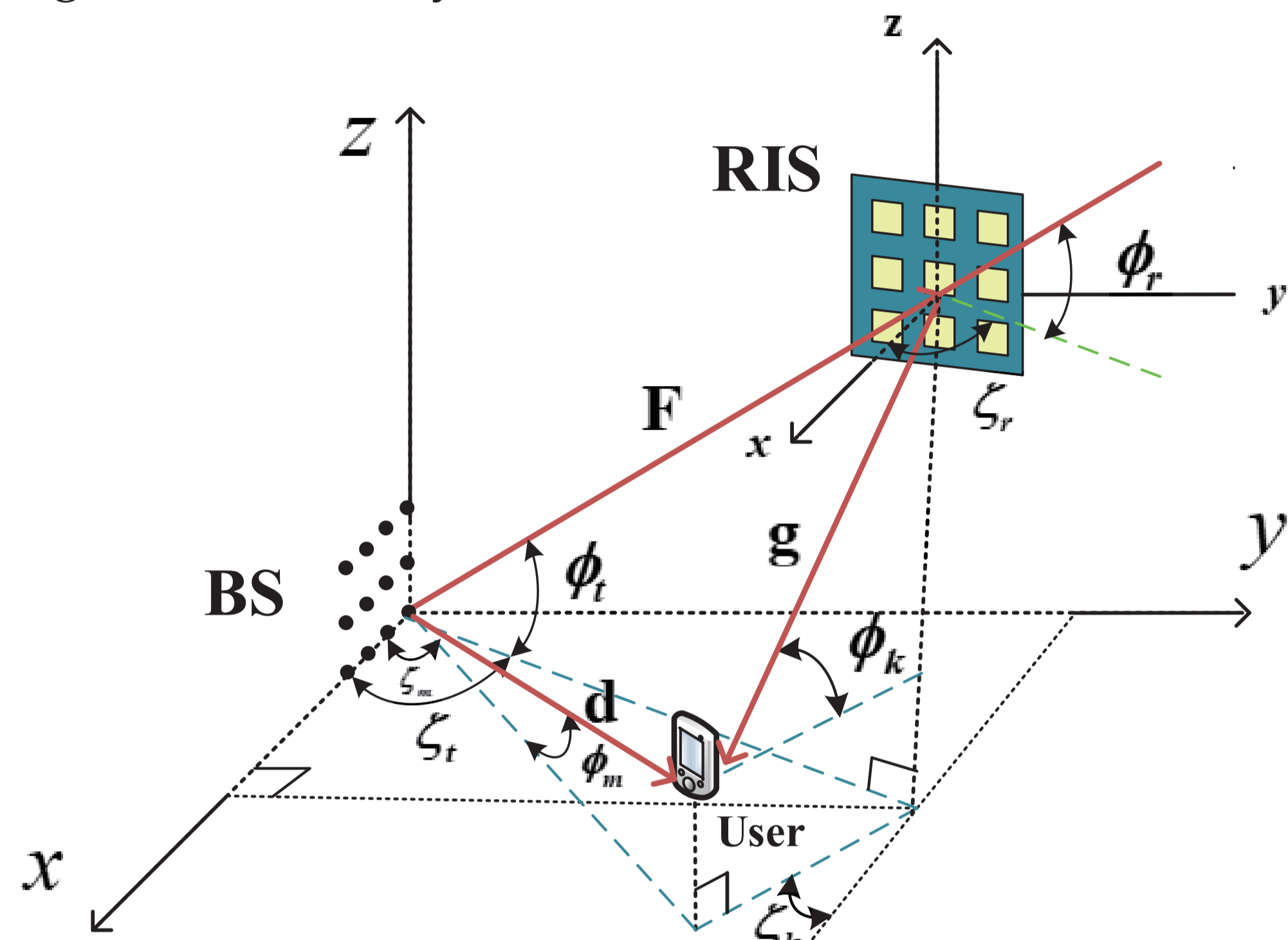


Location Optimization for RIS aided mmWave downlink network

Qian Xiang, Cong Sun, Danpu Liu, Beijing University of Posts and Telecommunications
suncong86@bupt.edu.cn

