



Sequential Bayesian Algorithms for Identification and Blind Equalization of Unit-Norm Channels

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1. Introduction

- In many problems the unknowns reside on **spherical manifolds**.
- We employ **Fisher-Bingham** (F-B) prior distributions to the unknowns.
- The F-B priors form a **conjugate model** that yields closed-form, **recursive** estimates, naturally constrained to the spherical manifold.
- We apply this model to a communication setup with multiple **gain-controlled** FIR channels, deriving a MAP **channel estimator** and a **blind equalizer** based on Rao-Blackwellized particle filters.

2. Problem Formulation

- Single transmitter, R receivers.
- Let $\{b_n\}$ be the **bits** and $\{x_n\}$ the corresponding differentially encoded **symbols**.
- Signal **observed** at the r -th receiver:

$$y_{r,n} = \mathbf{h}_r^T \mathbf{x}_n + v_{r,n}, \quad (1)$$

where $\mathbf{x}_t \triangleq [x_t \dots x_{t-L+1}]^T$, $v_{r,t} \sim \mathcal{N}(0; \sigma_r^2)$, and $\mathbf{h}_r \in \mathbb{R}^{L \times 1}$ is the (time-invariant) impulse response of the **channel** to the r -th receiver.

- The **random** quantities \mathbf{h}_r , $\{\mathbf{x}_n\}$, and $\{v_{r,n}\}$, $r \in \{1, \dots, R\}$, are presumed to be **mutually independent** a priori.
- Due to **automatic gain control** (AGC), $\|\mathbf{h}_r\| = 1$; to take this constraint into account, we assume that \mathbf{h}_r has an L -variate **F-B** prior

$$p(\mathbf{h}_r) = \mathcal{FB}(\mathbf{h}_r | \mathbf{a}_{r,0}, \mathbf{B}_{r,0}) \triangleq C(\mathbf{a}_{r,0}, \mathbf{B}_{r,0})^{-1} \times \exp(\mathbf{h}_r^T \mathbf{B}_{r,0} \mathbf{h}_r + \mathbf{h}_r^T \mathbf{a}_{r,0}) \mathcal{I}_{\|\mathbf{h}_r\|=1}, \quad (2)$$

- where $\mathbf{a}_{r,0} \in \mathbb{R}^L$ and $\mathbf{B}_{r,0} \in \mathbb{R}^{L \times L}$ are the **hyperparameters**, and $C(\mathbf{a}_{r,0}, \mathbf{B}_{r,0})$ is the **normalization** constant.
- Accurate numerical estimates to $C(\mathbf{a}_{r,0}, \mathbf{B}_{r,0})$ can be computed via the so-called **saddlepoint** approximations [1] (see Appendix).

3. Main Result

- As a consequence of (1) and (2), the **posterior distribution** of \mathbf{h}_r given $\mathbf{X}_n \triangleq \{\mathbf{x}_1, \dots, \mathbf{x}_n\}$ and $Y_{r,n} \triangleq \{y_{r,1}, \dots, y_{r,n}\}$ is **F-B**, i.e.,

$$p(\mathbf{h}_r | \mathbf{X}_n, Y_{r,n}) = \mathcal{FB}(\mathbf{h}_r | \mathbf{a}_{r,n}, \mathbf{B}_{r,n}), \quad (3)$$

where the parameters $\mathbf{a}_{r,n}$ and $\mathbf{B}_{r,n}$ can be **recursively** determined via

$$\mathbf{B}_{r,n} = \mathbf{B}_{r,n-1} - \mathbf{x}_n \mathbf{x}_n^T / 2\sigma_r^2, \quad (4)$$

$$\mathbf{a}_{r,n} = \mathbf{a}_{r,n-1} + \mathbf{x}_n y_{r,n} / \sigma_r^2. \quad (5)$$

- Sketch of the Proof:

