

# Global and Local Mode-domain Adaptive Algorithms for Spatial Active Noise Control Using Higher-order Sources

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## 1. Background & Abstract

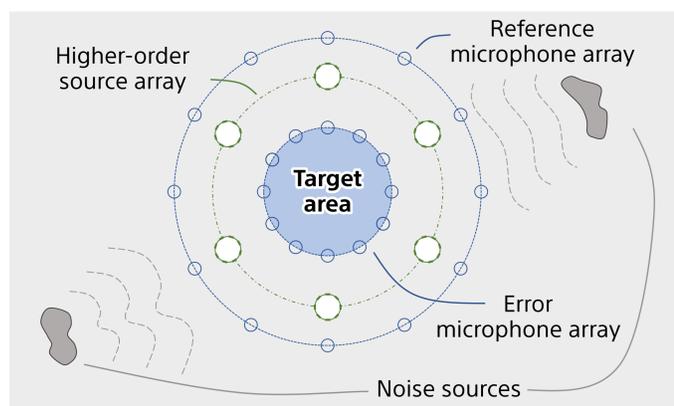
- **Spatial active noise control (ANC)** aims to attenuate noise over a certain space
- A large-scale system is required to achieve it
- **A higher-order source (HOS)** [1] has advantages in sound field control because it has controllable directivity patterns and occupies a smaller physical space

- Spatial ANC methods using **HOSs** are proposed
- Proposed methods are based on **the mode-domain signal processing** [2] which achieves fast convergence and a low computational cost

## 2. Problem Formulation

### 2.1. Array configuration

- Three concentric equiangular transducer arrays
- An objective of spatial ANC here is to attenuate noise in **the target area**



### 2.2. Harmonic representation of sound field

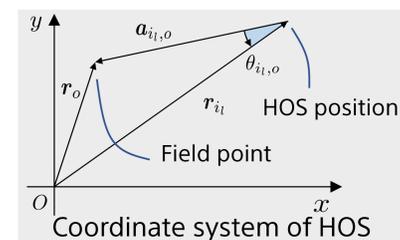
- The target sound field is represented using **the harmonic representation**

$$p(\mathbf{r}) = \sum_{m_g=-M_g}^{M_g} J_{m_g}(kr) \gamma_{m_g}(k) e^{-jm_g\phi}$$

Target sound field      Global mode coefficients

### 2.3. Higher-order sources

- The sound field generated by **HOSs** [1] (which are located at  $\mathbf{r}_{i1}, \dots, \mathbf{r}_{iL}$ ) is derived



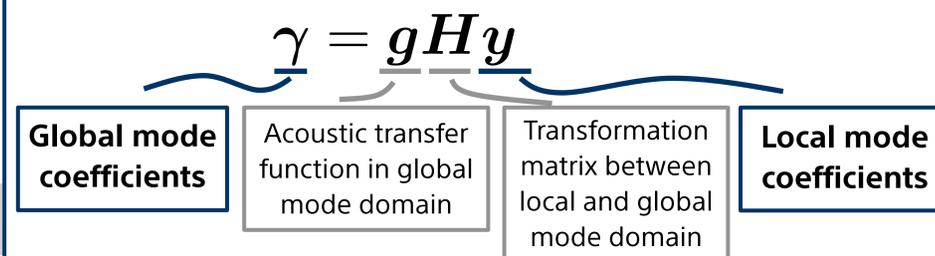
$$p(\mathbf{r}_o) = \sum_{i, m_i} \underline{y}_{m_i, i_l} H_{m_i}^{(2)}(ka_{i, o}) e^{-jm_i \theta_{i, o}}$$

Cylindrical addition theorem

$$= \sum_{m_g, i_l, m_i} \underline{y}_{m_i, i_l} H_{m_g+m_i}^{(2)}(kr_l) J_{m_g}(kr_o) e^{-jm_g(\phi_o - \phi_{i_l})}$$

