

Transmission Schemes for Enhanced DoF in Cache-Aided MIMO Communication Systems

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Outline

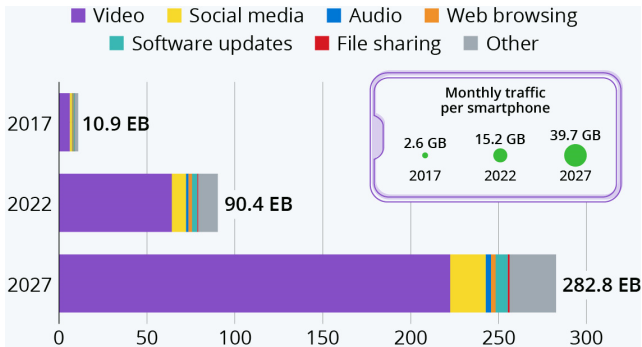
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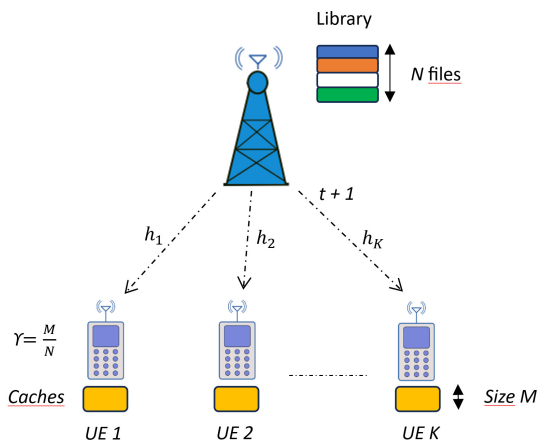
Coded Caching



- Mobile data traffic is growing continuously
 - multimedia content
 - mobile immersive viewing and extended reality.
- Wireless network infrastructure falls under considerable strain
 - demanding requirements: high throughput to ultra-low latency.

Coded Caching: SISO

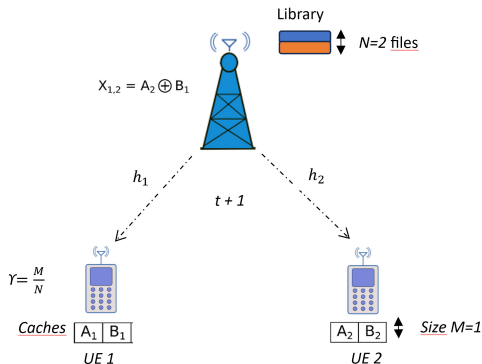
- Coded caching (CC) allows to use the UE memory as a new communication resource



- Offers a new performance (DoF) gain denoted by $t = K\gamma$.
- Enabled by:
 - 1) Cache-placement,
 - 2) Delivery phase
- The gain of CC is from multicasting codewords to groups of users of size $t + 1^a$,

^aA. Maddah-Ali, "Fundamental limits of caching," IEEE Trans. Inf. Theory, vol. 60, no. 5, pp. 2856-2867, 2014

SISO Coded Caching: Example 1



- BS carefully construct codewords within each Tx vector ($t = 1$)

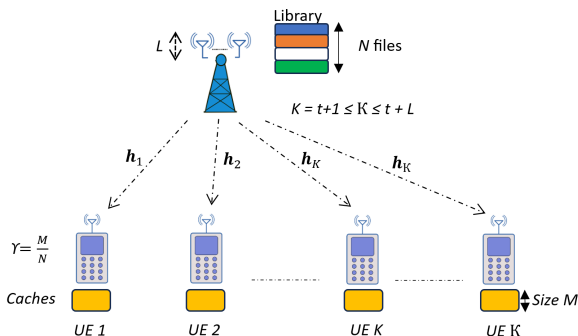
$$X_{1,2} = A_2 \oplus B_1,$$

- Interference-free decoding of desired signal is possible per user

$$y_1 = h_1(A_2 \oplus B_1) + z_1, \text{ CC-aided Intf cancellation : } \Rightarrow \bar{y}_1 = h_1 A_2 + z_1,$$

$$y_2 = h_2(A_2 \oplus B_1) + z_2, \text{ CC-aided Intf cancellation : } \Rightarrow \bar{y}_2 = h_2 B_1 + z_2,$$