INTRODUCTION

- **Reconfigurable Intelligent Surface (RIS)**: passive phase shift; high energy efficiency; popular candidate technique for 6G
- **Millimeter wave (mmWave)**: short wavelength; wide frequency band; great directivity



SYSTEM MODEL

Saleh-Valenzuela (SV) channels for UPAs at BS and RIS:

$$\mathbf{F} = L(l_B) \mathbf{a}_R(\zeta_r, \phi_r) \mathbf{a}_T(\zeta_t, \phi_t)^H, \mathbf{g} = L(l_U) \mathbf{a}_R(\zeta_k, \phi_k), \mathbf{d} = L(l_d) \mathbf{a}_T(\zeta_m, \phi_m),$$

where the path loss component, the normalized receive and transmit array response vector (ARV) are as follows respectively:

$$L(l_B) = C \left(\frac{l_B}{l_0} \right)^{-\alpha},$$

$$\mathbf{a}_R(\zeta_r, \phi_r) = \left[1, \dots, e^{j \frac{2\pi d_R}{\lambda} (N_x - 1) \sin \zeta_r \sin \phi_r} \right]^T \otimes \left[1, \dots, e^{j \frac{2\pi d_R}{\lambda} (N_z - 1) \cos \phi_r} \right]^T,$$

$$\mathbf{a}_T(\zeta_t, \phi_t) = \left[1, \dots, e^{j \frac{2\pi d_B}{\lambda} (M_x - 1) \sin \zeta_t \sin \phi_t} \right]^T \otimes \left[1, \dots, e^{j \frac{2\pi d_B}{\lambda} (M_z - 1) \cos \phi_t} \right]^T.$$

Here $\mathbf{r} = (x_r, y_r, z_r)^T$ represents the location of RIS, $\sin \zeta_r = \frac{y_r}{\|(\mathbf{b}-\mathbf{r})_{1,2}\|}$, $\sin \phi_r = \frac{z_r - z_b}{\|\mathbf{r}-\mathbf{b}\|}$, $\zeta_t = \zeta_r$ and $\phi_t = \phi_r$.

The signal received at the user is:

$$y = (\mathbf{d}^H + \mathbf{g}^H \Theta \mathbf{F}) \mathbf{w} s + n,$$

where $n \sim \mathcal{CN}(0, \sigma^2)$, s is the transmit signal for the user satisfying $\mathbb{E}(|s|^2) = 1$, $\mathbf{w} \in \mathbb{C}^M$ is the precoding vector and $\Theta = \text{Diag}(\theta_1, \dots, \theta_N)$ is the RIS matrix.

SNR MAXIMIZATION

$$\max_{\substack{\mathbf{w} \in \mathbb{C}^M, \mathbf{r} \in \mathbb{R}^3, \\ \Theta \in \mathbb{C}^{N \times N}}} \text{SNR} = \frac{|(\mathbf{d}^H + \mathbf{g}(\mathbf{r})^H \Theta \mathbf{F}(\mathbf{r})) \mathbf{w}|^2}{\sigma^2} \quad (1)$$

$$\text{s. t. } \|\mathbf{w}\|^2 \leq P_T,$$

$$\Theta = \text{Diag}(\theta_1, \dots, \theta_N), \theta_l \in \Omega, l = 1, \dots, N.$$

RIS constraint:

$$\Omega = \left\{ e^{i\varphi}, \varphi \in \left\{ 0, \frac{2\pi}{T}, \dots, \frac{2\pi(T-1)}{T} \right\} \right\} \text{ or } \{ \theta_l \in \mathbb{C} \mid |\theta_l| \leq 1 \}.$$