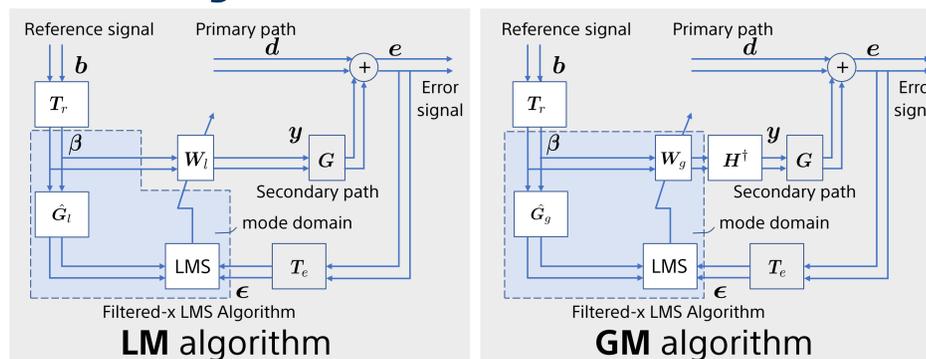
Local mode coefficients of each HOS

Relationship between global mode coefficients and local mode coefficients



## 3. Proposed Algorithms

### Block diagrams and error function



- Both algorithms minimize the instantaneous squared error of **the global mode coefficients**

$$J_g(\epsilon(n)) = \epsilon^H(n)\epsilon(n) \simeq \text{Approximation value of total error in target area [2]}$$

### 3.1. Local mode-domain adaptive algorithm (LM)

System model

$$\epsilon = \gamma + T_e G W_l \beta$$

Filter update rule derived by LMS

$$W_l(n+1) = W_l(n) - \mu \hat{G}_l^H \epsilon(n) \beta^H(n) \text{ where } \hat{G}_l = T_e \hat{G}$$

### 3.2. Global mode-domain adaptive algorithm (GM)

System model

$$\epsilon = \gamma + T_e G H^T W_g \beta$$

Filter update rule derived by LMS

$$W_g(n+1) = W_g(n) - \mu \hat{G}_g^H \text{diag}(\epsilon(n) \circ \overline{\beta(n)})$$

where  $\hat{G}_g = \text{diag}((T_e \hat{G} H^T)_{-M_g, -M_g}, \dots, (T_e \hat{G} H^T)_{M_g, M_g})$

Estimated secondary path in global mode domain

### 3.3. Comparison between proposed algorithms

Computational cost

Filtering: **GM** ≤ **LM** < MIMO

Filter update: **GM** ≪ **LM** < MIMO

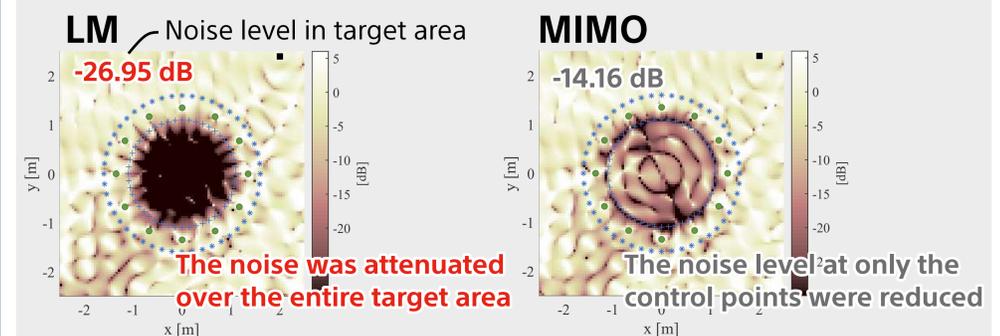
HOS array configuration

- LM algorithm does not necessarily require a **circular equiangular** HOS array

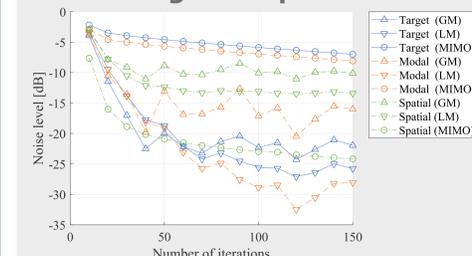
## 4. Experiment

- **LM, GM, and MIMO** (baseline) were compared

Noise level in dB after 500 iterations (at 500Hz)

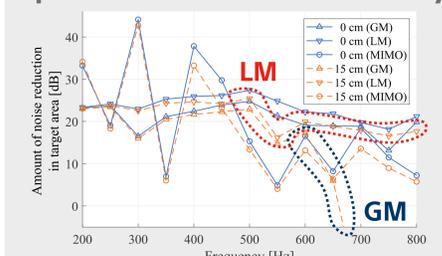


Comparison of convergence speed



**GM and LM achieved rapid convergence**

Robustness against positional deviation of arrays



**LM had robustness against the deviation**

Reference:

[1] M. A. Poletti, et al., *IEEE ICASSP*, 2011. [2] J. Zhang, et al., *IEEE TASLP*, 2018.