Graph Filter Banks with *M*-Channels, Maximal Decimation, and Perfect Reconstruction

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Caltech

Outline

Introduction to Graph Signal Processing Graph Signals

- 2 Purpose: *M*-Channel Filter Banks
 - Extending Classical Tools to Graph Case
- 3 Multirate Processing of Graph Signals
 - Decimation & Expansion
 - M-Block Cyclic Graphs
 - Concept of Spectrum Folding
 - More Results for M-Block Cyclic

4 Some Examples

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4 Some Examples

What is a Graph Signal ?



¹A. Sandryhaila and J. M. F. Moura. "Discrete Signal Processing on Graphs: Frequency Analysis". In: *IEEE Trans. Signal Process.* 62.12 (2014), pp. 3042–3054.

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A is considered as the graph operator, with $A = V \Lambda V^{-1}$

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What is a Graph Signal ?



A is the adjacency matrix^{1,2,3}, $A \in \mathcal{M}^N$

A is considered as the graph operator, with $A = V \Lambda V^{-1}$

V = graph Fourier Basis, V^{-1} = graph Fourier Transform

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Maximally Decimated *M*-Channel Filter Banks



M-Channel Filter Banks (Brute-Force)



M-Channel Filter Banks (Brute-Force)



M-Channel Filter Banks (Brute-Force)





 $F_{syn} H_{anal} = I$



$$\boldsymbol{F}_{syn} \ \boldsymbol{H}_{anal} = \boldsymbol{I}$$

$$oldsymbol{F}_{syn}=oldsymbol{H}_{anal}^{-1}$$



$$\boldsymbol{F}_{syn} \boldsymbol{H}_{anal} = \boldsymbol{I}$$

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Filter Design \$ Matrix Inversion Problem



Relation with A?

$$F_{syn} H_{anal} = I$$

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Filter Design \$ Matrix Inversion Problem



Relation with A?

$$\boldsymbol{F}_{syn} \; \boldsymbol{H}_{anal} = \boldsymbol{I}$$

$$\boldsymbol{F}_{syn} = \boldsymbol{H}_{anal}^{-1}$$

Computational Complexity = $O(N^2)$

Filter Design ↓ Matrix Inversion Problem

 $\boldsymbol{H}_{k}(\boldsymbol{A}) = h_{k}(0) \, \boldsymbol{A}^{0} + h_{k}(1) \, \boldsymbol{A}^{1} + h_{k}(2) \, \boldsymbol{A}^{2} + \ldots + h_{k}(L) \, \boldsymbol{A}^{L}$

$$\boldsymbol{H}_k(\boldsymbol{A}) = h_k(0) \, \boldsymbol{A}^0 + h_k(1) \, \boldsymbol{A}^1 + h_k(2) \, \boldsymbol{A}^2 + \ldots + h_k(L) \, \boldsymbol{A}^L$$



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Cost: $LN^2 + LN$

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Cost of $H_k(A)$ is O(LN) v.s. $O(N^2)$ in brute-force $L \ll N$

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 $\boldsymbol{D}: C^N \to C^{N/M}$

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Dx

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Definition (Canonical Decimator)

$$\boldsymbol{D} = \begin{bmatrix} \boldsymbol{I}_{N/M} & \boldsymbol{0}_{N/M} & \cdots & \boldsymbol{0}_{N/M} \end{bmatrix} \in \mathcal{C}^{(N/M) \times N},$$

which retains the first N/M samples of the given graph signal.

