

CONTRIBUTIONS

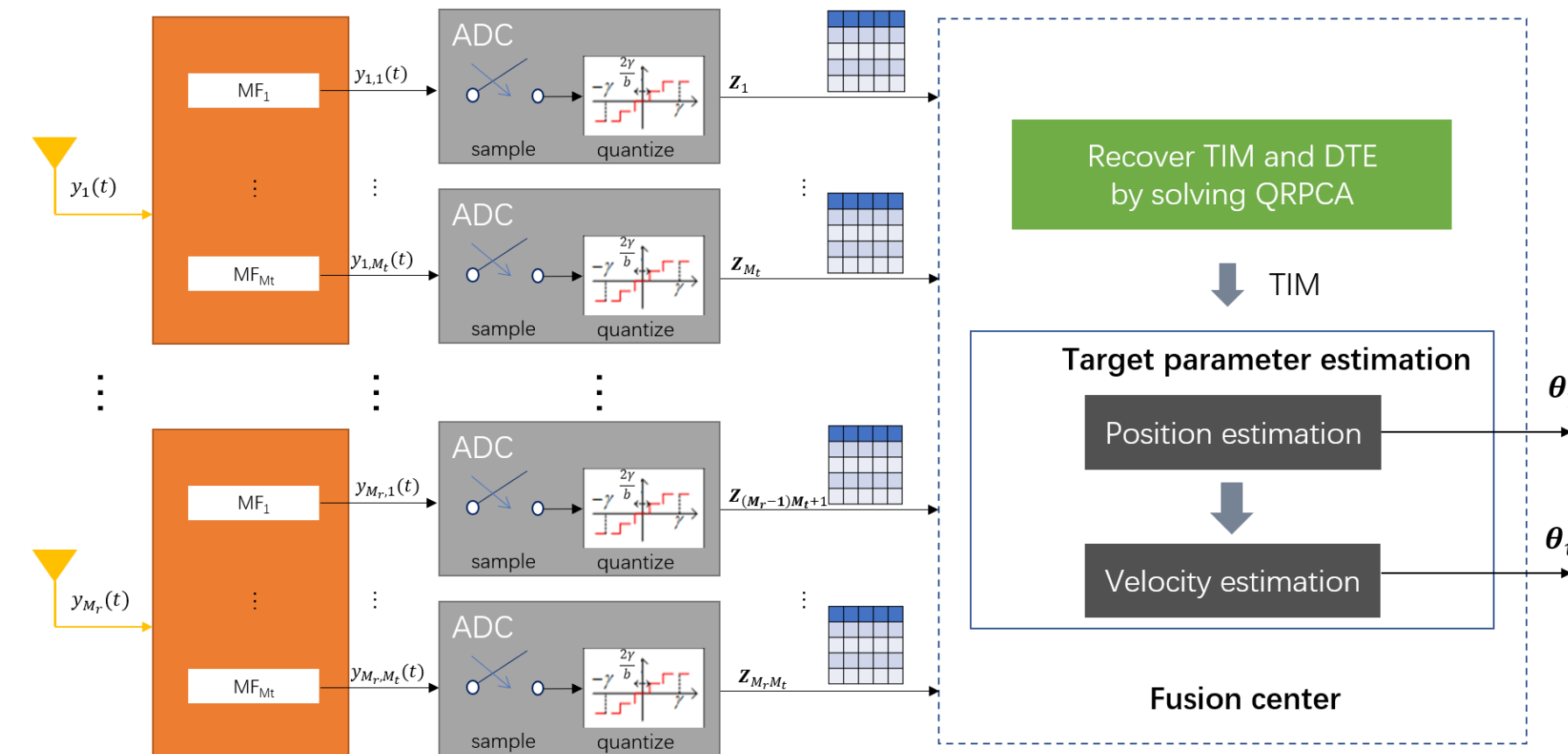
□ **Motivation:** Reduce the cost, energy consumption, and system complexity of distributed MIMO radar systems

□ **Key idea:**

- Using low-resolution ADCs to reduce the data transmission volume
- Formulate a QRPCA problem to recover the infinite-precision data
- Develop an APG-based algorithm to solve the QRPCA problem

□ **Result:** Develop a low-bit quantized distributed MIMO radar (LiQuiD-MIMO radar)

