

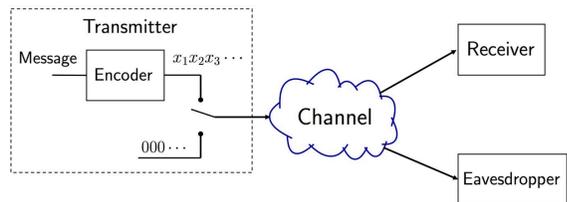
# On Covert Communication Over Infinite-Bandwidth Gaussian Channels

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## Covert Communication and Square-Root Law



Suppose

- Both Receiver and Eavesdropper observe same AWGN channel (with same noise power);
- Covert requirement is

$$D(P^n \| Q^n) \leq \delta$$

where  $P^n$  is average output distribution when Transmitter sends a codeword;

$Q^n$  is output distribution when Transmitter sends  $n$  zeros (pure Gaussian noise).

Then [1], [2]

$$\text{maximum number of nats over } n \text{ channel uses} = \sqrt{n\delta} + o(\sqrt{n}).$$

In particular, covert communication capacity (nats per channel use) is zero.

## Infinite Bandwidth: Simple Heuristics

Over  $W$  Hz and  $T$  seconds with white Gaussian noise, one has  $2WT$  independent samples.

⇒ Total number of nats  $\propto \sqrt{WT}$ .

⇒ Positive per-second rate possible if  $W \gtrsim T$ .

## Formal Treatment in Continuous Time

**Model 1:** Input  $X(\cdot)$  and output (at both Receiver and Eavesdropper)  $Y(\cdot)$  are related by

$$Y(t) = X(t) + Z(t), \quad t \in \mathbb{R},$$

where  $Z(\cdot)$  is a stationary Gaussian process to be further specified later.

- Transmitter is “approximately time-limited”: it maps a message to  $x(t)$ ,  $t \in \mathbb{R}$ , such that

$$\lim_{T \rightarrow \infty} \frac{\int_0^T |x(t)|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} = 1.$$

- Decoder is strictly time-limited: it maps  $y(t)$ ,  $t \in [0, T]$ , to decoded message.
- Eavesdropper is not time-limited: covertness constraint is

$$\lim_{T \rightarrow \infty} D(P_{-\infty}^{\infty} \| Q_{-\infty}^{\infty}) = 0,$$

where  $P_{-\infty}^{\infty}$  and  $Q_{-\infty}^{\infty}$  are resp. distributions of  $Y(t)$  and  $Z(t)$ ,  $t \in (-\infty, \infty)$ .

**Proposition 1.** Assume, for every  $T$ , the noise process  $Z(\cdot)$  has PSD  $N_0/2$  over  $[-W_T, W_T]$ , where  $W_T = T^2$ . Under the above conditions and power constraint

$$\mathbb{E} \left[ \int_{-\infty}^{\infty} |X(t)|^2 dt \right] \leq PT,$$

the covert communication capacity of the channel is  $P/N_0$  nats per second.

**Prolate Spheroidal Wave Functions (PSWFs)** [3], [4]:

There exist  $1 > \lambda_1 > \lambda_2 > \dots > 0$  (countably infinite) and functions  $\{\psi_i\}$  such that

1. Each  $\psi_i$  is band-limited to  $W$  Hz. Further, the functions  $\{\psi_i\}$  are orthonormal on  $\mathbb{R}$ , and complete in the space of functions that are band-limited to  $W$  Hz.
2. The restrictions of  $\{\psi_i\}$  to the interval  $[0, T]$  are orthogonal:

$$\int_0^T \psi_i(t) \psi_j(t) dt = \begin{cases} \lambda_i, & i = j, \\ 0, & i \neq j. \end{cases}$$

Restrictions of  $\{\psi_i\}$  to  $[0, T]$  are complete in the space of square integrable functions on  $[0, T]$ .

3. For any  $\epsilon \in (0, 1)$ , as  $WT \rightarrow \infty$ ,

$$\begin{aligned} \lambda_{2(1-\epsilon)WT} &\rightarrow 1 \\ \lambda_{2(1+\epsilon)WT} &\rightarrow 0. \end{aligned}$$

4. Let  $Z(\cdot)$  be stationary Gaussian noise with PSD

$$N(f) = \begin{cases} \frac{N_0}{2}, & |f| \leq W, \\ 0, & |f| > W \end{cases}$$

restricted to the interval  $[0, T]$ , then  $\mathbf{Z}$  can be written in the Karhunen-Loève expansion

$$Z(t) = \sum_{i=1}^{\infty} Z_i \psi_i(t), \quad t \in [0, T],$$

where  $\{Z_i\}$  are IID Gaussian random variables of mean zero and variance  $N_0/2$ .