 $\boldsymbol{U}: C^{N/M} \to C^N$

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Given the decimator $\boldsymbol{D} = \begin{bmatrix} \boldsymbol{I}_{N/M} & \boldsymbol{0}_{N/M} & \cdots & \boldsymbol{0}_{N/M} \end{bmatrix}$

$$oldsymbol{U} = oldsymbol{D}^T = \left[egin{array}{c} oldsymbol{I}_{N/M} \ oldsymbol{0}_{N/M} \ dots \ oldsymbol{0}_{N/M} \end{array}
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Upsample-then-downsample, DU, is unity DU = I

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$$\boldsymbol{D}^T = rgmin_{\boldsymbol{U}} \| \boldsymbol{U} \|_F$$
 s.t. $\boldsymbol{D}\boldsymbol{U} = \boldsymbol{I}$

M-block cyclic graphs

^aO. Teke and P. P. Vaidyanathan. "Extending Classical Multirate Signal Processing Theory to Graphs". *IEEE Trans. Signal Process. (Submitted). b*

M-block cyclic graphs

$$oldsymbol{A} = egin{bmatrix} oldsymbol{0} & \cdots & oldsymbol{0} & oldsymbol{A}_M \ oldsymbol{A}_1 & \cdots & oldsymbol{0} &$$

(Under suitable permutation)

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$$\boldsymbol{A}_{j} \in \mathcal{M}^{N/M}$$
(Under suitable permutation)

M-block cyclic graphs

$$A = \begin{bmatrix} \mathbf{0} & \cdots & \mathbf{0} & A_M \\ A_1 & \cdots & \mathbf{0} & \mathbf{0} \\ \vdots & \ddots & \mathbf{0} & \vdots \\ \mathbf{0} & \cdots & A_{M-1} & \mathbf{0} \end{bmatrix} \in \mathcal{M}^N$$

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1 If a graph is *M*-block cyclic, then it is *M*-partite, but not vice-versa.

M-Block Cyclic Graphs

M-block cyclic graphs

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- 2 A graph is 2-block cyclic if and only if it is bi-partite.

M-Block Cyclic Graphs

M-block cyclic graphs



- 1 If a graph is M-block cyclic, then it is
- 2 A graph is 2-block cyclic if and only if
- 3 An M-block cyclic graph is necessarily a directed graph for M > 2.

M-Block Cyclic Graphs

$M\operatorname{-block}$ cyclic graphs



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- A cyclic graph of size N, C_N, is an M-block cyclic graph for all M that divides N.

$M\operatorname{-Block}\,\operatorname{Cyclic}\,\operatorname{Graphs}$

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- 5 Unique eigenvalue & eigenvector structure ^{*a b*}

^aO. Teke and P. P. Vaidyanathan. "Extending Classical Multirate Signal Processing Theory to Graphs". *IEEE Trans. Signal Process. (Submitted).* ^bD. S. Watkins. "Product eigenvalue problems" *SIAM Review*, 2005

s Concept of Spectrum Folding

Spectrum Folding - Aliasing

$$\boldsymbol{y} = \boldsymbol{D}^T \, \boldsymbol{D} \, \boldsymbol{x}$$

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Concept of Spectrum Folding

Spectrum Folding - Aliasing



$$\boldsymbol{y} = \boldsymbol{D}^T \, \boldsymbol{D} \, \boldsymbol{x}$$



What is the relation between \hat{x} and \hat{y} ?

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 $\hat{y} = V^{-1}D^TDV \hat{x}$



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No "simple" relation in general!

Teke & Vaidyanathan











⁴S.K. Narang and A. Ortega. "Perfect Reconstruction Two-Channel Wavelet Filter Banks for Graph Structured Data". *IEEE Trans. Signal Process.* 60.6 (2012), pp. 2786–2799.

Teke & Vaidyanathan













More Results for $M\operatorname{-Block}\nolimits\operatorname{Cyclic}\nolimits$

Interpolation Filter on *M*-Block Cyclic Graphs

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Not true for an arbitrary x !