– Exploiting Markovian properties, it follows that:

$$p(\mathbf{h}_r | \mathbf{X}_n, Y_{r,n}) = \frac{p(y_{r,n} | \mathbf{x}_n, \mathbf{h}_r) p(\mathbf{h}_r | \mathbf{X}_{n-1}, Y_{r,n-1})}{\int_{\mathbf{h}_r \in \mathcal{S}^{L-1}} p(y_{r,n} | \mathbf{x}_n, \mathbf{h}_r) p(\mathbf{h}_r | \mathbf{X}_{n-1}, Y_{r,n-1}) d\mathcal{S}^{L-1}(\mathbf{h}_r)}$$

– Induction hypothesis:

$$p(\mathbf{h}_r | \mathbf{X}_{n-1}, Y_{r,n-1}) = \mathcal{FB}(\mathbf{h}_r | \mathbf{a}_{r,n-1}, \mathbf{B}_{r,n-1}).$$

– Observing that $p(y_{r,n} | \mathbf{x}_n, \mathbf{h}_r) = \mathcal{N}(y_{r,n} | \mathbf{h}_r^T \mathbf{x}_n; \sigma_r^2)$, we get after some algebra that

$$p(y_{r,n} | \mathbf{x}_n, \mathbf{h}_r) p(\mathbf{h}_r | \mathbf{X}_{n-1}, Y_{r,n-1}) = \frac{1}{\sqrt{2\pi\sigma_r^2}} \times \frac{C(\mathbf{a}_{r,n}, \mathbf{B}_{r,n})}{C(\mathbf{a}_{r,n-1}, \mathbf{B}_{r,n-1})} \exp\left(-\frac{y_{r,n}^2}{2\sigma_r^2}\right) \mathcal{FB}(\mathbf{h}_r | \mathbf{a}_{r,n}, \mathbf{B}_{r,n}). \quad (6)$$

4. Channel Identification

- The local **trained MAP estimate** of the channel parameters at the r -th receiver is given by

$$\hat{\mathbf{h}}_{r,n} \triangleq \arg \max_{\mathbf{h}_r} p(\mathbf{h}_r | \mathbf{X}_n, Y_{r,n}). \quad (7)$$

- The **constrained** optimization problem (7) can be recast as

$$\hat{\mathbf{h}}_{r,n} = \arg \max_{\mathbf{h}_r} \exp(\mathbf{h}_r^T \mathbf{B}_{r,n} \mathbf{h}_r + \mathbf{h}_r^T \mathbf{a}_{r,n}),$$

subject to $\|\mathbf{h}_r\|_2 = 1$

as the normalization constant $C(\mathbf{a}_{r,n}, \mathbf{B}_{r,n})$ does not depend on \mathbf{h}_r .

- The corresponding **Lagrange function** is given by

$$\Lambda(\mathbf{h}_r, \lambda) \triangleq \exp(\mathbf{h}_r^T \mathbf{B}_{r,n} \mathbf{h}_r + \mathbf{h}_r^T \mathbf{a}_{r,n}) + \lambda(\mathbf{h}_r^T \mathbf{h}_r - 1). \quad (8)$$

- Taking the **gradient** of (8) with respect to \mathbf{h}_r , dividing the result by the exponential term and equating it to zero, it follows that

$$\hat{\mathbf{h}}_{r,n} = -\frac{1}{2} (\mathbf{B}_{r,n} + \tilde{\lambda} \mathbf{I})^{-1} \mathbf{a}_{r,n}, \quad (9)$$

where $\tilde{\lambda} \triangleq \lambda \exp(-\hat{\mathbf{h}}_{r,n}^T \mathbf{B}_{r,n} \hat{\mathbf{h}}_{r,n} - \hat{\mathbf{h}}_{r,n}^T \mathbf{a}_{r,n})$.