*Proof Sketch of Proposition 1:* Fix  $\epsilon \in (0, 1)$ . Our coding scheme is to generate  $2(1-\epsilon)T^3$  IID Gaussian random variables  $\{X_i\}$  each of mean zero and variance  $PT^{-2}/2$ , and transmit the signal

$$X(t) = \sum_{i=1}^{(1-\epsilon)T^3} X_i \psi_i(t), \quad t \in \mathbb{R},$$

where  $\{\psi_i\}$  are PSWFs for the frequency band  $[-T^2, T^2]$  and time interval  $[0, T]$ . The proof then follows classic works [5], [6]; covertness, data rates, and other results all follow from the nice properties of the PSWFs. □

## Band-Limited Noise: Good and Bad Models

**Model 2:** make the following changes from Model 1.

- Transmitter is strictly time-limited:  $X(t) = 0$  w.p. 1 for all  $t \notin [0, T]$ .
- Eavesdropper is also time-limited: covertness constraint is

$$\lim_{T \rightarrow \infty} D(P_0^T \| Q_0^T) = 0.$$

**Proposition 2.** Let  $Z(\cdot)$  have PSD that equals  $N_0/2$  on  $[-W, W]$  and zero elsewhere, where  $W$  is a constant that does not grow with  $T$ . Under Model 1, the covert communication capacity of the channel is zero. Under Model 2, the covert communication capacity is infinity.

*Proof Sketch for Model 2:* Fix interval  $[0, T]$ . For any positive integer  $k$ , generate a sequence of  $k^3$  IID Gaussian random variables  $\{X_i\}$  of mean zero and variance  $k^{-2}$ . Let

$$X(t) = \begin{cases} \sum_i X_i \psi_i(t), & t \in [0, T], \\ 0, & \text{otherwise.} \end{cases}$$

By the orthogonality of the PSWFs on  $[0, T]$ , the channel can be reduced, for both Eavesdropper and Receiver, to a set of  $k^3$  parallel, independent Gaussian channels  $Y_i = X_i + Z_i$ . The claim then follows from the discrete-time AWGN results, and the fact that  $k$  can be arbitrarily large. □

**Lesson:** Model 2 is bad. When channel has memory (e.g., noise with limited bandwidth), covertness constraint must not be restricted to communication duration.

## Colored Noise

Infinite-bandwidth white noise does not exist, as it would have infinite power. Consider colored Gaussian noise  $Z(\cdot)$  with PSD  $N(f) > 0$  for all  $f \in \mathbb{R}$ , symmetric around  $f = 0$ , satisfying

$$\int_{-\infty}^{\infty} N(f) df < \infty.$$

Let us choose the input signal  $X(\cdot)$  to be generated from a stationary Gaussian process with PSD

$$S(f) = \begin{cases} T^{-7/4} \cdot N(f), & f \in [-W_T, W_T] \\ 0, & \text{otherwise,} \end{cases}$$

where again  $W_T = T^2$ . We then have [7]

$$D(P_{\mathbf{Y}} \| P_{\mathbf{Z}}) = T \cdot \frac{1}{2} \int_{-W_T}^{W_T} \left( \frac{S(f)}{N(f)} - \log \left( 1 + \frac{S(f)}{N(f)} \right) \right) df \leq \frac{T^{-1/2}}{2},$$

which tends to zero as  $T \rightarrow \infty$ ; while

$$\frac{1}{T} \cdot I(\mathbf{X}; \mathbf{Y}) = \int_{-W_T}^{W_T} \frac{1}{2} \log \left( 1 + \frac{S(f)}{N(f)} \right) df \approx \frac{T^{1/4}}{2}$$

which tends to infinity as  $T \rightarrow \infty$ . Note also  $\int_{-W_T}^{W_T} S(f) df \rightarrow 0$  as  $T \rightarrow \infty$ .

The following conjecture remains to be formulated and proven in a true continuous-time setting.

**Conjecture 3.** If  $Z(\cdot)$  is Gaussian noise as above, then the covert communication capacity of the channel without bandwidth constraint on the input is infinity. Furthermore, this should hold irrespective of whether an average-power constraint is imposed on the input or not.

## References

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