Low-resolution ADCs + Data recovery + Parameter Estimation



LiQuiD-MIMO Radars Model

Signal Model

□ Time delay

$$\tau_{mn}^{(k)} = \frac{\|\mathbf{p}^{(k)} - \mathbf{p}_t^{(k)}\| + \|\mathbf{p}^{(k)} - \mathbf{p}_r^{(k)}\|}{c}$$

□ Doppler frequency

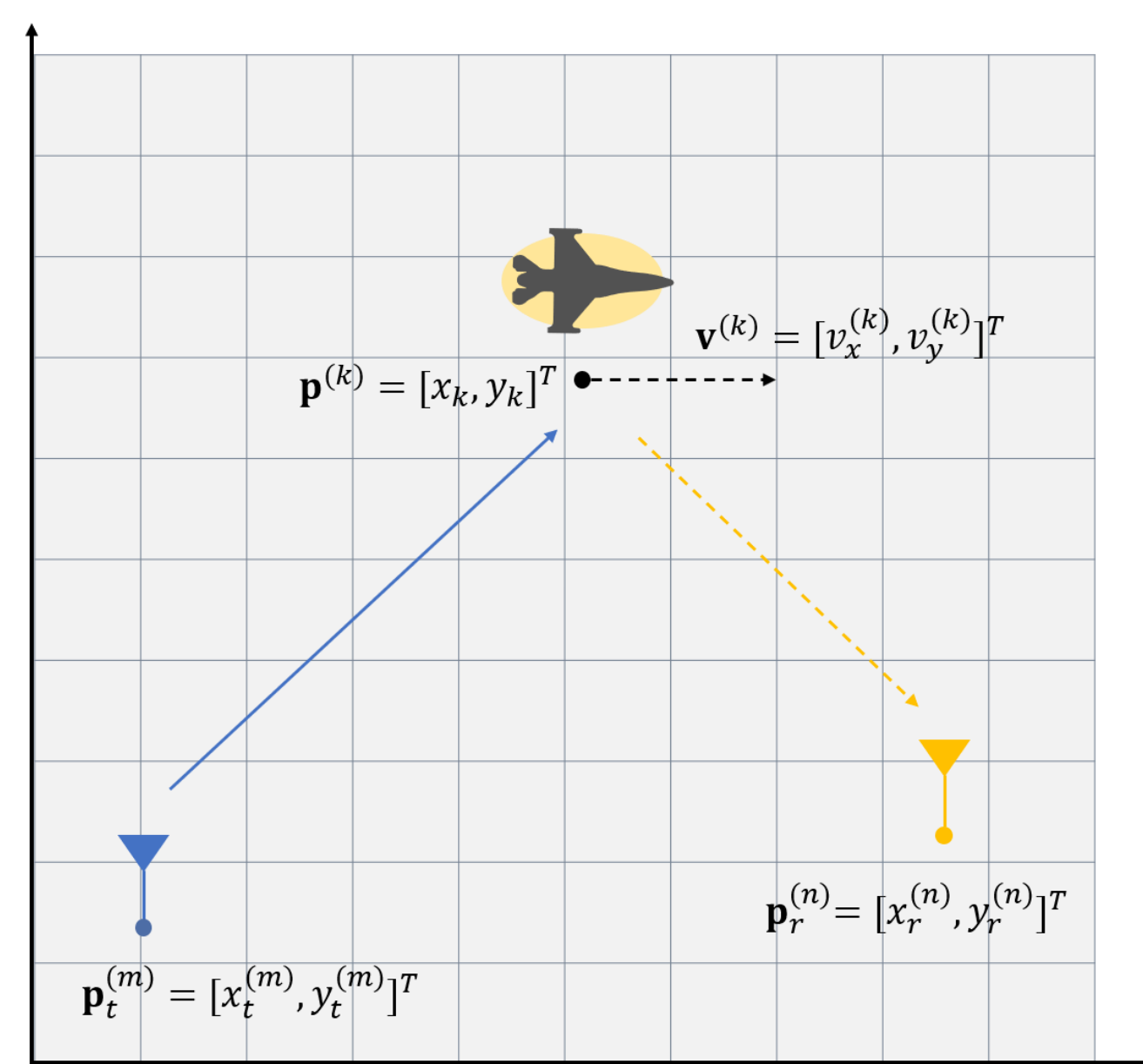
$$f_{mn}^{(k)} = \frac{f_m}{c} \left(\frac{\langle \mathbf{v}^{(k)}, \mathbf{p}^{(k)} - \mathbf{p}_t^{(m)} \rangle}{\|\mathbf{p}^{(k)} - \mathbf{p}_t^{(m)}\|} + \frac{\langle \mathbf{v}^{(k)}, \mathbf{p}^{(k)} - \mathbf{p}_r^{(m)} \rangle}{\|\mathbf{p}^{(k)} - \mathbf{p}_r^{(m)}\|} \right)$$

