

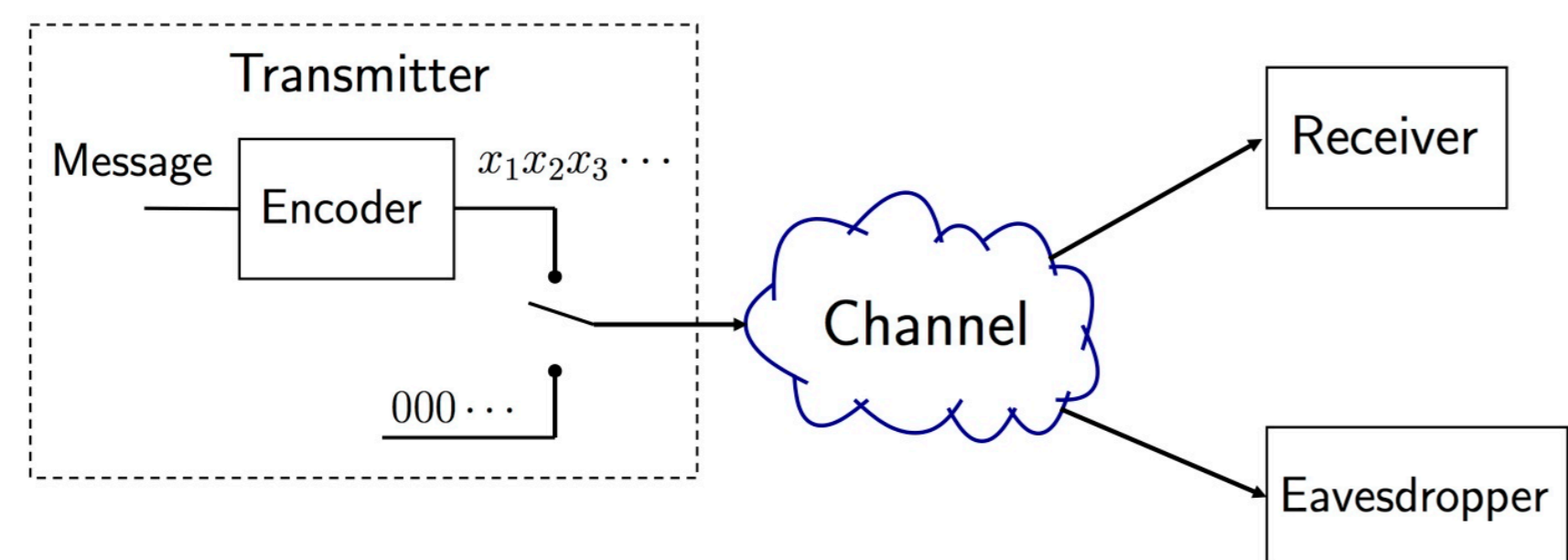
On Covert Communication Over Infinite-Bandwidth Gaussian Channels

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Covert Communication and Square-Root Law



Suppose

- Both Receiver and Eavesdropper observe same AWGN channel (with same noise power);
- Covert requirement is

$$D(P^n \| Q^n) \leq \delta$$

where P^n is average output distribution when Transmitter sends a codeword;
 Q^n is output distribution when Transmitter sends n zeros (pure Gaussian noise).

Then [1], [2]

$$\text{maximum number of nats over } n \text{ channel uses} = \sqrt{n\delta} + o(\sqrt{n}).$$

In particular, covert communication capacity (nats per channel use) is zero.

Infinite Bandwidth: Simple Heuristics

Over W Hz and T seconds with white Gaussian noise, one has $2WT$ independent samples.

⇒ Total number of nats $\propto \sqrt{WT}$.

⇒ Positive per-second rate possible if $W \gtrsim T$.

Formal Treatment in Continuous Time

Model 1: Input $X(\cdot)$ and output (at both Receiver and Eavesdropper) $Y(\cdot)$ are related by

$$Y(t) = X(t) + Z(t), \quad t \in \mathbb{R},$$

where $Z(\cdot)$ is a stationary Gaussian process to be further specified later.

- Transmitter is “approximately time-limited”: it maps a message to $x(t)$, $t \in \mathbb{R}$, such that

$$\lim_{T \rightarrow \infty} \frac{\int_0^T |x(t)|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} = 1.$$

- Decoder is strictly time-limited: it maps $y(t)$, $t \in [0, T]$, to decoded message.
- Eavesdropper is not time-limited: covertness constraint is

$$\lim_{T \rightarrow \infty} D(P_{-\infty}^{\infty} \| Q_{-\infty}^{\infty}) = 0,$$

where $P_{-\infty}^{\infty}$ and $Q_{-\infty}^{\infty}$ are resp. distributions of $Y(t)$ and $Z(t)$, $t \in (-\infty, \infty)$.

Proposition 1. Assume, for every T , the noise process $Z(\cdot)$ has PSD $N_0/2$ over $[-W_T, W_T]$, where $W_T = T^2$. Under the above conditions and power constraint

$$\mathbb{E} \left[\int_{-\infty}^{\infty} |X(t)|^2 dt \right] \leq PT,$$

the covert communication capacity of the channel is P/N_0 nats per second.