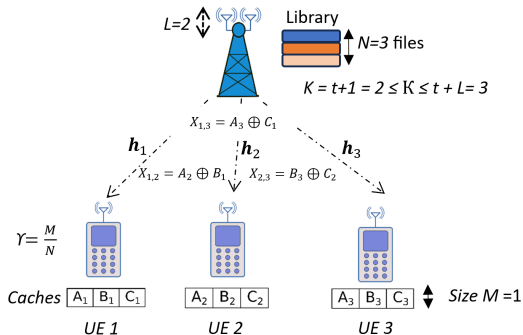
MISO Coded Caching



- The achievable DoF is increased to $t + L$ in a MISO setup ¹,
 - The interference by user k can be removed by cache contents at up to t users.
 - The spatial multiplexing gain of L at Tx-side nulls intf at other UEs
 - From receiver perspective of each UE, y_k is an equivalent G-MAC.

¹. S. P. Shariatpanahi, G. Caire, and B. Hossein Khalaj, "Physical-Layer Schemes for Wireless Coded Caching," IEEE Trans. Inf. Theory, vol. 65, no. 5, pp. 2792-2807, 2019

MISO Coded Caching: Example 1

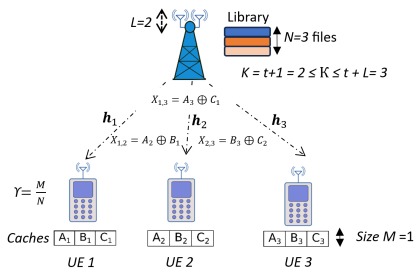


User	Available in Cache	Supp. by Beamformer	Useful data
1	A_1, B_1, C_1	$C_2 \oplus B_3$	A_2, A_3
2	A_2, B_2, C_2	$A_3 \oplus C_1$	B_1, B_3
3	A_3, B_3, C_3	$A_2 \oplus B_1$	C_1, C_2

Decoding Process for \mathbf{X} at different network UEs

²A. T. Iolli, "Multi-antenna interference management for coded caching," IEEE Trans. Wireless Commun., vol. 19, no. 3, pp. 2091â2106, 2020

MISO Coded Caching: Example 1



$$\mathbf{x} = (A_2 \oplus B_1)\mathbf{w}_{1,2} + (A_3 \oplus C_1)\mathbf{w}_{1,3} + (C_2 \oplus B_3)\mathbf{w}_{2,3}$$

$$y_1 = \underline{(\mathbf{h}_1^H \mathbf{w}_{1,2})\tilde{X}_{1,2}} + \underline{(\mathbf{h}_1^H \mathbf{w}_{1,3})\tilde{X}_{1,3}} + \underline{(\mathbf{h}_1^H \mathbf{w}_{2,3})\tilde{X}_{2,3}} + z_1$$

$$y_2 = \underline{(\mathbf{h}_2^H \mathbf{w}_{1,2})\tilde{X}_{1,2}} + \underline{(\mathbf{h}_2^H \mathbf{w}_{1,3})\tilde{X}_{1,3}} + \underline{(\mathbf{h}_2^H \mathbf{w}_{2,3})\tilde{X}_{2,3}} + z_2$$

$$y_3 = \underline{(\mathbf{h}_3^H \mathbf{w}_{1,2})\tilde{X}_{1,2}} + \underline{(\mathbf{h}_3^H \mathbf{w}_{1,3})\tilde{X}_{1,3}} + \underline{(\mathbf{h}_3^H \mathbf{w}_{2,3})\tilde{X}_{2,3}} + z_3$$