Plug the optimal MRT precoding vector $\mathbf{w}^* = \frac{\sqrt{P_T}}{\|\mathbf{h}\|} \mathbf{h}$ into (1), and the equivalent optimization problem becomes:

$$\max_{\mathbf{v} \in \mathbb{C}^N, \mathbf{r} \in \mathbb{R}^3} \|\mathbf{h}(\mathbf{v}, \mathbf{r})\|^2 \quad (2)$$

$$\text{s. t. } \theta_l \in \Omega, l = 1, \dots, N,$$

where $\mathbf{h}(\mathbf{v}, \mathbf{r}) = \mathbf{d} + \mathbf{F}^H \Theta^H \mathbf{g}$

$$= \mathbf{d} + \frac{C^2}{\|\mathbf{u}-\mathbf{r}\|^2 \|\mathbf{b}-\mathbf{r}\|^2} \left[\mathbf{a}_R(\zeta_r(\mathbf{r}), \phi_r(\mathbf{r}))^H \mathbf{B} \mathbf{v} \right] \mathbf{a}_T(\zeta_t(\mathbf{r}), \phi_t(\mathbf{r})).$$

Here $\mathbf{v} = (\theta_1, \dots, \theta_N)^H$, and $\mathbf{B} = \text{Diag}[\mathbf{a}_R(\zeta_k(\mathbf{r}), \phi_k(\mathbf{r}))]$.

ADMM METHOD

Augmented Lagrangian penalty function technique:

$$\max_{\mathbf{v}, \mathbf{r}, \mathbf{p}} \|\mathbf{h}(\mathbf{v}, \mathbf{r})\|^2 \quad \Rightarrow \quad \min_{\mathbf{v}, \mathbf{r}, \mathbf{p}} -\|\mathbf{h}(\mathbf{v}, \mathbf{r})\|^2 + \text{Re} \{ \boldsymbol{\mu}^H (\mathbf{p} - \mathbf{v}) \} + \frac{\rho}{2} \|\mathbf{p} - \mathbf{v}\|^2$$

$$\text{s. t. } \mathbf{p} = \mathbf{v}, \quad \text{s. t. } p_l \in \Omega, l = 1, \dots, N.$$

- **v subproblem**: $\min_{\mathbf{v}} \frac{1}{2} \mathbf{v}^H \mathbf{Q} \mathbf{v} + \text{Re}(\mathbf{v}^H \mathbf{q})$
closed-form solution: $\mathbf{v}_{k+1} = -\mathbf{Q}^{-1} \mathbf{q}$, or
solution with sufficient function reduction: $\mathbf{v}_{k+1} = \mathbf{v}_k + \alpha_k \mathbf{d}_k$.

- **p subproblem**:

$$\min_{\mathbf{p} \in \mathbb{C}^N} \frac{\rho}{2} \mathbf{p}^H \mathbf{p} + \text{Re} \{ (\boldsymbol{\mu} - \rho \mathbf{v})^H \mathbf{p} \} \quad \text{closed-form solution} \quad p_l^* = \begin{cases} \eta_l, & \text{if } |\eta_l| \leq 1, \\ \frac{\eta_l}{|\eta_l|}, & \text{otherwise.} \end{cases}$$

$$\text{s. t. } p_l \in \Omega, l = 1, \dots, N.$$

$$\text{or } p_l^* = e^{i\varphi_l^*}, \varphi_l^* = \arg \min_{\varphi_l \in \{0, \frac{2\pi}{T}, \dots, \frac{2\pi(T-1)}{T}\}} |\varphi_l - \text{Arg}(\rho \eta_l)|, \eta_l = v_l - \frac{\mu_l}{\rho}.$$

- **r subproblem**: $\min_{\mathbf{r}} -\|\mathbf{h}(\mathbf{v}, \mathbf{r})\|^2$

Levenberg-Marquardt (LM) method:

$$\mathbf{r}^{k+1} = \mathbf{r}^k + \left(\gamma_k \mathbf{I} - \text{Re} \{ \mathbf{J}_k^H \mathbf{J}_k \} \right)^{-1} \text{Re} \{ \mathbf{J}_k^H \mathbf{h}_k \},$$

from subproblem:

$$\min_{\mathbf{r}} -\|\mathbf{h}(\mathbf{r}) + \mathbf{J}(\mathbf{r}^k)(\mathbf{r} - \mathbf{r}^k)\|^2 \quad \text{s. t. } \|\mathbf{r} - \mathbf{r}^k\| \leq \Delta_k.$$

