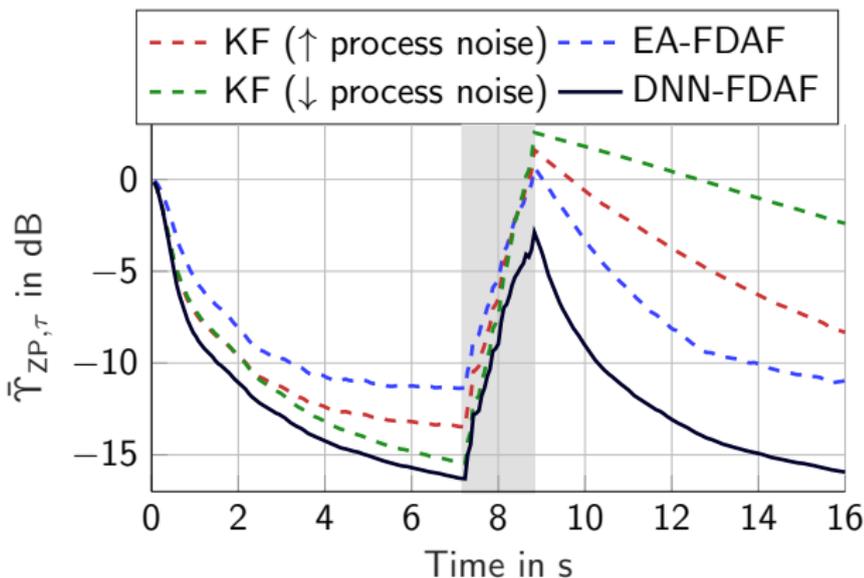
Static and frequency-independent
raw step-size

System Identification Performance



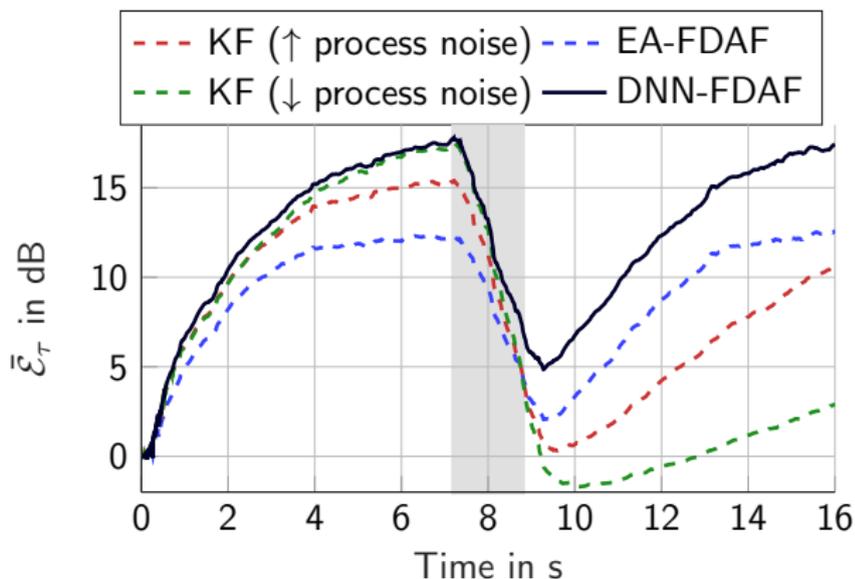
Steady-state and tracking performance trade-off

System Identification Performance



DNN-FDAF: Fast convergence and high steady-state performance

Echo Cancellation Performance



DNN-FDAF: Improved echo cancellation performance

Conclusion

Summary

- ▶ Novel adaptation control for online system identification by using **DNN-based step-size inference**
- ▶ **End-to-end optimization** of DNN parameters w.r.t. average system identification performance

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- ▶ Extension to **unsupervised system identification** applications

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Thank you for watching!