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# Robust FIR Filters for Wireless Low-Frequency Sound Zones

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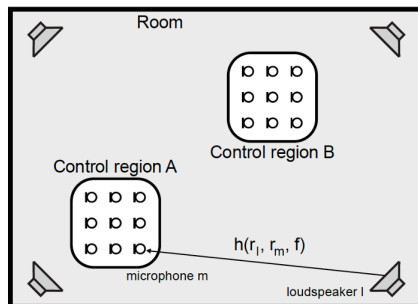
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# Outline

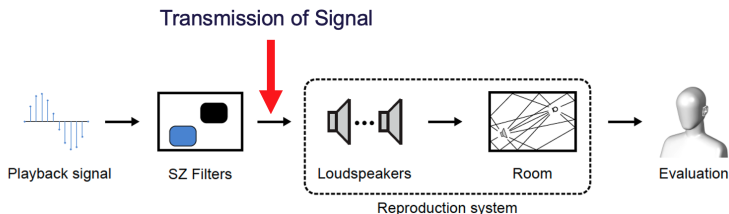
- Introduction of Sound Zone Systems
- Model with Packet Loss
- Filter Derivation
- Simulation
  - ▶ Simulation Setting
  - ▶ Contrast
  - ▶ Sound Quality
- Discussion

# Sound Zone



- Bright zone: where the sound is desired
- Dark zones: where the sound is suppressed
- Different control strategies are usually considered in different frequency ranges
- we focus on the creation of sound zones at **low frequencies**

# Sound Zone Systems



- Cable: high speed, robustness
- Wireless:
  - ▶ portability, increased flexibility, lower installation costs
  - ▶ bit errors and loss of packets

# Effect of Packet Loss on Sound Zones

- Lower contrast
- Lower audio quality
- More leakage to the dark zone

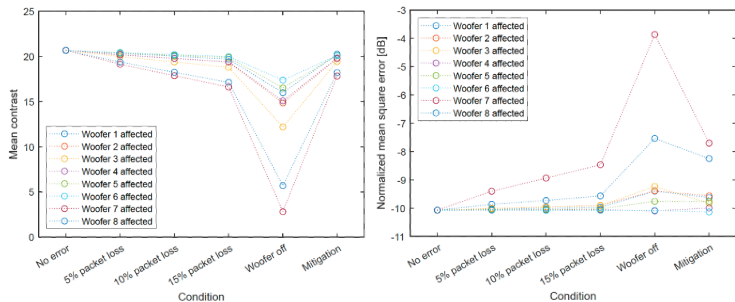


Figure: Contrast and normalized mean square error for 10s gaussian white noise depending on error and affected woofers.<sup>1</sup>

<sup>1</sup>C. S. Pedersen, M. B. Møller, and J. Østergaard, "Effect of wireless transmission errors on sound zone performance at low frequencies," EUROREGIO BNAM2022 Joint Acoustic Conference, pp. 115–124, May 2022.

## Preliminaries

- The sound pressure at time  $n$  recorded by microphone  $m$  due to loudspeaker  $l$ :

$$p_{m,l}(n) = \sum_{j=0}^{J-1} h_{m,l}(j) \sum_{i=0}^{l-1} w_l(i) x_s(n - i - j), \quad (1)$$

- ▶  $x_s$  is the input audio signal,
  - ▶  $\mathbf{h}_{m,l} = (h_{m,l}(0), \dots, h_{m,l}(J-1))^T$  is the room impulse response (RIR),
  - ▶  $\mathbf{w}_l = (w_l(0), \dots, w_l(l-1))^T$  is the FIR filters.
- Assume the source signal  $x_s$  to be spectrally flat:

$$\mathbf{p}_{m,l} = \mathbf{H}_{m,l} \mathbf{w}_l, \quad (2)$$

- ▶  $\mathbf{H}_{m,l} \in \mathbb{R}^{(l+J-1) \times l}$  is a Toeplitz matrix.

## Model with Packet Loss

- Assume the packet size is 1 and packet loss in each channel  $l$  is independent:

$$P(\delta_l(t) = 0) = p_l, P(\delta_l(t) = 1) = 1 - p_l,$$

where  $\delta_l(t)$  is the observation of whether a packet is lost.

- Sound Pressure under packet loss can be written as:

$$\mathbf{p}_{m,l} = \mathbf{H}_{m,l} \text{diag}(\boldsymbol{\delta}_l) \mathbf{w}_l,$$

where  $\boldsymbol{\delta}_l = (\delta_l(0), \dots, \delta_l(l-1))^T$ .