SIMULATION RESULTS

Parameter settings: $\mathbf{b} = (0, 0, 10)^T$, $\mathbf{u} = (50, 50, 5)^T$, $M_x = 4$, $M_z = 4$, $N_x = 10$, $N_z = \frac{N}{N_x}$, $\sigma^2 = -110\text{dBm}$, $P_T = 2.5\text{dBm}$, $T = 4$.

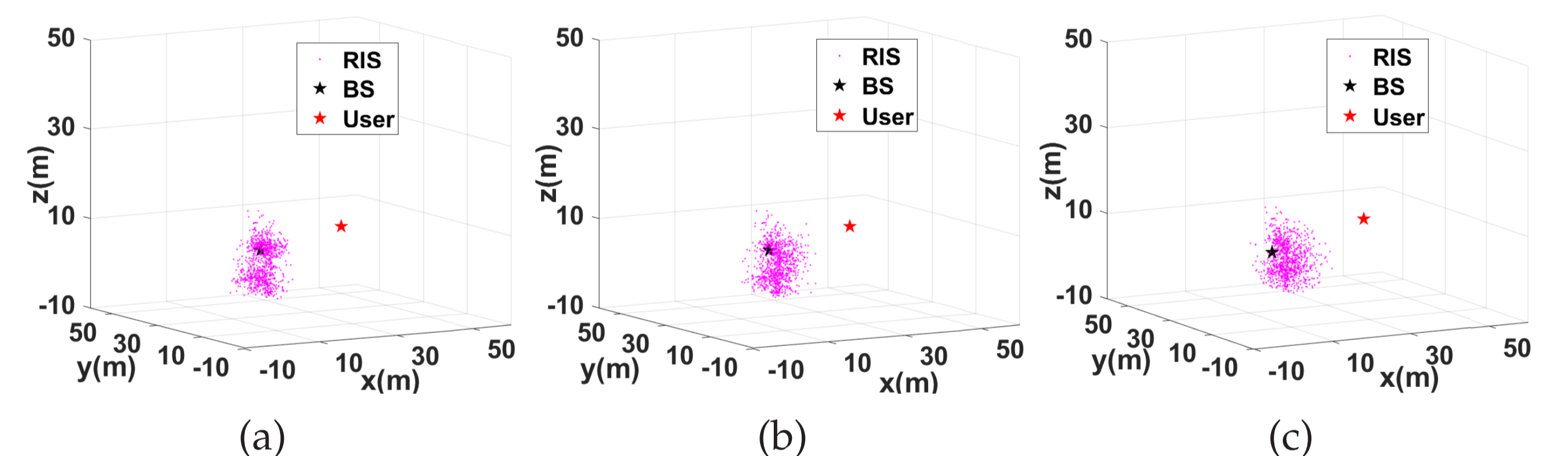


Figure 1: RIS locations with different initial setting

The initial RIS location \mathbf{r}^0 is randomly selected from the 10m radius circle centered at: (a) BS; (b) midpoint between BS and the user; (c) the user.