□ Received signal:

$$y_{mn}(t) = \sum_{q=0}^{Q-1} \sum_{k=1}^K \beta_{mn}^{(k)} s_m(t - \tau_{mn}^{(k)} - qT_{PRI}) e^{j2\pi f_{mn}^{(k)} qT_{PRI}} + w_{mn}(t)$$

□ $s_m(t)$ could be FDMA waveforms

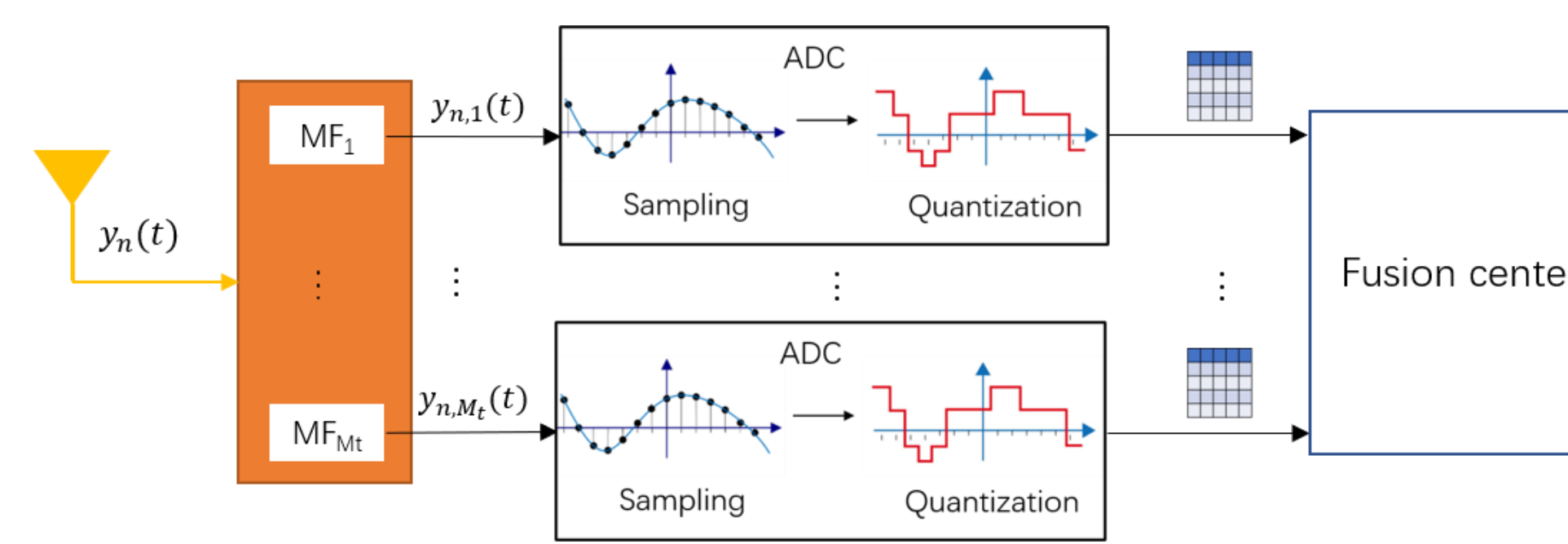
□ Task: Resolve the K position and velocity pairs $\{\mathbf{p}^{(k)}, \mathbf{v}^{(k)}\}_{k=1}^K$ from received signals.