- For  $M$  microphone positions:

$$\mathbf{p} = \mathbf{H} \Delta \mathbf{w}, \quad (3)$$

where

$$\mathbf{H} = (\mathbf{H}_1^T, \dots, \mathbf{H}_M^T)^T \text{ with } \mathbf{H}_m = (\mathbf{H}_{m,1}, \dots, \mathbf{H}_{m,L}),$$
$$\boldsymbol{\delta} = (\boldsymbol{\delta}_1^T, \dots, \boldsymbol{\delta}_L^T)^T, \mathbf{w} = (\mathbf{w}_1^T, \dots, \mathbf{w}_L^T)^T, \Delta = \text{diag}(\boldsymbol{\delta}).$$

# Cost Function

- Sound Pressure in the bright zone and dark zone:

$$\mathbf{p}_B = \mathbf{H}_B \Delta \mathbf{w}, \quad \mathbf{p}_D = \mathbf{H}_D \Delta \mathbf{w}. \quad (4)$$

- We will use the following cost function:

$$J_{pl}(\mathbf{w}) = (1 - \beta) \mathbb{E}\{\|\mathbf{p}_B - \mathbf{p}_T\|_2^2\} + \beta \mathbb{E}\{\|\mathbf{p}_D\|_2^2\} + \lambda_w \mathbf{w}^T \mathbf{R}_w \mathbf{w}, \quad (5)$$

- ▶  $\mathbf{p}_T$  is the target sound pressure in the bright zone
- ▶  $\mathbf{R}_w$  is a weighting matrix for controlling the shape of the FIR filters
- ▶ The expectation  $\mathbb{E}(\cdot)$  is with respect to the packet loss  $\Delta$ .



# Filter Derivation

- The FIR filters  $\mathbf{w}$  can be estimated by minimizing (5):

$$\begin{aligned}\mathbf{w}_{opt} &= [(1 - \beta)\mathbb{E}(\Delta\mathbf{H}_B^T\mathbf{H}_B\Delta) + \beta\mathbb{E}(\Delta\mathbf{H}_D^T\mathbf{H}_D\Delta) \\ &\quad + \lambda_w\mathbf{R}_w]^{-1}(1 - \beta)\mathbb{E}(\Delta)\mathbf{H}_B^T\mathbf{p}_T. \\ &= [(1 - \beta)\mathbf{H}_B^T\mathbf{H}_B \odot \Omega + \beta\mathbf{H}_D^T\mathbf{H}_D \odot \Omega \\ &\quad + \lambda_w\mathbf{R}_w]^{-1}(1 - \beta)(\Psi \otimes \mathbf{I}_l)\mathbf{H}_B^T\mathbf{p}_T.\end{aligned}\tag{6}$$

where

- $\mathbb{E}(\Delta) = \Psi \otimes \mathbf{I}_l$ , where  $\mathbf{I}_l$  is an  $l$ -by- $l$  identity matrix and  $\Psi = \text{diag}(1 - \rho_1, \dots, 1 - \rho_L)$ ,  $\otimes$  denotes Kronecker product, and

$$\Omega = \begin{pmatrix} (1 - \rho_1)^2\Omega_1 & \cdots & (1 - \rho_1)(1 - \rho_L)\mathbf{1}_l \\ (1 - \rho_1)(1 - \rho_2)\mathbf{1}_l & \cdots & (1 - \rho_2)(1 - \rho_L)\mathbf{1}_l \\ \vdots & \ddots & \vdots \\ (1 - \rho_1)(1 - \rho_L)\mathbf{1}_l & \cdots & (1 - \rho_L)^2\Omega_L \end{pmatrix}.$$

## Simulation Setting

- simulate a 5.5 m by 8.65 m by 2.7 m room using Green's function for point sources in rectangular rooms, with 0.6s  $T_{60}$  reverberation time and  $L = 8$  loudspeakers.
- The number of microphone positions sampled in the bright and dark zones are  $M_B = M_D = 75$ .

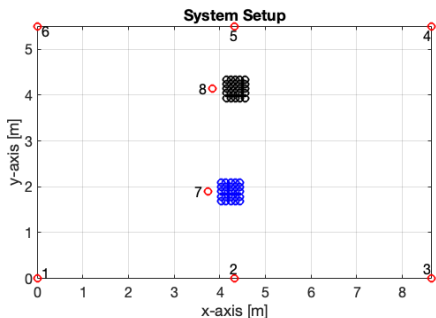


Figure: The blue and black circles are the microphones in the bright zone and dark zone respectively. The red circles are the loudspeakers.

# Simulation Setting

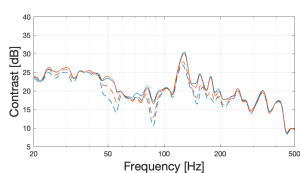
- focus cases where only one channel is subject to packet loss,
- $\omega_{i,p}, i = 1, \dots, 8$ : proposed filters derived by assuming Channel  $i$  has packet loss rate  $p$ , where  $p = 5\%, 10\%, 15\%$ ,
- $\omega_{old}$ : original filters derived without packet loss,
- The RIRs and the filters are of length  $J = 600$  and  $I = 300$ ,
- $\beta = 0.97$  and  $\lambda_w = 10^{-7}$ ,
- Input signal: 10s signal sampled at 1200 Hz, encoded into consecutive and non-overlapping blocks, each consists of 24 samples (20 ms).

