

# Optimal Local Thresholds for Distributed Detection in Energy Harvesting Wireless Sensor Networks

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- Introduction
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- The designs of wireless sensor networks to perform the task of distributed detection are often based on the conventional battery-powered sensors, leading into designs with a short lifetime, due to battery depletion.
- Energy harvesting, which can collect energy from renewable resources of environment (e.g., solar, wind, and geothermal energy) promises a self-sustainable system with a lifetime.



# System Model

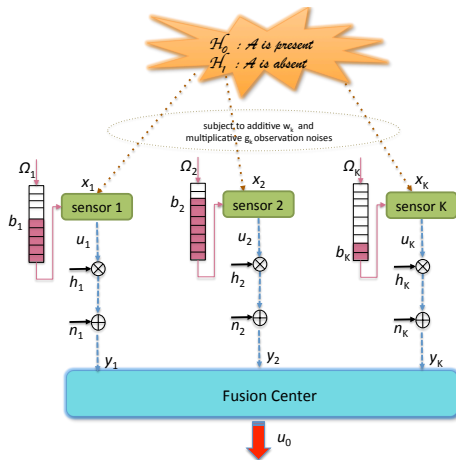


Figure 1: Our System model during one observation period.



Let  $x_k$  denote the local observation at sensor  $k$ :

$$x_k = \begin{cases} g_k \mathcal{A} + w_k & \mathcal{H}_1 \\ w_k & \mathcal{H}_0 \end{cases} \quad (1)$$

- $\mathcal{A}$  is a known scalar signal
- $w_k \sim \mathcal{N}(0, \sigma_{w_k}^2)$   $\rightarrow$  Additive noise
- $g_k \sim \mathcal{N}(0, \gamma_{g_k})$   $\rightarrow$  Multiplicative noise
- All observation noises are independent over time and among  $K$  sensors.



# System Model

During each observation period, sensor  $k$  takes  $N$  samples of  $x_k$  to measure the received signal energy and applies an energy detector to make a binary decision, i.e., sensor  $k$  decides whether or not signal  $\mathcal{A}$  is present.

$$\Lambda_k = \frac{1}{N} \sum_{n=1}^N |x_{k,n}|^2 \underset{d_k=0}{\overset{d_k=1}{\geq}} \theta_k \quad (2)$$

- $P_{f_k} = \Pr(\Lambda_k > \theta_k | \mathcal{H}_0) = \frac{\Gamma(N/2, \frac{N\theta_k}{\sigma_{w_k}^2})}{\Gamma(N/2)}$
- $P_{d_k} = \Pr(\Lambda_k > \theta_k | \mathcal{H}_1) = Q_{N/2}\left(\frac{\sqrt{\eta_k}}{\sigma_{w_k}}, \frac{\sqrt{N\theta_k}}{\sigma_{w_k}}\right)$
- **Our goal is optimize the local decision threshold  $\theta_k$**



## Assumptions:

- Each sensor is able to harvest energy from the environment and stores it in a battery with the capacity  $\mathcal{K}$  units of energy.
- The sensors communicate with the FC through orthogonal fading channels with channel gains  $|h_k|$ 's with parameters  $\gamma h_k$ .
- The sensors employ on-off keying signaling.
- We use the channel-inversion power, the number of energy units spent to convey a decision be inversely proportional to  $|h_k|$ .
- To avoid the battery depletion when  $|h_k|$  is too small, we impose an extra constraint for channel quality.



Let  $u_{k,t}$  represent the sensor output corresponding to the observation period  $t$ .

$$u_{k,t} = \begin{cases} \lceil \frac{\lambda}{|h_k|} \rceil & \Lambda_k > \theta_k, b_{k,t} > \lceil \frac{\lambda}{|h_k|} \rceil, |h_k|^2 > \zeta_k \\ 0 & \text{Otherwise} \end{cases} \quad (3)$$

- $b_{k,t}$  denote the battery state of sensor  $k$
- $|h_k|$  is channel gain
- $\zeta_k$  is threshold of the channel quality
- $\lambda$  is a power regulation constant