User	Available in Cache	Supp. by Beamformer	Useful data
1	A_1, B_1, C_1	$C_2 \oplus B_3$	A_2, A_3
2	A_2, B_2, C_2	$A_3 \oplus C_1$	B_1, B_3
3	A_3, B_3, C_3	$A_2 \oplus B_1$	C_1, C_2

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MIMO CC: System Model

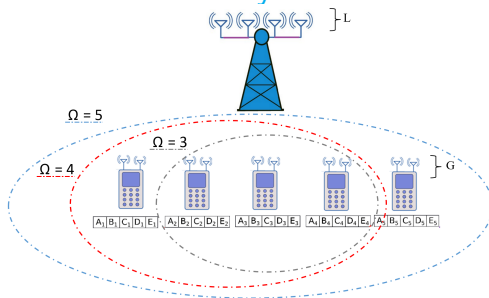


Figure: MIMO CC system model for arbitrary L and G and served users Ω

- Propose a flexible framework with enhanced DoF
 - The number of served UEs selected to optimize the achievable DoF for any given L , G , and t .
- Placement: split each file $W \in \mathcal{F}$ into $\binom{K}{t}$ subfiles $W_{\mathcal{P}}$, where $\mathcal{P} \subseteq [K]$, $|\mathcal{P}| = t$.
- Delivery: The server constructs and transmits the demand set via $\mathbf{x}(n) \in \mathbb{C}^L$,

$$\mathbf{y}_k(n) = \mathbf{H}_k \mathbf{x}(n) + \mathbf{z}_k(n)$$

- $\mathcal{K}(n) \subseteq [K]$ with $\Omega \in \{t+1, \dots, t+L\}$ members ($|\mathcal{K}(n)| = \Omega$), and $n \in \left[\binom{K}{\Omega}\right]$,

Achievable DoF Analysis I

Theorem 1: Achievable Bound for the number of streams linearly decodable

Consider a MIMO-CC system with K users, CC gain t , L Tx-antennas, and G Rx-antennas. A subset \mathcal{K} of users with size $\Omega \leq K$, where $t + 1 \leq \Omega \leq t + L$ are served in each interval.

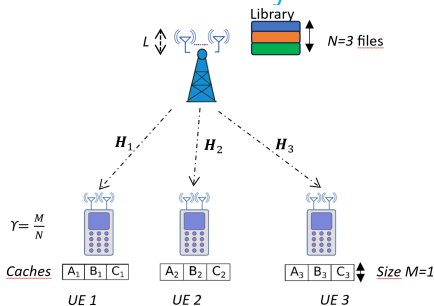
$$\beta \leq \min \left(G, (L - (\Omega - t - 1)\beta) \binom{\Omega - 1}{t} \right), \quad (1)$$

Every user in \mathcal{K} is able to decode β parallel streams interference-free.

Interference caused by every stream sent to user k can be removed:

- By cache content at up to t UEs
- Suppressed by Tx-side precoding for the rest of $\Omega - t - 1$ target users;
 - ⇒ The remaining spatial DoF : $L - (\Omega - t - 1)\beta$.
- Maximum value of β directly depends on sufficient L ;
 - To keep the direction of subpackets intended to the same null-space orthogonal
 - Otherwise; to keep the null-space of intended packets linearly

Achievable DoF Analysis I: Example



If $G = \beta = 2$, $\Omega = 3$, $t = 1$, then:

$$\mathbf{x} = (A_2 \oplus B_1)\mathbf{w}_{1,2} + (A_3 \oplus C_1)\mathbf{w}_{1,3} + (C_2 \oplus B_3)\mathbf{w}_{2,3}$$

$$y_{1,1} = \mathbf{u}_{1,1}^H \mathbf{H}_1 \mathbf{w}_{1,2} (A_2 \oplus B_1) + \mathbf{u}_{1,1}^H \mathbf{H}_1 \mathbf{w}_{1,3} (A_3 \oplus C_1) + \mathbf{u}_{1,1}^H \mathbf{H}_1 \mathbf{w}_{2,3} (C_2 \oplus B_3) + z_{1,1},$$