- As $\mathbf{B}_{r,n}$ is symmetrical, we can plug its **eigenvalue decomposition** $\mathbf{B}_{r,n} \triangleq \mathbf{U}_{r,n} \mathbf{D}_{r,n} \mathbf{U}_{r,n}^T$ into (9) and rewrite the constraint $\|\hat{\mathbf{h}}_{r,n}\|_2^2 = 1$ as

$$\mathbf{a}_{r,n}^T \mathbf{U}_{r,n} (\mathbf{D}_{r,n} + \tilde{\lambda} \mathbf{I})^{-2} \mathbf{U}_{r,n}^T \mathbf{a}_{r,n} = 4. \quad (10)$$

- Equation (10) is equivalent to

$$\sum_{k=1}^L \left(\frac{[\mathbf{a}_{r,n}^T \mathbf{U}_{r,n}]^{[k]}}{[\mathbf{D}_{r,n}]^{[k]} + \tilde{\lambda}} \right)^2 = 4 \quad (11)$$

which can be rewritten as a $2L$ degree **polynomial equation** and numerically solved for $\tilde{\lambda}$.

- We verified experimentally that the most negative real $\tilde{\lambda}$ that solves (11) *almost* always leads to the MAP solution when replaced in (9).

5. Multichannel Blind Equalization

- We wish now to obtain the **joint MAP estimate**

$$\hat{b}_n = \arg \max_{b_n} p(b_n | Y_n),$$

that employs the observations available at all receivers, $Y_n \triangleq \{Y_{1,n}, \dots, Y_{r,n}\}$, via a **particle filtering** method

$$p(\mathbf{X}_n | Y_n) \approx \sum_{p=1}^P w_n^{(p)} \mathcal{I}_{\{\mathbf{X}_n = \mathbf{X}_n^{(p)}\}},$$

where P denotes the number of particles $\mathbf{X}_n^{(p)}$ and $w_n^{(p)}$ the normalized weights.

- Adopting the **prior importance function**, the particles are extended as $\mathbf{x}_n^{(p)} \sim p(\mathbf{x}_n | \mathbf{x}_{n-1}^{(p)})$ and the weights updated as

$$w_n^{(p)} \propto w_{n-1}^{(p)} p(y_n | \mathbf{X}_n^{(p)}, Y_{n-1}).$$

- The **a priori** independence of the channel parameters and noise samples at each receiver imply that

$$p(y_n | \mathbf{X}_n^{(p)}, Y_{n-1}) = \prod_{r=1}^R p(y_{r,n} | \mathbf{X}_n^{(p)}, Y_{r,n-1}).$$

- Likewise, **conditional independence** relations induced by the model result that

$$p(y_{r,n} | \mathbf{X}_n^{(p)}, Y_{r,n-1}) = \int_{\mathbf{h}_r \in \mathcal{S}^{L-1}} p(y_{r,n} | \mathbf{x}_n^{(p)}, \mathbf{h}_r) \times p(\mathbf{h}_r | \mathbf{X}_{n-1}^{(p)}, Y_{r,n-1}) d\mathcal{S}^{L-1}(\mathbf{h}_r). \quad (12)$$

- The integrand in (12) is identical to (6), which depends on \mathbf{h}_r only through a F-B density. Therefore

$$p(y_{r,n} | \mathbf{X}_n^{(p)}, Y_{r,n-1}) = \frac{1}{\sqrt{2\pi\sigma_r^2}} \cdot \frac{C(\mathbf{a}_{r,n}^{(p)}, \mathbf{B}_{r,n}^{(p)})}{C(\mathbf{a}_{r,n-1}^{(p)}, \mathbf{B}_{r,n-1}^{(p)})} \exp\left(-\frac{y_{r,n}^2}{2\sigma_r^2}\right), \quad (13)$$

where $\mathbf{a}_{r,n}^{(p)}$ and $\mathbf{B}_{r,n}^{(p)}$ are defined as in (4)-(5) replacing \mathbf{x}_n with $\mathbf{x}_n^{(p)}$.