We model  $b_{k,t}$  in (3) as the following

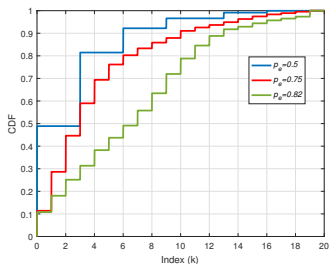
$$b_{k,t} = \min\left\{b_{k,t-1} - \left\lceil \frac{\lambda}{|h_k|} \right\rceil I_{u_{k,t-1}} + \Omega_{k,t}, \mathcal{K}\right\} \quad (4)$$

- $\Omega_{k,t} \in \{0, 1\}$  indicates units of harvesting energy a Bernoulli random variable, with  $\Pr(\Omega_{k,t} = 1) = p_e$
- $I_{u_{k,t-1}} = \begin{cases} 1 & u_{k,t-1} > 0 \\ 0 & \text{Otherwise} \end{cases}$

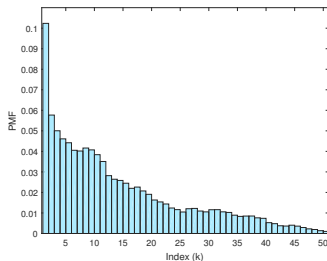


# Battery Model

- Assuming  $b_k$  in (4) is a stationary random process, one can compute the CDF and the pmf of  $b_k$  in terms of  $\mathcal{K}$ ,  $p_e$ ,  $\gamma_{h_k}$ . Further, we use pmf of  $b_k$  for our numerical results.



(a)



(b)

Figure 2: (a) CDF of  $b_k$  for  $\mathcal{K}=20$  and  $p_e=0.5, 0.75, 0.82$ , (b) pmf of  $b_k$  for  $\mathcal{K}=50$  and  $p_e=0.8$ .



We consider two detection performance metrics to find the optimal  $\theta_k$ 's:

- The detection probability at the FC, assuming that the FC utilizes the optimal fusion rule based on Neyman-Pearson optimality criterion.
- the KL distance between the two distributions of the received signals at the FC conditioned on hypothesis  $\mathcal{H}_0, \mathcal{H}_1$



# Optimal LRT Fusion Rule and $P_D, P_F$ Expressions

The received signal at the FC from sensor  $k$  is  $y_k = h_k u_k + n_k$ , where the additive communication channel noise  $n_k \sim \mathcal{N}(0, \sigma_{n_k}^2)$ . The likelihood ratio at the FC is

$$\Delta_{\text{LRT}} = \sum_{k=1}^K \log \left( \frac{\sum_{u_k} f(y_k | u_k, \mathcal{H}_1) \Pr(u_k | \mathcal{H}_1)}{\sum_{u_k} f(y_k | u_k, \mathcal{H}_0) \Pr(u_k | \mathcal{H}_0)} \right) \quad (5)$$

Given  $u_k, y_k$  is Gaussian, i.e.,  $y_k | u_k=0 \sim \mathcal{N}(0, \sigma_{n_k}^2)$  and  $y_k | u_k=\lceil \frac{\lambda}{|h_k|} \rceil \sim \mathcal{N}(\lceil \frac{\lambda}{|h_k|} \rceil h_k, \sigma_{n_k}^2)$ .



# Optimal LRT Fusion Rule and $P_D, P_F$ Expressions

The probabilities  $\Pr(u_k|\mathcal{H}_1)$ ,  $\Pr(u_k|\mathcal{H}_0)$  in (5) are

- $\Pr(u_k = \lceil \frac{\lambda}{|h_k|} \rceil | \mathcal{H}_1) = P_{d_k} \rho_k q_k = \alpha_k$
- $\Pr(u_k = \lceil \frac{\lambda}{|h_k|} \rceil | \mathcal{H}_0) = P_{f_k} \rho_k q_k = \beta_k$

where  $\rho_k = \Pr(b_k > \lceil \frac{\lambda}{|h_k|} \rceil)$  and  $q_k = \Pr(|h_k|^2 > \zeta_k)$ .