$$y_{1,2} = \mathbf{u}_{1,2}^H \mathbf{H}_1 \mathbf{w}_{1,2} (A_2 \oplus B_1) + \mathbf{u}_{1,2}^H \mathbf{H}_1 \mathbf{w}_{1,3} (A_3 \oplus C_1) + \mathbf{u}_{1,2}^H \mathbf{H}_1 \mathbf{w}_{2,3} (C_2 \oplus B_3) + z_{1,2},$$

- Subpackets delivered using linearly independent multicast beamformers,
 $\Rightarrow \mathbf{w}_{1,2} \in \text{Null}(\mathbf{U}_3^H \mathbf{H}_3)$, and $\mathbf{w}_{1,3} \in \text{Null}(\mathbf{U}_2^H \mathbf{H}_2)$;
 $\min_L (2 \times (L - 2)) \rightarrow L = 3$ is sufficient, and Thus, $\beta = 2 \times (3 - 2) = 2$.

Achievable DoF Analysis II

Corollary 1: Achievable DoF

The DoF of $\beta\Omega$ is necessarily achievable in every given MIMO setup, as long as β and Ω satisfy the given condition in Theorem 1. Using Theorem 1, the maximum achievable DoF for the proposed MIMO-CC transmission design is given by solving

$$\begin{aligned} \text{DoF}_{\max}(\beta^*, \Omega^*) &= \max_{\beta, \Omega} \Omega\beta, \\ \text{s.t. } \beta &\leq \min\left(G, \frac{L \binom{\Omega-1}{t}}{1 + (\Omega-t-1) \binom{\Omega-1}{t}}\right), \end{aligned} \quad (2)$$

where β^* and Ω^* represent the optimal parameters chosen to achieve DoF_{\max} .

Corollary 2: Simplified Achievable DoF

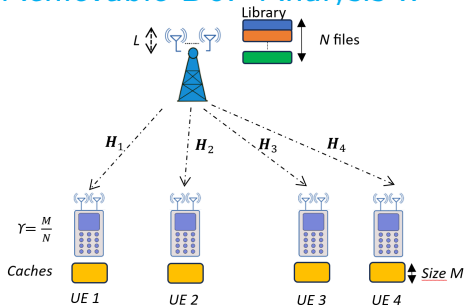
One can obtain the upperbound for Ω w.r.t β in MIMO setup from Theorem 1, and search across one dimension.

$$\text{DoF}_{\max}(\beta^*) = \max_{\beta \leq G} \left(\beta \left\lceil \frac{L}{\beta} \right\rceil + \beta t\right) \quad (3)$$

where the bound for Ω is:

$$\Omega \leq \left\lceil \frac{L}{\beta} \right\rceil + t \quad (4)$$

Achievable DoF Analysis II

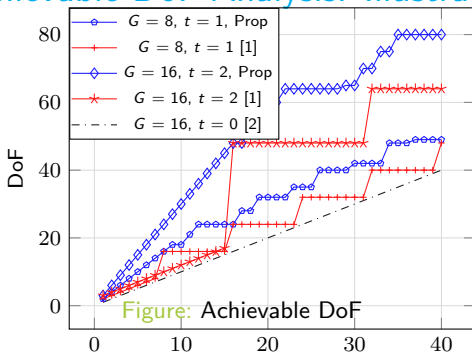


$$\text{DoF}_{\max}(\beta^*) = \max_{\beta \leq G} (\beta \lceil \frac{L}{\beta} \rceil + \beta t)$$

Given Theorem 1 and $G = 2, t = 1, L = 3$, Then;