6. Simulation Results

- To evaluate the performance of the proposed algorithms, we ran numerical experiments with $L = 3$. We drew \mathbf{h}_r by sampling independently in each realization from a **uniform distribution** on the unit sphere. Accordingly, the **hyperparameters** $\mathbf{a}_{r,0}$ and $\mathbf{B}_{r,0}$ were set to zero.

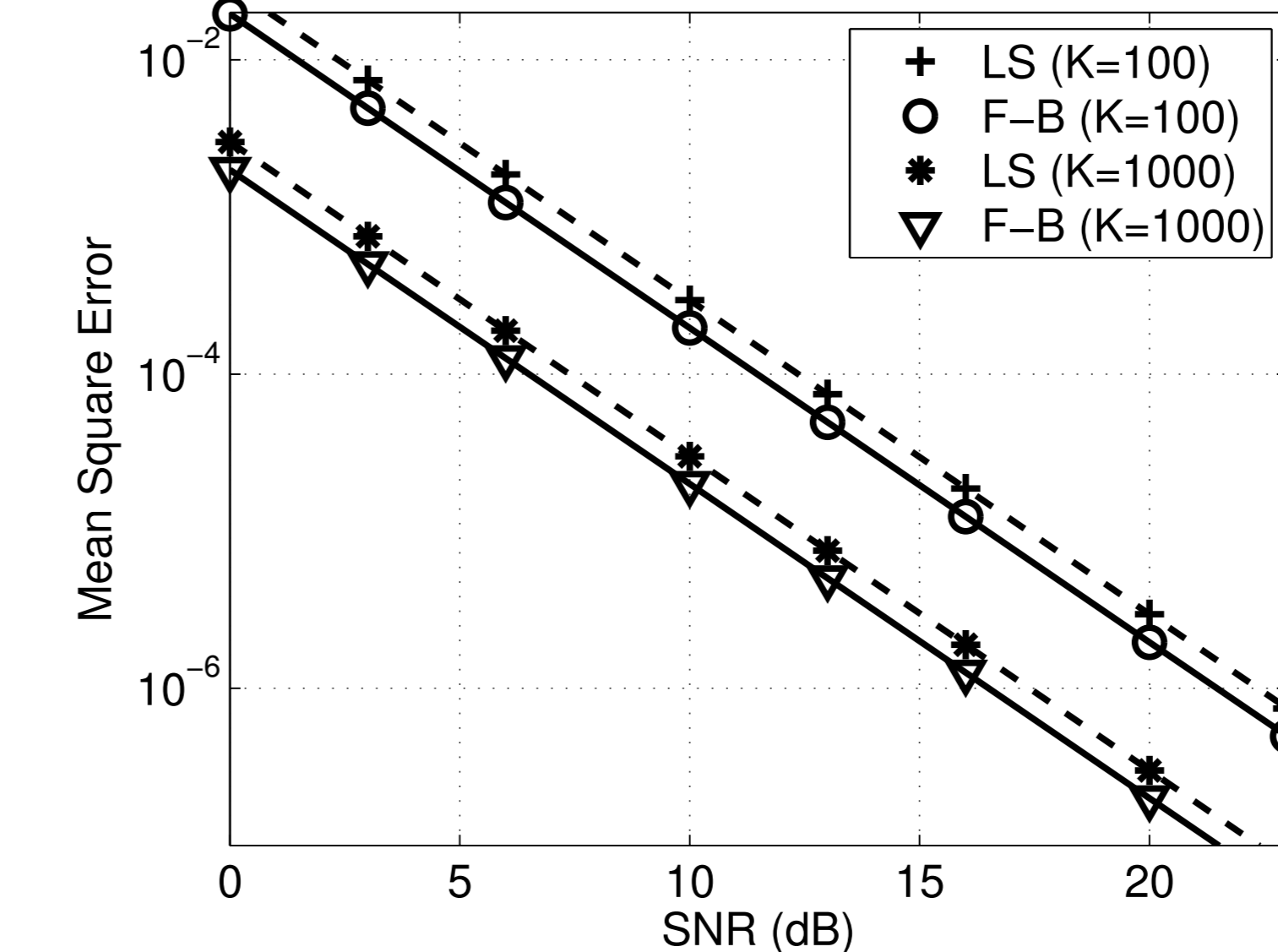


Figure 1: Average m.s.e. in the identification of the parameter \mathbf{h}_r as a function of the number of observed samples (K) and the SNR.