Prolate Spheroidal Wave Functions (PSWFs) [3], [4]:

There exist $1 > \lambda_1 > \lambda_2 > \dots > 0$ (countably infinite) and functions $\{\psi_i\}$ such that

1. Each ψ_i is band-limited to W Hz. Further, the functions $\{\psi_i\}$ are orthonormal on \mathbb{R} , and complete in the space of functions that are band-limited to W Hz.
2. The restrictions of $\{\psi_i\}$ to the interval $[0, T]$ are orthogonal:

$$\int_0^T \psi_i(t) \psi_j(t) dt = \begin{cases} \lambda_i, & i = j, \\ 0, & i \neq j. \end{cases}$$

Restrictions of $\{\psi_i\}$ to $[0, T]$ are complete in the space of square integrable functions on $[0, T]$.

3. For any $\epsilon \in (0, 1)$, as $WT \rightarrow \infty$,

$$\begin{aligned} \lambda_{2(1-\epsilon)WT} &\rightarrow 1 \\ \lambda_{2(1+\epsilon)WT} &\rightarrow 0. \end{aligned}$$

4. Let $Z(\cdot)$ be stationary Gaussian noise with PSD

$$N(f) = \begin{cases} \frac{N_0}{2}, & |f| \leq W, \\ 0, & |f| > W \end{cases}$$

restricted to the interval $[0, T]$, then \mathbf{Z} can be written in the Karhunen-Loève expansion

$$Z(t) = \sum_{i=1}^{\infty} Z_i \psi_i(t), \quad t \in [0, T],$$

where $\{Z_i\}$ are IID Gaussian random variables of mean zero and variance $N_0/2$.

Proof Sketch of Proposition 1: Fix $\epsilon \in (0, 1)$. Our coding scheme is to generate $2(1-\epsilon)T^3$ IID Gaussian random variables $\{X_i\}$ each of mean zero and variance $PT^{-2}/2$, and transmit the signal

$$X(t) = \sum_{i=1}^{(1-\epsilon)T^3} X_i \psi_i(t), \quad t \in \mathbb{R},$$

where $\{\psi_i\}$ are PSWFs for the frequency band $[-T^2, T^2]$ and time interval $[0, T]$. The proof then follows classic works [5], [6]; covertness, data rates, and other results all follow from the nice properties of the PSWFs. □

Band-Limited Noise: Good and Bad Models

Model 2: make the following changes from Model 1.

- Transmitter is strictly time-limited: $X(t) = 0$ w.p. 1 for all $t \notin [0, T]$.
- Eavesdropper is also time-limited: covertness constraint is

$$\lim_{T \rightarrow \infty} D(P_0^T \| Q_0^T) = 0.$$

Proposition 2. Let $Z(\cdot)$ have PSD that equals $N_0/2$ on $[-W, W]$ and zero elsewhere, where W is a constant that does not grow with T . Under Model 1, the covert communication capacity of the channel is zero. Under Model 2, the covert communication capacity is infinity.

Proof Sketch for Model 2: Fix interval $[0, T]$. For any positive integer k , generate a sequence of k^3 IID Gaussian random variables $\{X_i\}$ of mean zero and variance k^{-2} . Let

$$X(t) = \begin{cases} \sum_i X_i \psi_i(t), & t \in [0, T], \\ 0, & \text{otherwise.} \end{cases}$$

By the orthogonality of the PSWFs on $[0, T]$, the channel can be reduced, for both Eavesdropper and Receiver, to a set of k^3 parallel, independent Gaussian channels $Y_i = X_i + Z_i$. The claim then follows from the discrete-time AWGN results, and the fact that k can be arbitrarily large. □

Lesson: Model 2 is bad. When channel has memory (e.g., noise with limited bandwidth), covertness constraint must not be restricted to communication duration.

Colored Noise

Infinite-bandwidth white noise does not exist, as it would have infinite power. Consider colored Gaussian noise $Z(\cdot)$ with PSD $N(f) > 0$ for all $f \in \mathbb{R}$, symmetric around $f = 0$, satisfying

$$\int_{-\infty}^{\infty} N(f) df < \infty.$$

Let us choose the input signal $X(\cdot)$ to be generated from a stationary Gaussian process with PSD

$$S(f) = \begin{cases} T^{-7/4} \cdot N(f), & f \in [-W_T, W_T] \\ 0, & \text{otherwise,} \end{cases}$$

where again $W_T = T^2$. We then have [7]

$$D(P_{\mathbf{Y}} \| P_{\mathbf{Z}}) = T \cdot \frac{1}{2} \int_{-W_T}^{W_T} \left(\frac{S(f)}{N(f)} - \log \left(1 + \frac{S(f)}{N(f)} \right) \right) df \leq \frac{T^{-1/2}}{2},$$

which tends to zero as $T \rightarrow \infty$; while

$$\frac{1}{T} \cdot I(\mathbf{X}; \mathbf{Y}) = \int_{-W_T}^{W_T} \frac{1}{2} \log \left(1 + \frac{S(f)}{N(f)} \right) df \approx \frac{T^{1/4}}{2}$$

which tends to infinity as $T \rightarrow \infty$. Note also $\int_{-W_T}^{W_T} S(f) df \rightarrow 0$ as $T \rightarrow \infty$.

The following conjecture remains to be formulated and proven in a true continuous-time setting.

Conjecture 3. If $Z(\cdot)$ is Gaussian noise as above, then the covert communication capacity of the channel without bandwidth constraint on the input is infinity. Furthermore, this should hold irrespective of whether an average-power constraint is imposed on the input or not.

References

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