Sampling and Quantization with Low-Resolution ADCs

- Using the low-resolution ADCs: each data is quantized into \tilde{b} bits, e.g., $\tilde{b} = 2, 3, 4$;
- Send quantized data to fusion center.
- \mathbf{X}_{mn} : Target information matrix (TIM); \mathbf{W}_{mn} : White Gaussian Noise (WGN); $\tilde{\mathbf{T}}_{mn}$: Data transmission error (DTE).

$$\mathbf{Y}_{mn} = \mathbf{X}_{mn} + \mathbf{W}_{mn} \rightarrow \tilde{\mathbf{Z}}_{mn} = \mathcal{Q}_C^{y,b}(\mathbf{X}_{mn} + \mathbf{W}_{mn}) \rightarrow \mathbf{Z}_{mn} = \mathcal{Q}_C^{y,b}(\mathbf{X}_{mn} + \mathbf{W}_{mn}) + \tilde{\mathbf{T}}_{mn}$$



QRPCA Problem Formulation

□ $\mathbf{Z} = \mathcal{Q}_C^{y,b}(\mathbf{X} + \mathbf{W}) + \tilde{\mathbf{T}}$ can be equivalent to $\mathbf{Z} = \mathcal{Q}_C^{y,b}(\mathbf{X} + \mathbf{T} + \mathbf{W})$ (omitting the subscript mn).

- \mathbf{X} : Low rank. Its rank depends on the number of targets with different distances or different velocities.
- $\tilde{\mathbf{T}}$: Sparse. It is generally sparse since the bit error rate (BER) is generally quite low.
- \mathbf{T} : Sparse. It is an equivalent sparse DTE before quantization.

□ Recover the low-rank matrix \mathbf{X} and the sparse matrix \mathbf{T} by solving QRPCA problem.

- Function $D(\cdot, \cdot)$ is similarity metric which measures the similarity between the quantized data \mathbf{Z} and the unquantized data $\mathbf{Y} = \mathbf{X} + \mathbf{T}$.

$$-\frac{\Delta}{2} \leq \Re\{\mathbf{Y} - \mathbf{Z}\} \leq \frac{\Delta}{2} \quad \Rightarrow \quad D(\mathbf{Z}, \mathbf{X} + \mathbf{T}) = \left\| \rho \left(\left[\Re\{\mathbf{X} + \mathbf{T} - \mathbf{Z}\} + \frac{\Delta}{2}; \Im\{\mathbf{X} + \mathbf{T} - \mathbf{Z}\} + \frac{\Delta}{2} \right] \right) \right\|_F^2 + \left\| \rho \left(\left[\Re\{\mathbf{Z} - \mathbf{X} - \mathbf{T}\} + \frac{\Delta}{2}; \Im\{\mathbf{Z} - \mathbf{X} - \mathbf{T}\} + \frac{\Delta}{2} \right] \right) \right\|_F^2$$

where $\rho(\cdot)$ is an element-wise function with $\rho(x) = \max\{-x, 0\}$.

$$\min_{\mathbf{X}, \mathbf{T}} \frac{1}{2} D(\mathbf{Z}, \mathbf{X} + \mathbf{T}) + \mu \|\mathbf{X}\|_* + \lambda \|\mathbf{T}\|_1$$

Low rank Sparse

Method

APG-QRPCA Algorithm

□ Define $h(\mathbf{X}, \mathbf{T}) = \mu \|\mathbf{X}\|_* + \lambda \|\mathbf{T}\|_1$ and $g(\mathbf{X}, \mathbf{T}) = \frac{1}{2} D(\mathbf{Z}, \mathbf{X} + \mathbf{T})$,

□ $h(\mathbf{X}, \mathbf{T})$ is convex, and $g(\mathbf{X}, \mathbf{T})$ is differentiable.

□ The QRPCA problem can be rewritten as $\min_{\mathbf{X}, \mathbf{T}} h(\mathbf{X}, \mathbf{T}) + g(\mathbf{X}, \mathbf{T})$

□ Iteratively calculate

- Calculate momentum $\bar{\mathbf{X}}_l = \mathbf{X}_l + \frac{\zeta_l - 1}{\zeta_l} (\mathbf{X}_l - \mathbf{X}_{l-1}), \bar{\mathbf{T}}_l = \mathbf{T}_l + \frac{\zeta_l - 1}{\zeta_l} (\mathbf{T}_l - \mathbf{T}_{l-1})$

- Gradient descent $\mathbf{X}_p = \bar{\mathbf{X}}_l - \delta \nabla_{\mathbf{X}} g(\bar{\mathbf{X}}_l, \bar{\mathbf{T}}_l), \mathbf{T}_p = \bar{\mathbf{T}}_l - \delta \nabla_{\mathbf{T}} g(\bar{\mathbf{X}}_l, \bar{\mathbf{T}}_l)$