- As one may verify (Figure 1), the algorithm employing **F-B prior** led to a lower m.s.e. (about 66% of the LS estimator m.s.e.) for all SNR levels; this remains true when $K = 1, 000$.
- To assess the mean BER of the proposed **blind equalization** algorithm, the SNR was set equal on all receivers.
- In each realization, an i.i.d. sequence of 250 differentially encoded binary symbols was transmitted, being the first 150 bits discarded. The simulated system has $R = 4$. \mathbf{h}_r was sampled independently for each r . The particle filter uses $P = 300$ and performs systematic resampling at each time step.
- Figure 2 displays the **mean BER** of the proposed algorithm (F-B) and that of the equivalent method that employs **mismatched** Gaussian priors [2, Sec. 3.1].

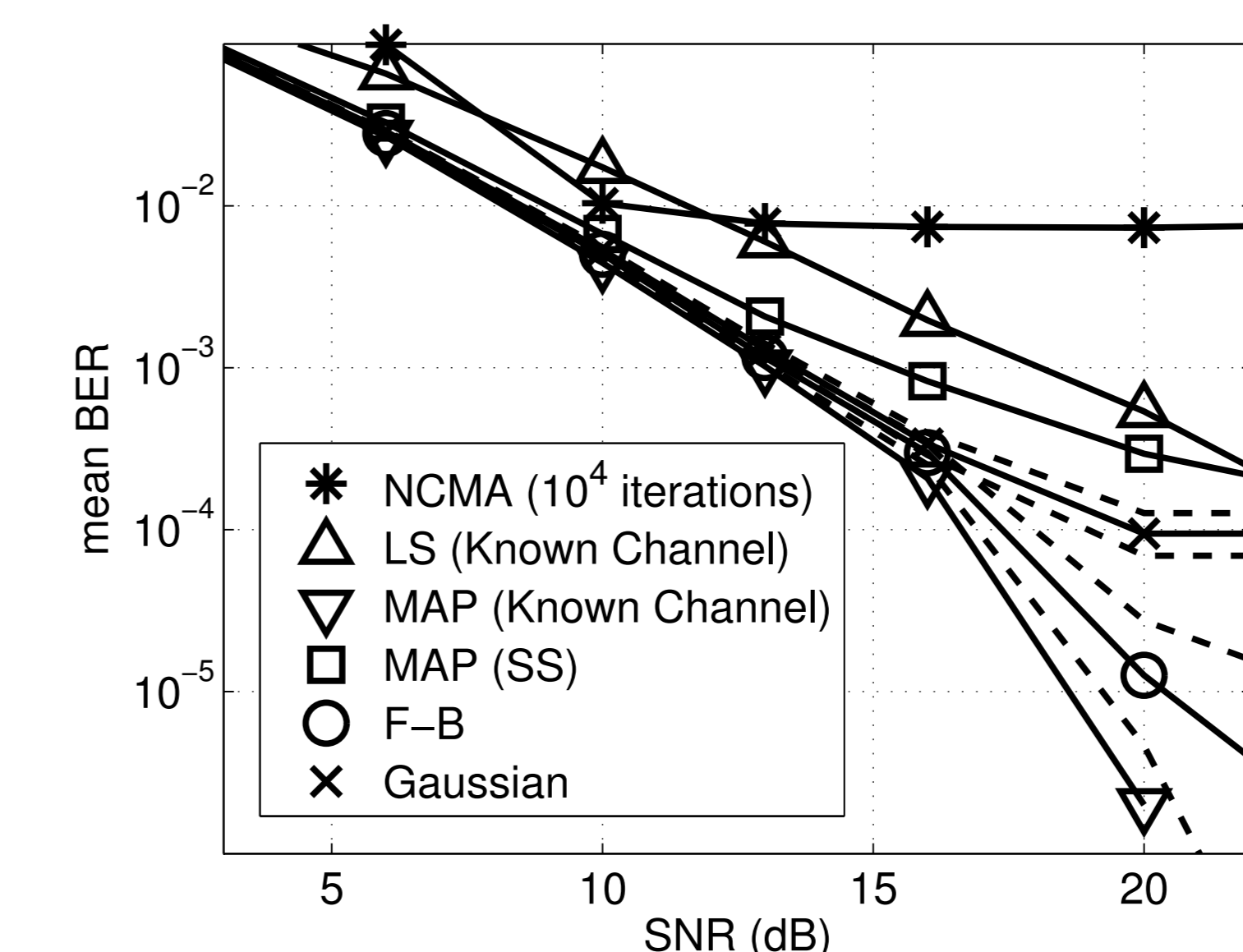


Figure 2: Mean BER estimated in 5,000 Monte Carlo runs as a function of the SNR. The dashed lines surrounding the solid ones display 95% confidence intervals.

- As one may note, the mean BER of the new (F-B) algorithm is equivalent to the optimal MAP equalizer and Gaussian particle filter method for low SNR. For high SNR, the new method **outperforms** the others.

7. Conclusions

- We introduced new algorithms for channel identification and blind equalization using a **Fisher-Bingham** prior model for the unknown parameters.
- F-B priors lead to a **conjugate model** that results in closed-form expressions for the parameters of the posterior densities, dropping with the need for **approximations** performed by previous works that employed sphere-constrained distributions.