- $\Omega = 2, \beta \leq \min(2, \frac{3 \binom{2-1}{1}}{1+(2-1-1)\binom{2-1}{1}}) = 1 \Rightarrow \text{DoF} = (1 \lceil \frac{3}{1} \rceil + 1) = 4$
- $\Omega = 3, \beta \leq \min(2, \frac{3 \binom{3-1}{1}}{1+(3-1-1)\binom{3-1}{1}}) = 2 \Rightarrow \text{DoF} = (2 \lceil \frac{3}{2} \rceil + 2) = 6 = \text{DoF}_{\max}$
- $\Omega = 4, \beta \leq \min(2, \frac{3 \binom{4-1}{1}}{1+(4-1-1)\binom{4-1}{1}}) = 1 \Rightarrow \text{DoF} = (1 \lceil \frac{3}{1} \rceil + 1) = 4$

Achievable DoF Analysis: Illustration



- The proposed scheme relaxes the integer constraint on $\frac{L}{G}$
- Gain boost of MIMO-CC setups: with smaller TX-side SM gains than [3]³.
- This flexibility becomes evident when enhancing the RX SM per UE
- Proposed DoF consistently achieves higher levels v.s MIMO scheme ($t = 0$)

³E. Parrinello, "Fundamental Limits of Coded Caching with Multiple Antennas, Shared Caches and Uncoded Prefetching," IEEE Trans. Inf. Theory, vol. 66, no. 4, pp. 2252-2268, 2020.

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General Transmission Design for MIMO-CC using Covariance Matrices

- Consider a generic multicast transmission signal vector $\mathbf{x}(i)$, built as

$$\mathbf{x}(i) = \sum_{\mathcal{T} \subseteq \mathcal{K}(i), |\mathcal{T}|=t+1} \tilde{\mathbf{x}}_{\mathcal{T}}(i), \forall i, \quad (5)$$

- $\tilde{\mathbf{x}}_{\mathcal{T}}(i)$ is the corresponding multicast signal at time instant i , chosen from a complex Gaussian codebook $\tilde{\mathbf{x}}_{\mathcal{T}} \sim \mathcal{CN}(\mathbf{0}, \mathbf{K}_{\tilde{\mathbf{x}}_{\mathcal{T}}})$.
- $\mathbf{x} = \sum_{\mathcal{T} \subseteq \mathcal{K}, |\mathcal{T}|=t+1} \tilde{\mathbf{x}}_{\mathcal{T}}$, and $\mathbf{K}_{\mathbf{x}} = \sum_{\mathcal{T} \subseteq \mathcal{K}, |\mathcal{T}|=t+1} \mathbf{K}_{\tilde{\mathbf{x}}_{\mathcal{T}}}$.
- This allows us to rewrite (10) as follows:

$$\mathbf{y}_k = \mathbf{H}_k \sum_{\mathcal{T} \in \mathcal{S}_k^{\mathcal{K}}} \tilde{\mathbf{x}}_{\mathcal{T}} + \mathbf{H}_k \sum_{\mathcal{T} \in \bar{\mathcal{S}}_k^{\mathcal{K}}} \tilde{\mathbf{x}}_{\mathcal{T}} + \mathbf{z}_k, \quad (6)$$

General Transmission Design for MIMO-CC using Covariance Matrices: Example

- Given $L = 4$, $\Omega = 3$ users, $t = 1$ (i.e., DoF = 6), and files $A-C$ requested by users 1-3, respectively. The MC transmission vector is

$$\mathbf{x} = \mathbf{x}_{1,2} + \mathbf{x}_{1,3} + \mathbf{x}_{2,3}, \quad (7)$$