For k^{th} -band-limited signals, x can be recovered from Dx

M-Channel Filter-Banks on *M*-Block Cyclic Graphs

M-Channel Filter-Banks on *M*-Block Cyclic Graphs










$$\boldsymbol{x} \xrightarrow{H_0(\boldsymbol{A})} \overbrace{\boldsymbol{D}} \xrightarrow{\boldsymbol{D}} \overbrace{\boldsymbol{D}}^T \overbrace{F_0(\boldsymbol{A})} \xrightarrow{+} \xrightarrow{\boldsymbol{y}} \xrightarrow{\boldsymbol{y}} \xrightarrow{\boldsymbol{y}} \xrightarrow{\boldsymbol{H}_{M-1}(\boldsymbol{A})} \xrightarrow{+} \overbrace{\boldsymbol{D}} \xrightarrow{\boldsymbol{D}} \xrightarrow{\boldsymbol{D}} \xrightarrow{\boldsymbol{D}} \xrightarrow{\boldsymbol{F}_{M-1}(\boldsymbol{A})} \xrightarrow{+} \xrightarrow{\boldsymbol{y}} \xrightarrow{\boldsymbol{$$

 $\boldsymbol{F}_k = M \boldsymbol{H}_k$

5

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 $\boldsymbol{H}_{k-1} = \boldsymbol{V} \left(\boldsymbol{I} \otimes \boldsymbol{e}_k \boldsymbol{e}_k^T \right) \boldsymbol{V}^{-1}$

 $\boldsymbol{F}_{k} = M \boldsymbol{H}_{k}$ $\boldsymbol{H}_{k-1} = \boldsymbol{V} \left(\boldsymbol{I} \otimes \boldsymbol{e}_{k} \boldsymbol{e}_{k}^{T} \right) \boldsymbol{V}^{-1} \qquad H_{k}(e^{j\omega}) = \begin{cases} 1, & \frac{2\pi k}{M} \leq \omega \leq \frac{2\pi (k+1)}{M}, \\ 0, & \text{otherwise,} \end{cases}$

6

 $\boldsymbol{F}_k = M \boldsymbol{H}_k$

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$$\boldsymbol{A} \text{ needs to have distinct eigenvalues.}^{5 \ 6}$$

⁵O. Teke and P. P. Vaidyanathan. "Extending Classical Multirate Signal Processing Theory to Graphs – Part I: Fundamentals". *IEEE Trans. Signal Process. (Submitted).*

⁶A. Sandryhaila and J. M. F. Moura. "Discrete Signal Processing on Graphs". *IEEE Trans. Signal Process.* 61.7 (2013)

M-Block Cyclic ⁷
$$\rightleftharpoons$$

 $\begin{cases}
\text{Eigenvector Property}: \quad v_{i,j+k} = \Omega^k \ v_{i,j} \\
\text{Eigenvalue Property}: \quad \lambda_{i,j+k} = w^k \ \lambda_{i,j}
\end{cases}$

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For any polynomial H(A) $H(QAQ^{-1}) = Q H(A) Q^{-1}$ for any invertible Q

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$$x \longrightarrow H_0(A) \xrightarrow{\widetilde{D}} \widetilde{D} \longrightarrow \widetilde{U} \xrightarrow{F_0(A)} \xrightarrow{+} y$$

$$\vdots \vdots \vdots \vdots \vdots$$

$$H_{M-1}(A) \xrightarrow{\widetilde{D}} \widetilde{D} \longrightarrow \widetilde{U} \xrightarrow{F_{M-1}(A)}$$

\widetilde{D} and \widetilde{U} have higher complexity, but no restrictive assumptions on A.

⁷O. Teke and P. P. Vaidyanathan. "Extending Classical Multirate Signal Processing Theory to Graphs – Part I: Fundamentals". *IEEE Trans. Signal Process. (Submitted).*

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Conclusions

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- Decimator
- M-Block Cyclic Graph
 - Unique Eigenvalue-Eigenvector Structure
- Spectrum Folding
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 - Interpolation
- M-Channel Filter Banks
- Further Directions
 - How does this compare with alternative ways?
 - From ideal to non-ideal?

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Any questions?

Please email me: oteke@caltech.edu