- As we assessed via Monte Carlo simulations, the new channel identification and blind equalization algorithms outperformed conventional algorithms that adopt **mismatched Gaussian priors**, at the cost of increased computational complexity.

References

- [1] A. Kume, "Saddlepoint approximations for the Bingham and Fisher-Bingham normalising constants," *Biometrika*, n. 2, vol. 92, pp. 465-476, 2005.
- [2] C. J. Bordin Jr. and M. G. S. Bruno, "Consensus-Based Distributed Particle Filtering Algorithms for Cooperative Blind Equalization in Receiver Networks," *Proc. of ICASSP*, pp. 3968-3971, May 2011.

Appendix: Saddlepoint Approximation

- Kume [1] developed a method to compute the normalization constant

$$C(\mathbf{a}, \mathbf{B}) = \int_{\mathbf{h} \in \mathcal{S}^{L-1}} \exp(\mathbf{h}^T \mathbf{B} \mathbf{h} + \mathbf{h}^T \mathbf{a}) d\mathcal{S}^{L-1}(\mathbf{h})$$

that exploits the fact that the F-B density arises when an L -variable Gaussian r.v. \mathbf{x} is conditioned to have unit norm.

- Assuming (without loss of generality) that B is diagonal and introducing the change of variables $\nu \triangleq \mathbf{x}^T \mathbf{x}$ and $\mathbf{h} \triangleq \mathbf{x} / \nu^{1/2}$, it follows that $C(\mathbf{a}, \mathbf{B})$ depends on $p(\nu)$ (i.e., the p.d.f. of ν), evaluated at $\nu = 1$.

- As ν can be shown to be a linear combination of noncentral χ_1^2 r.v.'s, there is a closed-form expression for $K(t)$, the **cumulant generating function** of the distribution of ν , from which one can derive the **saddlepoint approximation**

$$\hat{p}(\nu) \triangleq (2\pi K''(\hat{t}))^{-\frac{1}{2}} \exp(K(\hat{t}) - \hat{t}\nu),$$

where \hat{t} denotes the solution to $K'(\hat{t}) = \nu$, and $K'(\hat{t})$ and $K''(\hat{t})$ are the 1st and 2nd derivatives of $K(t)$, respectively.