# Smoothed Contrast

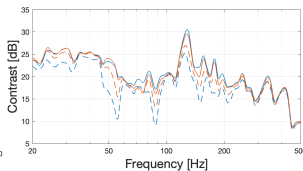
- Input signal: Gaussian white noise
- Smoothed Contrast:

$$SC(\omega) = 10 \times \log_{10} \left( \frac{S(\sum_{m=1}^M (\sum_{l=1}^L \tilde{\hat{p}}_{B,m,l}(\omega))^2 / M_B)}{S(\sum_{m=1}^M (\sum_{l=1}^L \tilde{\hat{p}}_{D,m,l}(\omega))^2 / M_D)} \right),$$

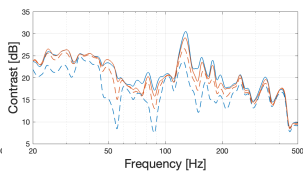
where  $\tilde{\hat{p}}_{B,m,l}$ ,  $\tilde{\hat{p}}_{D,m,l}$  are the Fourier transform of  $\hat{p}_{B,m,l}$ ,  $\hat{p}_{D,m,l}$  respectively, and  $S(\cdot)$  is the smoothing function.



(a)  $p = 5\%$



(b)  $p = 10\%$



(c)  $p = 15\%$

# Mean Contrast

- Input signal: Gaussian white noise
- The Mean Contrast is defined as

$$MC = \frac{1}{N} \sum_{n=1}^N C_T(n),$$

where

$$C_T(n) = 10 \times \log_{10} \left( \frac{\sum_{m=1}^M (\sum_{l=1}^L \hat{p}_{B,m,l}(n))^2}{\sum_{m=1}^M (\sum_{l=1}^L \hat{p}_{D,m,l}(n))^2} \right).$$

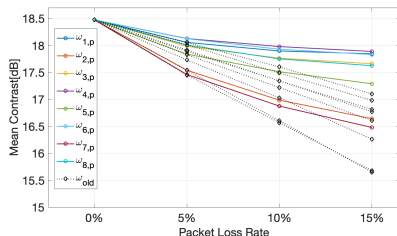


Figure: Mean Contrast for  $\omega_{old}$  and  $\omega_{i,p}$  under different packet loss rates.

## Sound Quality in Bright Zone

- Input signal: 10-second segment of *Dazed and Confused* from the Album “Led Zeppelin” .
- We use the Perceptual Evaluation of Audio Quality (PEAQ) model to predict sound quality in the bright zone.

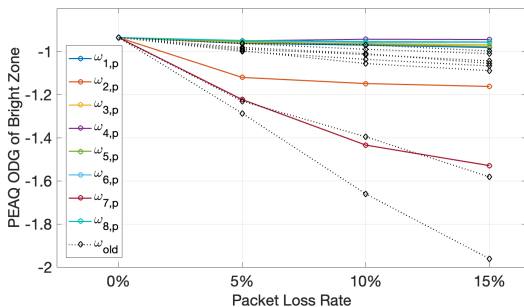


Figure: PEAQ ODG of bright zone for  $\omega_{old}$  and  $\omega_{i,p}$  under different packet loss rates.

## Sound Quality Reduction due to Leakage

- We evaluate the sound quality reduction of listening to the audio when exposed to leakage as well as the intended audio.
- Assume that we have two zones (Zone A and Zone B), we reproduce audio signal A in Zone A and seek to reduce the leakage towards Zone B. Zone B has a loudspeaker reproducing audio signal B for Zone B.
- We then evaluate how a person in Zone B experiences the quality of listening to the combination of the reproduced audio signal B and the leaked audio from Zone A under different packet loss patterns.

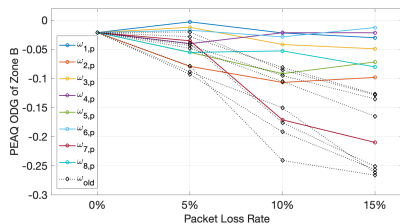


Figure: PEAQ ODG of Zone B for  $\omega_{old}$  and  $\omega_{i,p}$  under different packet loss rates.

# Discussion

- Incorporating packet loss information can improve the performance of the wireless low-frequency sound zone system. With packet loss rate increases, the improvement is more significant,
- Our proposed filters not only improves the overall contrast when packet losses occur, but also has comparable performance when evaluated with no packet loss,
- Future investigation: incorporating bursty packet loss, cases when all channels have packet loss.



Thank you!