- Proximal map $\mathbf{X}_{l+1} = \arg \min_{\mathbf{X}} \left\{ \mu \|\mathbf{X}\|_* + \frac{1}{2\delta} \|\mathbf{X} - \mathbf{X}_p\|_F^2 \right\} = \mathbf{U}_p \mathcal{S}_{\mu\delta}(\mathbf{\Sigma}_p) \mathbf{V}_p^T$,

$$\mathbf{T}_{l+1} = \arg \min_{\mathbf{T}} \left\{ \lambda \|\mathbf{T}\|_1 + \frac{1}{2\delta} \|\mathbf{T} - \mathbf{T}_p\|_F^2 \right\} = \mathcal{S}_{\lambda\delta}(\mathbf{T}_p)$$

LS-based Target Parameter Estimation

□ $M_t \times M_r$ TIM matrixes can be recovered at the fusion center.

□ A sequential LS method introduced to sequentially estimate the position and velocity parameters.

- $\theta_p = \{\mathbf{p}^{(k)}\}_{k=1}^K$ and $\theta_v = \{\mathbf{v}^{(k)}\}_{k=1}^K$ are implicitly determined by \mathbf{A}_{mn} and \mathbf{B}_{mn}

Original problem:

$$\hat{\theta} = \arg \min_{\theta_p, \theta_v} \sum_{m=1}^{M_t} \sum_{n=1}^{M_r} \|\bar{\mathbf{x}}_{mn} - \mathbf{A}_{mn}(\theta_p) \mathbf{A}_{mn} \mathbf{B}_{mn}(\theta_v)\|_2^2$$

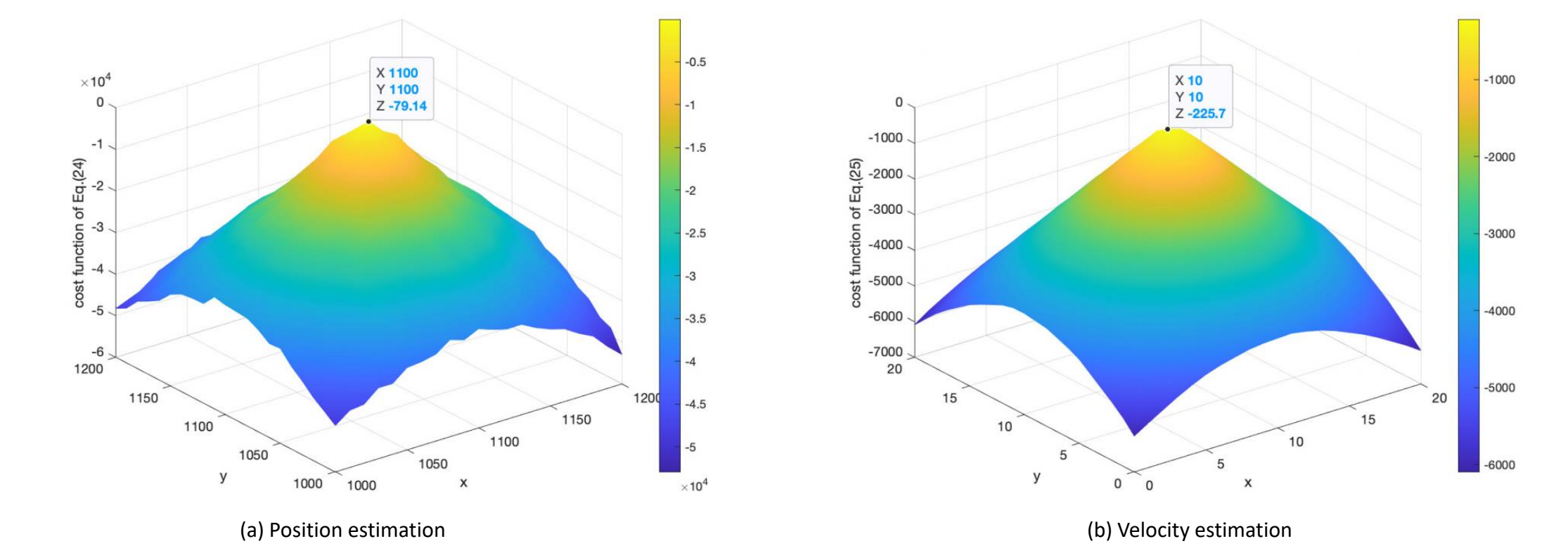
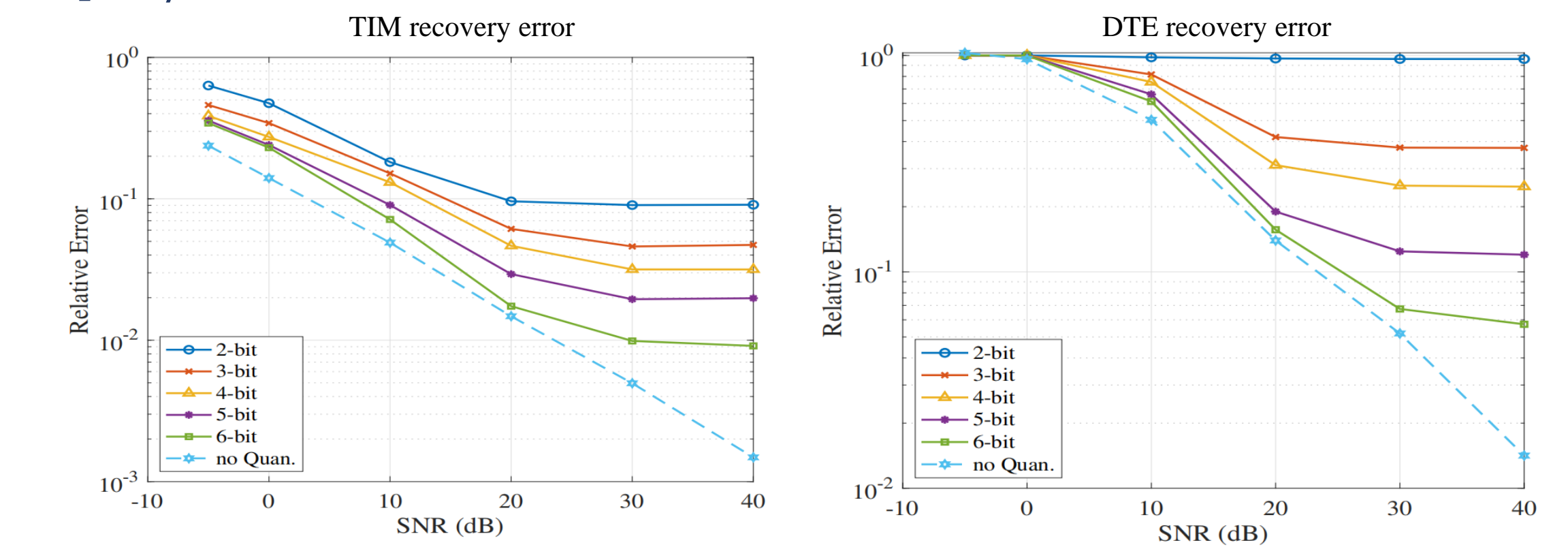
Two subproblems:

$$\hat{\theta}_p = \arg \min_{\theta_p} \sum_{m=1}^{M_t} \sum_{n=1}^{M_r} \|\mathbf{P}_{p,mn}^\perp(\theta_p) \bar{\mathbf{x}}_{mn}\|_2^2$$

$$\hat{\theta}_v = \arg \min_{\theta_v} \sum_{m=1}^{M_t} \sum_{n=1}^{M_r} \sum_{k=1}^K \left(\hat{f}_{mn}^{(k)} - f_{mn}^{(k)}(\mathbf{v}^{(k)}) \right)^2$$

Numerical Example

$M_t = 3$ transmit antennas, $M_r = 10$ receive antennas, uniformly distributed on the concentric circles with radius 5km and 3km, respectively. The reference carrier frequency parameters $f_0 = 5\text{GHz}$ and the frequency increment $\Delta f = 50\text{MHz}$. One CPI consists of $Q=128$ pulses with $T_{PRI} = 0.5\text{ms}$ and $T_p = 6.4\mu\text{s}$. The transmitters emit Hadamard sequences with length of $N = 64$. 1% symbol error rate is assumed to lead sparse data transmission error matrix. One target is located at $\mathbf{p}^{(1)} = [1100, 1100]^T \text{m}$ with $\mathbf{v}^{(1)} = [10, 10]^T \text{m/s}$.



Conclusions

- Propose a low-bit quantized distributed MIMO radar system;
- Formulate a QRPCA problem to recover the infinite-precision target information matrix and the data transmission errors simultaneously;
- Demonstrate the feasibility of implementing a low-bit quantized distributed MIMO radar system;
- The analysis of the performance bound of the proposed LiQuiD-MIMO radar will be our future work.