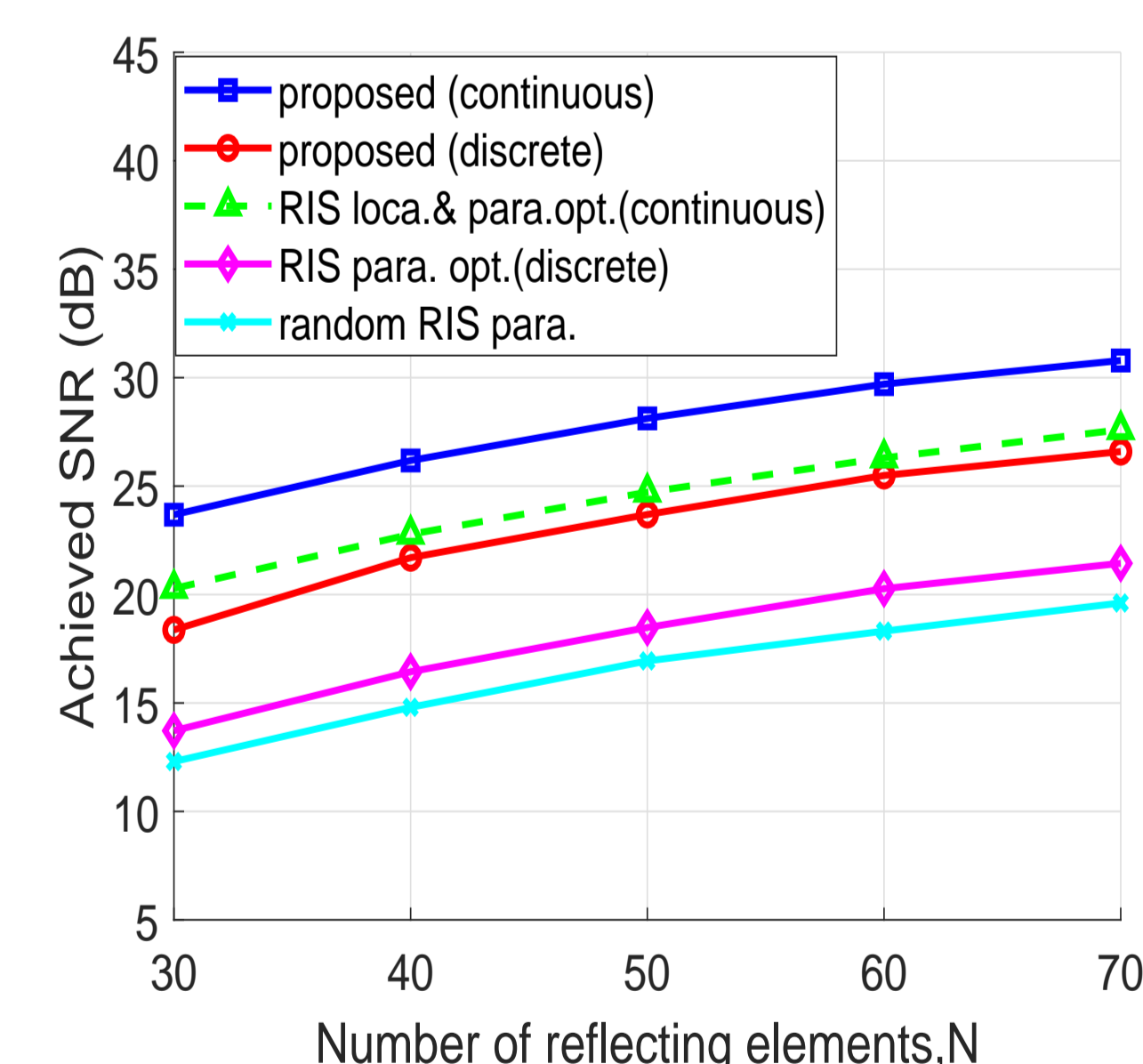


Figure 2: Achieved SNR with respect to N

Joint optimization of RIS location and parameters greatly outperforms other benchmarks.

REFERENCES

- [1] H. Lu, Y. Zeng, S. Jin and R. Zhang, "Aerial Intelligent Reflecting Surface: Joint Placement and Passive Beamforming Design With 3D Beam Flattening," IEEE Transactions on Wireless Communications, vol. 20, no. 7, pp. 4128-4143, July 2021.
- [2] Q. Wu and R. Zhang, "Beamforming Optimization for Intelligent Reflecting Surface with Discrete Phase Shifts," IEEE International Conference on Acoustics, Speech and Signal Processing (ICASSP), pp. 7830-7833, Apr. 2019.