Given a threshold  $\tau$ , the optimal likelihood ratio test (LRT) is

$\Delta_{\text{LRT}} \underset{\mathcal{H}_0}{\underset{\mathcal{H}_1}{\geq}} \tau$ . The  $P_F, P_D$  at the FC

$$P_F = \Pr(\Delta_{\text{LRT}} > \tau | \mathcal{H}_0) = Q\left(\frac{\tau - \mu_{\Delta|\mathcal{H}_0}}{\sigma_{\Delta|\mathcal{H}_0}}\right) \quad (6)$$

$$\begin{aligned} P_D &= \Pr(\Delta_{\text{LRT}} > \tau | \mathcal{H}_1) \\ &= Q\left(\frac{Q^{-1}(a)\sigma_{\Delta|\mathcal{H}_0} + \mu_{\Delta|\mathcal{H}_0} - \mu_{\Delta|\mathcal{H}_1}}{\sigma_{\Delta|\mathcal{H}_1}}\right) \end{aligned}$$



Kullback-Leibler distance (KL) between the two distributions of the received signals at the FC

$$KL_k = \int_{y_k} f(y_k|\mathcal{H}_1) \log \left( \frac{f(y_k|\mathcal{H}_1)}{f(y_k|\mathcal{H}_0)} \right) dy_k \quad (8)$$

One can approximate  $KL_k$  in (8) by the KL distance of two Gaussian distributions

$$KL_k \approx \frac{1}{2} \log \left( \frac{\sigma_{y_k|\mathcal{H}_0}^2}{\sigma_{y_k|\mathcal{H}_1}^2} \right) + \frac{\sigma_{y_k|\mathcal{H}_1}^2 - \sigma_{y_k|\mathcal{H}_0}^2 + (\mu_{y_k|\mathcal{H}_1} - \mu_{y_k|\mathcal{H}_0})^2}{2\sigma_{y_k|\mathcal{H}_0}^2} \quad (9)$$



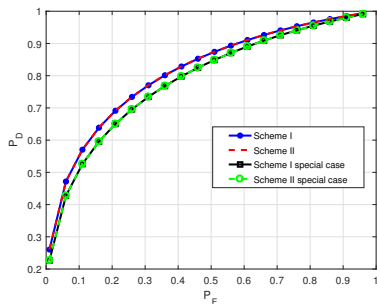
In this section, we consider:

- Case I: Numerically find  $\theta_k$ 's which maximize  $P_D$  in (7)  $\rightarrow$   $K$ -dimensional search is required  $\rightarrow$  computational complexity!
- Case II: Finding  $\theta_k$ 's which maximize  $KL_{tot} = \sum_{k=1}^K KL_k$ , using the  $KL_k$  approximations in (9)  $\rightarrow$  Only one dimensional search  $\rightarrow$  computationally efficient.
- Special cases: Assume all sensors employ the same local threshold  $\theta_k = \theta$ .

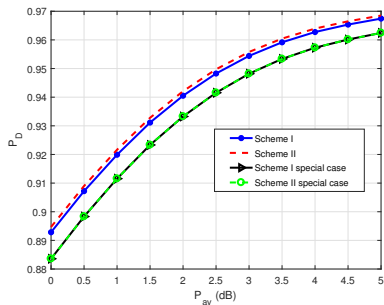
We then compare  $P_D$  evaluated at the  $\theta_k$ 's obtained from mentioned cases.



# Simulation results



(a)



(b)

Figure 3: (a)  $P_D$  vs.  $P_F$   
(b)  $P_D$  vs.  $P_{av}$





# Conclusion

- We studied a distributed detection problem in a wireless network with  $K$  heterogeneous energy harvesting sensors and investigated the optimal local decision thresholds for given transmission and battery state models.
- Our numerical results indicate that the thresholds obtained from maximizing the KL distance are near-optimal and computationally very efficient, as it requires only  $K$  one-dimensional searches, as opposed to a  $K$ -dimensional search required to find the thresholds that maximize the detection probability.
- The performance gap between each scheme and its corresponding special case indicates that when sensors are heterogeneous, it is advantageous to use different local thresholds according to sensors' statistics.



*Questions?*