- Each signal is proportional $\mathbf{x}_{1,2} \propto (A_2 \oplus B_1)$, $\mathbf{x}_{1,3} \propto (A_3 \oplus C_1)$, $\mathbf{x}_{2,3} \propto (B_3 \oplus C_2)$.⁴
- Then, the received signals at users 1-3:

$$\mathbf{y}_1 = \underline{\mathbf{H}}_1 \mathbf{x}_{1,2} + \underline{\mathbf{H}}_1 \mathbf{x}_{1,3} + \underline{\mathbf{H}}_1 \mathbf{x}_{2,3} + \mathbf{z}_1,$$

$$\mathbf{y}_2 = \underline{\mathbf{H}}_2 \mathbf{x}_{1,2} + \underline{\mathbf{H}}_2 \mathbf{x}_{2,3} + \underline{\mathbf{H}}_2 \mathbf{x}_{1,3} + \mathbf{z}_2,$$

$$\mathbf{y}_3 = \underline{\mathbf{H}}_3 \mathbf{x}_{1,3} + \underline{\mathbf{H}}_3 \mathbf{x}_{2,3} + \underline{\mathbf{H}}_3 \mathbf{x}_{1,2} + \mathbf{z}_3,$$

- UE 1, both $\mathbf{x}_{1,2}$ and $\mathbf{x}_{1,3}$: desired signals, and $\mathbf{x}_{2,3}$: Gaussian interference
 $\Rightarrow \mathbf{y}_1$: an equivalent G-MAC channel, equal rate decoding of its msgs

$$R_{MAC}^1 = \min \left(R_{\{1,2\}}, R_{\{1,3\}}, \frac{1}{2} R_{\{\{1,2\}, \{1,3\}\}} \right), \quad (8)$$

⁴As a special case, in (7), the transmission vector $\mathbf{x} = (A_2 \oplus B_1)\mathbf{w}_{1,2} + (A_3 \oplus C_1)\mathbf{w}_{1,3} + (B_3 \oplus C_2)\mathbf{w}_{2,3}$ [4].

MULTIGROUP MULTICAST TRANSMISSION

Symmetric rate

Assuming that all requests for files are served \Rightarrow The worst-case delivery rate.

$$R_{sym} = \frac{K}{\sum_n T_n} = \frac{K \binom{K}{t} \binom{K-t-1}{\Omega-t-1}}{\sum_n \frac{1}{R_{\max-\min}(n)}}, \quad (9)$$

The goal: Design transmission parameters, R_{sym} is maximized in the delivery scheme.

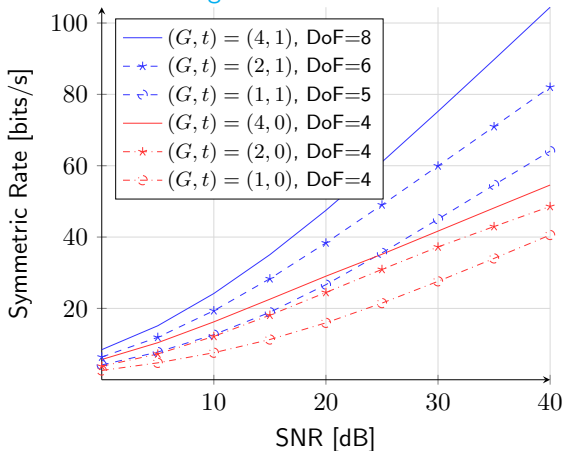
Max-Min Rate Optimization Problem

$$R_{\max-\min}(n) = \max_{\mathbf{K}_{\mathbf{x}_T}, R_T, T \in \mathcal{B}} \min_{k \in \mathcal{K}} \min_{\mathcal{B} \subseteq S_k^{\mathcal{K}}} \left[\frac{1}{|\mathcal{B}|} \sum_{T \in \mathcal{B}} R_T \right]$$

$$s.t. \quad \sum_{T \in \mathcal{B}} R_T \leq \log |\mathbf{I} + \mathbf{H}_k \sum_{T \in \mathcal{B}} \mathbf{K}_{\mathbf{x}_T} \mathbf{H}_k^H \mathbf{Q}_k^{-1}|, \quad \forall k \in \mathcal{K}, \mathcal{B} \subseteq S_k^{\mathcal{K}}$$

$$\sum_{T \in S^{\mathcal{K}}} \text{Tr}(\mathbf{K}_{\mathbf{x}_T}) \leq P_T, \quad \text{where} \quad \mathbf{Q}_k = (N_0 \mathbf{I} + \mathbf{H}_k \sum_{T \in S^{\mathcal{K}}} \mathbf{K}_{\mathbf{x}_T} \mathbf{H}_k^H)$$

General Transmission Design for MIMO-CC Cov Matrices: Simulation



- Covariance of the transmission signal, aiming for improved performance while achieving the enhanced DoF in Theorem 1.
- DoF scales with t . and G . when $L = 4$ for all setups.

General Transmission Design for MIMO-CC Cov Matrices: Simulation

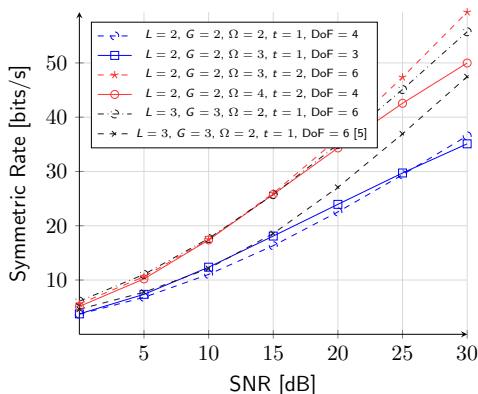


Figure: MIMO multicast design for $L = G$

- Cov-based design achieves the enhanced DoF, if params selected based on Theorem 1.
- Proposed approach achieves a DoF boost even with smaller Tx-spatial SM.

General Transmission Design for MIMO-CC Cov Matrices: Simulation

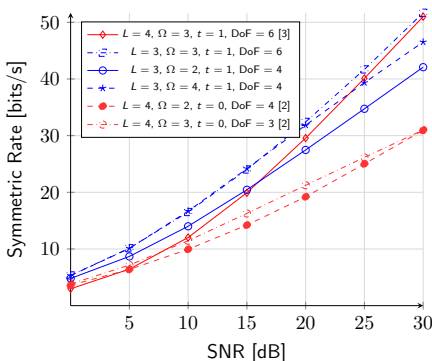


Figure: MIMO multicast design for $L > G - G = 2, t = \{1, 0\}$

- Compare Extended-Shared cache model, achieving DoF $Gt + L$ for divisible ratio of $\frac{L}{G}$
 $\Rightarrow L = 3, t = 1, G = \beta = 2, \text{DoF} = 6$, while [3] requires higher Tx SM $L = 4^5$,

⁵E. Parrinello, "Fundamental Limits of Coded Caching with Multiple Antennas, Shared Caches and Uncoded Prefetching," IEEE Trans. Inf. Theory, vol. 66, no. 4, pp. 2252-2268, 2020.

Conclusion and Future Work

- We proposed a flexible CC scheme for MIMO setups
 - Optimizing the number of users served in each transmission to maximize the achievable DoF.
- Proposed a high-performance beamformer design for MIMO-CC setups
 - Formulating the problem of maximizing the symmetric rate w.r.t transmit covariance matrices for the multicast signals.
 - Solved the non-convex problem using SCA and verified the performance enhancement through simulations.
- Optimality Proof of the proposed bound; (Converse Theorem)
- Generalized form of the proposed DoF with different SM per UE G_k ;

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References

- [1] M. J. Salehi, E. Parrinello, S. P. Shariatpanahi, P. Elia, and A. Tolli, "Low-Complexity High-Performance Cyclic Caching for Large MISO Systems," vol. 21, no. 5, pp. 3263–3278, 2022.
- [2] E. Biglieri, R. Calderbank, A. Constantinides, A. Goldsmith, A. Paulraj, and H. V. Poor, *MIMO wireless communications*. Cambridge university press, 2007.
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- [4] A. Tolli, S. P. Shariatpanahi, J. Kaleva, and B. Khalaj, "Multicast Beamformer Design for Coded Caching," vol. 2018-June, 6, pp. 1914–1918.
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